

CMOS 8-Bit Microcontroller

TMP86CM25F, TMP86CS25F

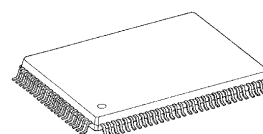
The TMP86CM25/S25 are the high-speed, high-performance and low power consumption 8-bit microcomputer, including ROM, RAM, Dot matrix LCD driver, multi-function timer/counter, serial interface (UART/SIO), a 8-bit AD converter and two clock generators on chip.

Product No.	ROM	RAM	Package	OTP MCU
TMP86CM25F	32 K × 8 bits	2 K × 8 bits	P-QFP100-1420-0.65A	TMP86PS25F
TMP86CS25F	60 K × 8 bits			

Features

- ◆ 8-bit single chip microcomputer TLCS-870/C series
- ◆ Instruction execution time: 0.25 μ s (at 16 MHz)
122 μ s (at 32.768 kHz)
- ◆ 132 types and 731 basic instructions
- ◆ 20 interrupt sources (External: 5, Internal: 15)
- ◆ Input/Output ports (42 pins)
(Out of which 20 pins are also used as SEG pins.)
- ◆ 18-bit timer counter: 1 ch
 - Timer, Event counter, Pulse width measurement, Frequency measurement modes
- ◆ 8-bit timer counter: 4 ch
 - Timer, Event counter, PWM output, Programmable Divider Output, PPG modes
- ◆ Time Base Timer
- ◆ Divider output function
- ◆ Watchdog Timer
 - Interrupt source/internal reset generate (programmable)

P-QFP100-1420-0.65A



TMP86CM25F
TMP86CS25F

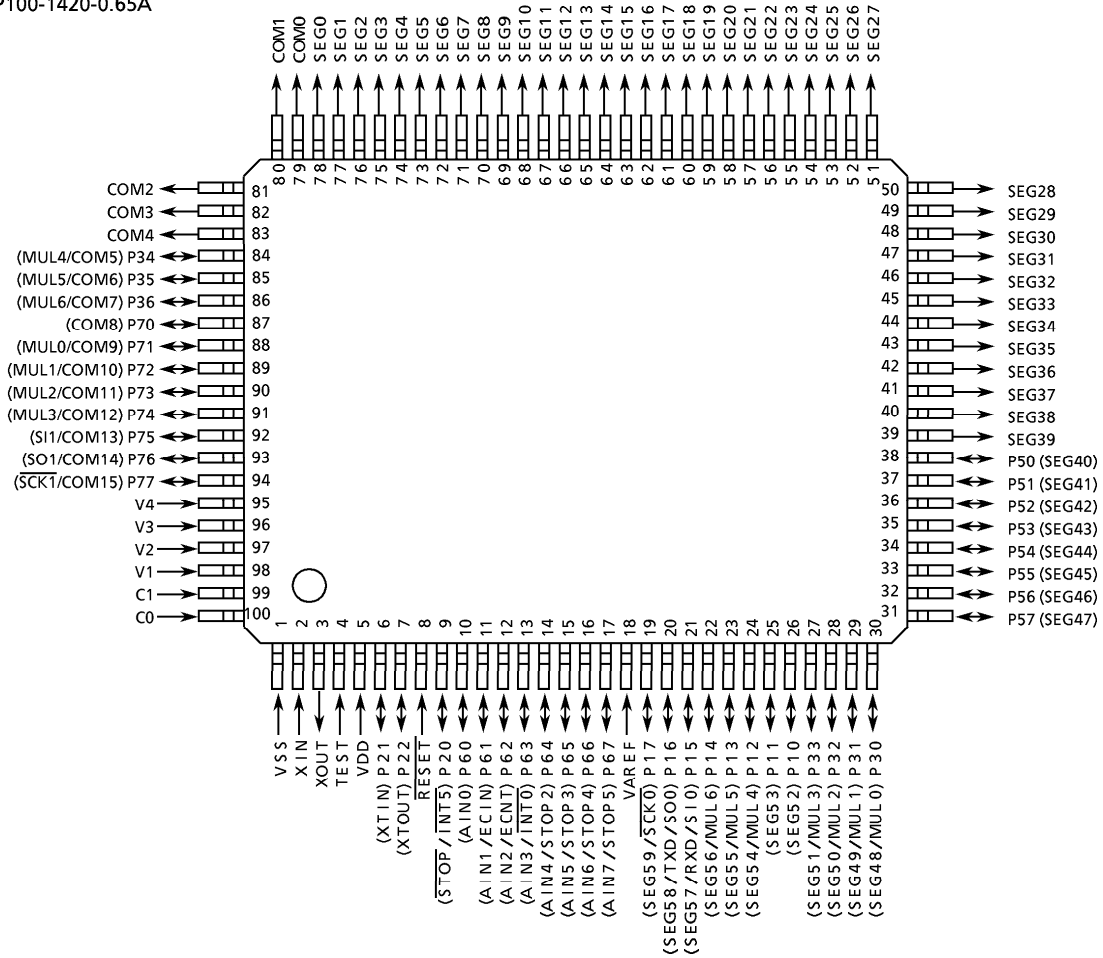
000707EBP1

- For a discussion of how the reliability of microcontrollers can be predicted, please refer to Section 1.3 of the chapter entitled Quality and Reliability Assurance / Handling Precautions.
- TOSHIBA is continually working to improve the quality and reliability of its products. Nevertheless, semiconductor devices in general can malfunction or fail due to their inherent electrical sensitivity and vulnerability to physical stress. It is the responsibility of the buyer, when utilizing TOSHIBA products, to comply with the standards of safety in making a safe design for the entire system, and to avoid situations in which a malfunction or failure of such TOSHIBA products could cause loss of human life, bodily injury or damage to property. In developing your designs, please ensure that TOSHIBA products are used within specified operating ranges as set forth in the most recent TOSHIBA products specifications. Also, please keep in mind the precautions and conditions set forth in the "Handling Guide for Semiconductor Devices," or "TOSHIBA Semiconductor Reliability Handbook" etc..
- The TOSHIBA products listed in this document are intended for usage in general electronics applications (computer, personal equipment, office equipment, measuring equipment, industrial robotics, domestic appliances, etc.). These TOSHIBA products are neither intended nor warranted for usage in equipment that requires extraordinarily high quality and/or reliability or a malfunction or failure of which may cause loss of human life or bodily injury ("Unintended Usage"). Unintended Usage include atomic energy control instruments, airplane or spaceship instruments, transportation instruments, traffic signal instruments, combustion control instruments, medical instruments, all types of safety devices, etc.. Unintended Usage of TOSHIBA products listed in this document shall be made at the customer's own risk.
- The products described in this document are subject to the foreign exchange and foreign trade laws.
- The information contained herein is presented only as a guide for the applications of our products. No responsibility is assumed by TOSHIBA CORPORATION for any infringements of intellectual property or other rights of the third parties which may result from its use. No license is granted by implication or otherwise under any intellectual property or other rights of TOSHIBA CORPORATION or others.
- The information contained herein is subject to change without notice.

- ◆ Serial interface: 2ch
 - 8-bit UART/SIO: 1ch
 - 8-bit SIO: 1ch
- ◆ 8-bit successive approximation type AD converter
 - Analog input: 8 ch
- ◆ Four Key On Wake Up pins
- ◆ LCD driver/controller
 - Built-in voltage booster for LCD driver
 - With displaymemory
 - LCD direct drive capability (60 seg × 16 com, 60 seg × 8 com, 60 seg × 4 com)
 - 1/16, 1/8, 1/4 duties drive are programmably selectable
- ◆ Dual clock operation
 - Single/Dual-clock mode
- ◆ Nine power saving operating modes
 - STOP mode: Oscillation stops. Battery/capacitor back-up. Port output hold/high-impedance.
 - SLOW 1, 2 mode: Low power consumption operation using low-frequency clock (32.768 kHz)
 - IDLE0 mode: CPU stops, and peripherals stop except Time-Base-Timer.
Release by falling edge of TBTCR < TBTCCK > setting.
 - IDLE 1 mode: CPU stops, and peripherals operate using high-frequency clock. Release by interrupts.
 - IDLE 2 mode: CPU stops, and peripherals operate using high and low frequency clock. Release by interrupts.
 - SLEEP 0 mode: CPU stops, and peripherals stop except Time-Base-Timer.
Release by falling edge of TBTCR < TBTCCK > setting.
 - SLEEP 1 mode: CPU stops, and peripherals operate using low-frequency clock. Release by interrupts.
 - SLEEP 2 mode: CPU stops, and peripherals operate using high and low frequency clock. Release by interrupts.
- ◆ Wide operating voltage: 1.8 to 5.5 V at 4.2 MHz/32.768 kHz,
2.7 to 5.5 V at 8 MHz/32.768 kHz,
4.5 to 5.5 V at 16 MHz/32.768 kHz

Pin Assignments (Top View)

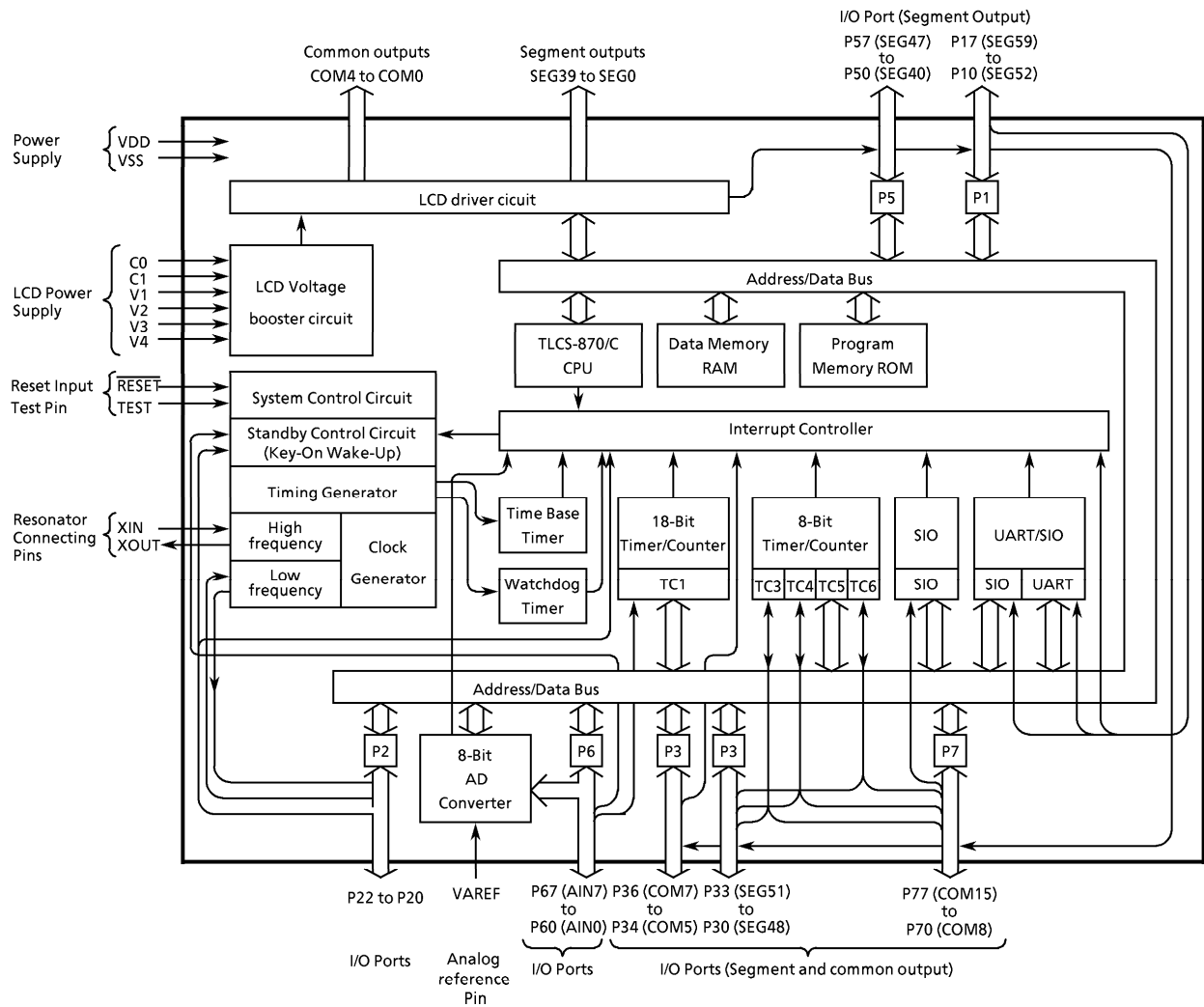
P-QFP100-1420-0.65A



Note: Ports assigned as MUL6 to MUL0 can switch pin assignment by the multifunction register (MULSEL). For functions assigned to each pin, see the table below.

Pin name	Function	Pin assignment
MUL0	\overline{DVO}	P30 or P71
MUL1	PWM3, $\overline{PDO3}$, TC3	P31 or P72
MUL2	PPG4, PWM4, $\overline{PDO4}$, TC4	P32 or P73
MUL3	PPG6, PWM6, $\overline{PDO6}$, TC6	P33 or P74
MUL4	INT1	P12 or P34
MUL5	INT2	P13 or P35
MUL6	INT3	P14 or P36

Block Diagram



Pin Function

Pin Name	Input/Output	Function	
P17 (SEG59, SCK0)	I/O (I/O)	8-bit input/output port with latch. When used as input port, an external interrupt input, serial clock input/output, serial data input/output or UART data input/output, the latch must be set to "1". When used as a LCD segment output, the P1LCR must be set to "1".	Serial clock input/Output
P16 (SEG58, TxD, SO0)	I/O (Output)		UART data output Serial data output
P15 (SEG57, RxD, S10)	I/O (I/O)		UART data input Serial data input
P14 (SEG56, MUL6)	I/O (I/O)		External interrupt 3 input
P13 (SEG55, MUL5)	I/O (I/O)		External interrupt 2 input
P12 (SEG54, MUL4)	I/O (I/O)		External interrupt 1 input
P11 (SEG53)	I/O (Output)		
P10 (SEG52)	I/O (Output)		
P22 (XTOUT)	I/O (Output)	3-bit input/output port with latch. When used as an input port, the latch must be set to "1".	Resonator connecting pins (32.768 kHz) For inputting external clock, XTIN is used and XTOUT is opened.
P21 (XTIN)	I/O (Input)		
P20 (INT5, STOP)	I/O (Input)		External interrupt input 5 or STOP mode release signal input
P36 (COM7, MUL6)	I/O (I/O)	7-bit I/O port with latch. When used as input port, an external interrupt input or timer/counter input/output, the latch must be set to "1". When used as a LCD segment output, the P3LCR must be set to "1".	External interrupt 3 input
P35 (COM6, MUL5)	I/O (I/O)		External interrupt 2 input
P34 (COM5, MUL4)	I/O (I/O)		External interrupt 1 input
P33 (SEG51, MUL3)	I/O (I/O)		Timer/counter 6 input/output
P32 (SEG50, MUL2)	I/O (I/O)		Timer/counter 4 input/output
P31 (SEG49, MUL1)	I/O (I/O)		Timer/counter 3 input/output
P30 (SEG48, MUL0)	I/O (Output)		Divider output
P57 (SEG16) to P50 (SEG23)	I/O (Output)		8-bit input/output port with latch. When used as a LCD segment output, the P5LCR must be set to "1".
P67 (AIN7, STOP5)	I/O (Input)	8-bit programmable input/output ports (tri-state). Each bit of these ports can be individually configured as an input or an output under software control. When used as a key on wake up input, an external interrupt input and timer/counter input, the P6CR must be set to "1".	STOP5 input
P66 (AIN6, STOP4)	I/O (Input)		STOP4 input
P65 (AIN5, STOP3)	I/O (Input)		STOP3 input
P64 (AIN4, STOP2)	I/O (Input)		STOP2 input
P63 (AIN3, INT0)	I/O (Input)		External interrupt 0 input
P62 (AIN2, ECNT)	I/O (Input)		
P61 (AIN1, ECIN)	I/O (Input)		Timer/counter 1 input
P60 (AIN0)	I/O (Input)		
P70 (COM8)	I/O (Output)	8-bit I/O port. When used common output, P7 port control register (P7LCR) should be set to 1.	Divider output
P71 (COM9, MUL0)	I/O (I/O)		Timer/counter 3 input/output
P72 (COM10, MUL1)	I/O (I/O)		Timer/counter 4 input/output
P73 (COM11, MUL2)	I/O (I/O)		Timer/counter 6 input/output
P74 (COM12, MUL3)	I/O (I/O)		
P75 (COM13, S11)	I/O (I/O)		Serial data input
P76 (COM14, SO1)	I/O (Output)		Serial data output
P77 (COM15, SCK1)	I/O (I/O)		Serial clock input/output
SEG39 to SEG0	Output	LCD segment outputs	
COM4 to COM0		LCD common outputs	
V4 to V1 C1 to C0	LCD voltage booster pin	LCD voltage booster pin. Capacitors are required between C0 and C1 pin and V1/V2/V3/V4 pin and GND.	
XIN, XOUT	Input Output	Resonator connecting pins for high-frequency clock. For inputting external clock, XIN is used and XOUT is opened.	
RESET	Input	Reset signal input	
TEST	Input	Test pin for out-going test. Be fixed to low.	
VDD, VSS	Power Supply	+ 5 V, 0 (GND)	
VAREF		Analog reference voltage input.	

Operational Description

1. CPU Core Functions

The CPU core consists of a CPU, a system clock controller, and an interrupt controller.

This section provides a description of the CPU core, the program memory, the data memory, the external memory interface, and the reset circuit.

1.1 Memory Address Maps

The TMP86CM25/S25 memory consist of 4 blocks: ROM, RAM, DBR (Data Buffer Register) and SFR (Special Function Register). They are all mapped in 64 Kbyte address space. Figure 1-1 shows the TMP86CM25/S25 memory address maps. The general-purpose registers are not assigned to the RAM address space.

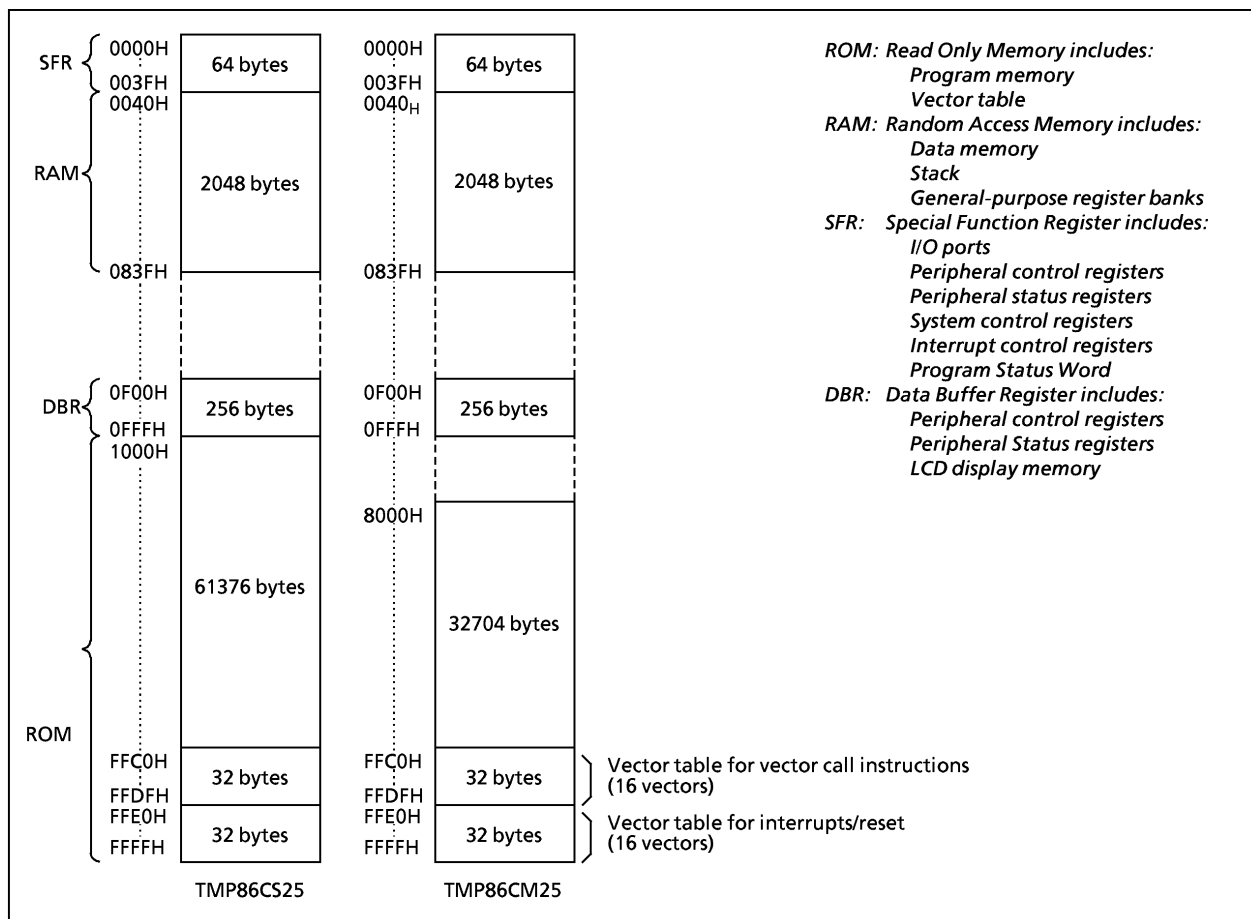


Figure 1-1. Memory Address Maps

1.2 Program Memory (ROM)

The TMP86CM25 has a 32 K×8-bit (address 8000H to FFFFH), and the TMP86CS25 has a 60 K×8-bit (address 1000H to FFFFH) of program memory (mask programmed ROM). However, placing program memory on the internal RAM is deregulated if a certain procedure is executed (See 2.5.5 Address Trap).

1.3 Data Memory (RAM)

Data memory consists of internal data memory (internal ROM or RAM). The TMP86CM25/S25 have 2 Kbytes of internal RAM. The first 192 bytes (0040H to 00FFH) of the internal RAM are located in the direct area; instructions with shorten operations are available against such an area.

The data memory contents become unstable when the power supply is turned on; therefore, the data memory should be initialized by an initialization routine.

Example: TMP86CM25/S25 RAM CLR

```
LD      HL, 0040H      ; Start address setup
LD      A, H           ; initial value (00H) setup
LD      BC, 07FFH     ; byte (-1) setup
SRAMCLR: LD      (HL), A
INC     HL
DEC     BC
JRS    F, SRAMCLR
```

1.4 System Clock Controller

The system clock controller consists of a clock generator, a timing generator, and a stand-by controller.

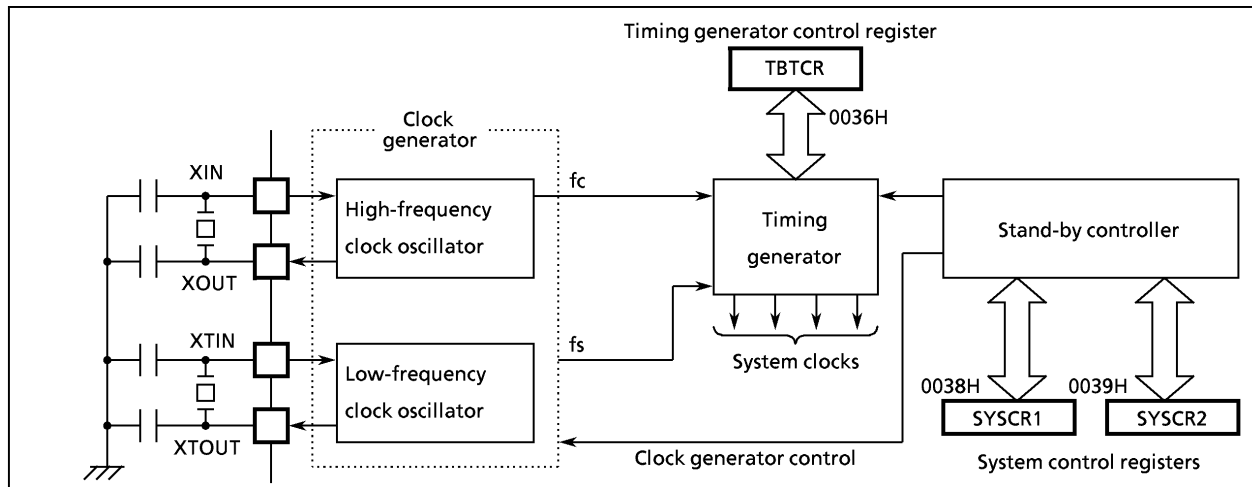


Figure 1-2. System Clock Control

1.4.1 Clock Generator

The Clock generator generates the basic clock which provides the system clocks supplied to the CPU core and peripheral hardware. It contains two oscillation circuits: one for the high-frequency clock and one for the low-frequency clock. Power consumption can be reduced by switching of the standby controller to low-power operation based on the low-frequency clock.

The high-frequency (f_c) and low-frequency (f_s) clocks can easily be obtained by connecting a resonator between the XIN/XOUT and XTIN/XTOUT pins respectively. Clock input from an external oscillator is also possible. In this case, external clock is applied to XIN/XTIN pin with XOUT/XOUT pin not connected.

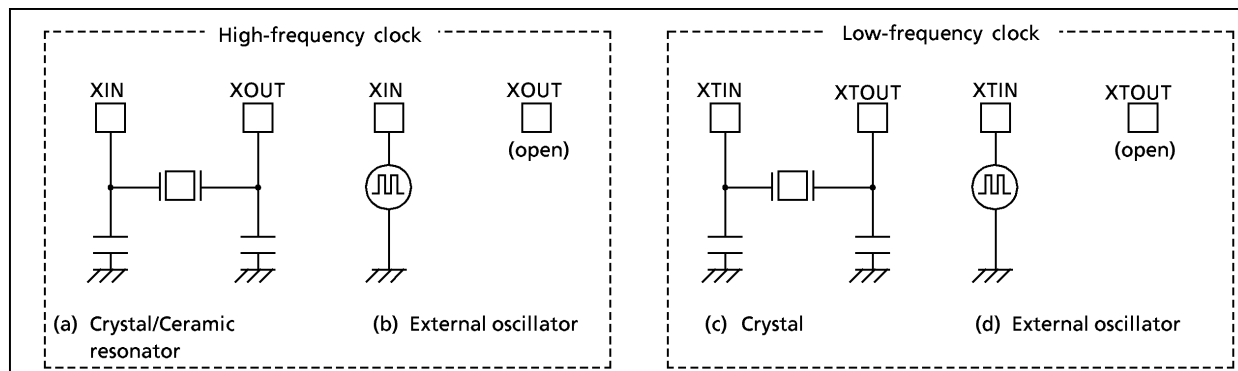


Figure 1-3. Examples of Resonator Connection

Note: The function to monitor the basic clock directly at external is not provided for hardware, however, with disabling all interrupts and watchdog timers, the oscillation frequency can be adjusted by monitoring the pulse which the fixed frequency is outputted to the port by the program.

The system to require the adjustment of the oscillation frequency should create the program for the adjustment in advance.

1.4.2 Timing Generator

The timing generator generates the various system clocks supplied to the CPU core and peripheral hardware from the basic clock (fc or fs). The timing generator provides the following functions.

- ① Generation of main system clock (fm)
- ② Generation of divider output (DVO) pulses
- ③ Generation of source clocks for time base timer
- ④ Generation of source clocks for watchdog timer
- ⑤ Generation of internal source clocks for timer/counters
- ⑥ Generation of warm-up clocks for releasing STOP mode

(1) Configuration of timing generator

The timing generator consists of a 2-stage prescaler, a 21-stage divider, a main system clock generator, and machine cycle counters.

An input clock to the 7th stage of the divider depends on the operating mode, DV7CK (bit 4 in TBTCR), that is shown in Figure 1-5. As reset and STOP mode started/canceled, the prescaler and the divider are cleared to "0".

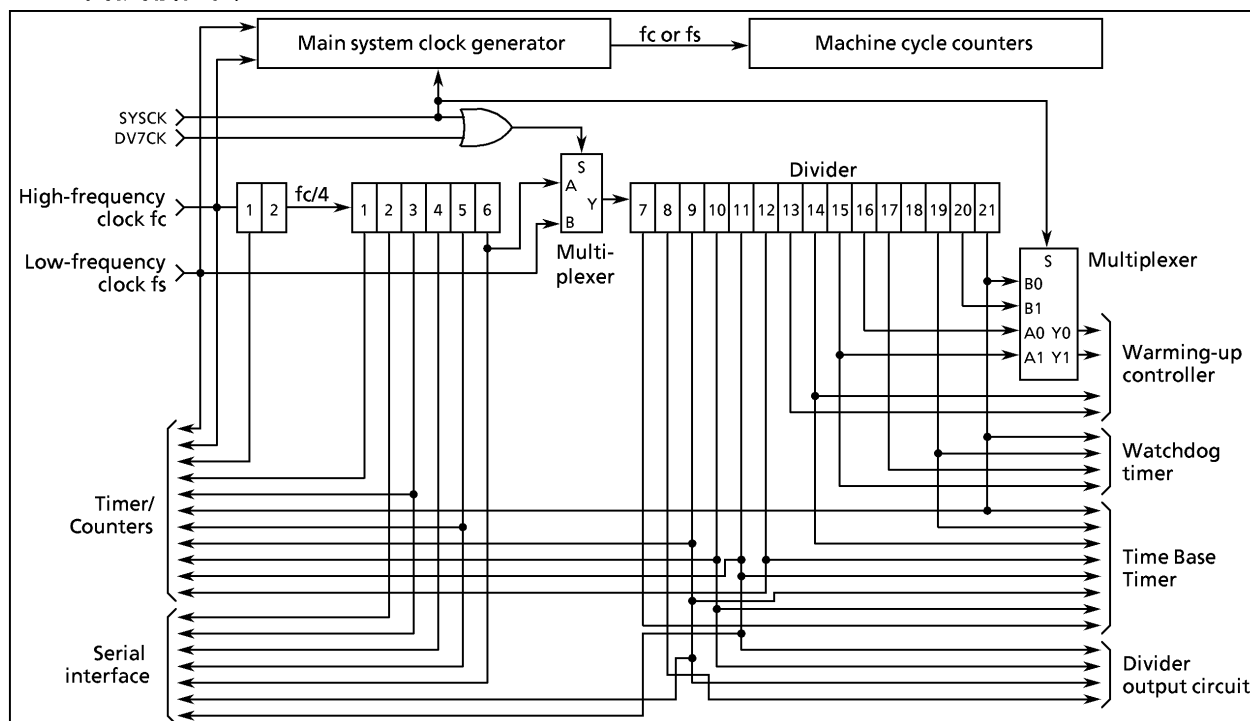


Figure 1-4. Configuration of Timing Generator

TBTCR (0036H)	7	6	5	4	3	2	1	0	(Initial value: 0**0 0***)
	(DVOEN)	(DVQCK)	DV7CK	(TBTEN)		(TBTCK)			
	DV7CK	Selection of input to the 7th stage of the divider					0: $fc/2^8$ [Hz] 1: fs		R/W

Note 1: In Single Clock mode, do not set DV7CK to "1".
 Note 2: Do not set "1" on DV7CK while the low-frequency clock is not operated stably.
 Note 3: fc: High-frequency clock [Hz], fs: Low-frequency clock [Hz], *: Don't care
 Note 4: In SLOW1/2 and SLEEP1/2 modes, the DV7CK setting is ineffective, and fs is input to the 7th stage of the divider.
 Note 5: When STOP mode is entered from NORMAL 1/2 mode, the DV7CK setting is ineffective during the warm-up period after release of STOP mode, and the 6th stage of the divider is input to the 7th stage during this period.

Figure 1-5. Timing Generator Control Register

(2) Machine cycle

Instruction execution and peripheral hardware operation are synchronized with the main system clock. The minimum instruction execution unit is called a “machine cycle”. There are a total of 10 different types of instructions for the TLCS-870/C Series: ranging from 1-cycle instructions which require one machine cycle for execution to 10-cycle instructions which require 10 machine cycles for execution. A machine cycle consists of 4 states (S0 to S3), and each state consists of one main system clock.

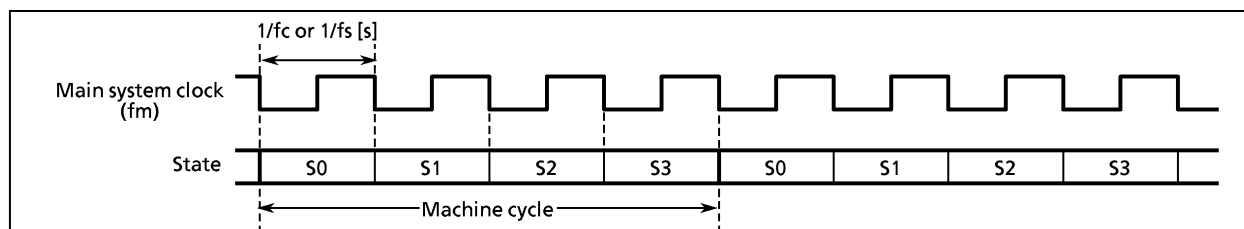


Figure 1-6. Machine Cycle

1.4.3 Operation Mode Control Circuit

The operation mode control circuit starts and stops the oscillation circuits for the high-frequency and low-frequency clocks, and switches the main system clock. There are two operating modes: single-clock and dual-clock. These modes are controlled by the system control registers (SYSCR1 and SYSCR2). Figure 1-7 shows the operating mode transition diagram and Figure 1-8 shows the system control registers.

(1) Single-clock mode

Only the oscillation circuit for the high-frequency clock is used, and P21 (XTIN) and P22 (XTOUT) pins are used as input/output ports. The main-system clock is obtained from the high-frequency clock. In the single-clock mode, the machine cycle time is $4/f_c$ [s].

① NORMAL1 mode

In this mode, both the CPU core and on-chip peripherals operate using the high-frequency clock. The TMP86CM25/S25 are placed in this mode after reset.

② IDLE1 mode

In this mode, the internal oscillation circuit remains active. The CPU and the watchdog timer are halted; however on-chip peripherals remain active (operate using the high-frequency clock). IDLE1 mode is started by SYSCR2<IDLE>, and IDLE1 mode is released to NORMAL1 mode by an interrupt request from the on-chip peripherals or external interrupt inputs. When the IMF (interrupt master enable flag) is “1” (interrupt enable), the execution will resume with the acceptance of the interrupt, and the operation will return to normal after the interrupt service is completed. When the IMF is “0” (interrupt disable), the execution will resume with the instruction which follows the IDLE1 mode start instruction.

③ IDLE0 mode

In this mode, all the circuit, except oscillator and the Timer-Base-Timer, stops operation.

This mode is enabled by setting “1” on bit TGHALT on the system control register 2 (SYSCR2).

When IDLE0 mode starts, the CPU stops and the timing generator stops feeding the clock to the peripheral circuits other than TBT. Then, upon detecting the falling edge of the source clock selected with TBTCCR < TBTCK >, the timing generator starts feeding the clock to all peripheral circuits.

When returned from IDLE0 mode, the CPU restarts operating, entering NORMAL1 mode back again. IDLE0 mode is entered and returned regardless of how TBTCCR < TBTEN > is set. When IMF = “1”, EF₆ (TBT interrupt individual enable flag) = “1”, and TBTCCR < TBTEN > = “1”, interrupt processing is performed. When IDLE0 mode is entered while TBTCCR < TBTEN > = “1”, the INTTBT interrupt latch is set after returning to NORMAL1 mode.

(2) Dual-clock mode

Both the high-frequency and low-frequency oscillation circuits are used in this mode. P21 (XTIN) and P22 (XTOUT) pins cannot be used as input/output ports. The main system clock is obtained from the high-frequency clock in NORMAL2 and IDLE2 modes, and is obtained from the low-frequency clock in SLOW and SLEEP modes. The machine cycle time is $4/f_c$ [s] in the NORMAL2 and IDLE2 modes, and $4/f_s$ [s] ($122 \mu\text{s}$ at $f_s = 32.768 \text{ kHz}$) in the SLOW and SLEEP modes.

The TLCS-870/C is placed in the signal-clock mode during reset. To use the dual-clock mode, the low-frequency oscillator should be turned on at the start of a program.

① NORMAL2 mode

In this mode, the CPU core operates with the high-frequency clock. On-chip peripherals operate using the high-frequency clock and/or low-frequency clock.

② SLOW2 mode

In this mode, the CPU core operates with the low-frequency clock, while both the high-frequency clock and the low-frequency clock are operated. On-chip peripherals are triggered by the low-frequency clock. As the SYSCK on SYSCR2 becomes “0”, the hardware changes into NORMAL2 mode. As the XEN on SYSCR2 becomes “0”, the hardware changes into SLOW1 mode. Do not clear XTEN to “0” during SLOW2 mode.

③ SLOW1 mode

This mode can be used to reduce power-consumption by turning off oscillation of the high-frequency clock. The CPU core and on-chip peripherals operate using the low-frequency clock.

Switching back and forth between SLOW1 and SLOW2 modes are performed by XEN bit on the system control register 2 (SYSCR2). In SLOW1 and SLEEP mode, the input clock to the 1st stage of the divider is stopped; output from the 1st to 6th stages is also stopped.

④ IDLE2 mode

In this mode, the internal oscillation circuit remain active. The CPU and the watchdog timer are halted ; however, on-chip peripherals remain active (operate using the high-frequency clock and/or the low-frequency clock). Starting and releasing of IDLE2 mode are the same as for IDLE1 mode, except that operation returns to NORMAL2 mode.

⑤ SLEEP1 mode

In this mode, the internal oscillation circuit of the low-frequency clock remains active. The CPU, the watchdog timer, and the internal oscillation circuit of the high-frequency clock are halted; however, on-chip peripherals remain active (operate using the low-frequency clock). Starting and releasing of SLEEP mode are the same as for IDLE1 mode, except that operation returns to SLOW mode. In SLOW and SLEEP mode, the input clock to the 1st stage of the divider is stopped; output from the 1st to 6th stages is also stopped.

⑥ SLEEP2 mode

The SLEEP2 mode is the idle mode corresponding to the SLOW2 mode. The status under the SLEEP2 mode is same as that under the SLEEP1 mode, except for the oscillation circuit of the high-frequency clock.

⑦ SLEEP0 mode

In this mode, all the circuit, except oscillator and the Timer-Base-Timer, stops operation.

This mode is enabled by setting "1" on bit TGHALT on the system control register 2 (SYSCR2).

When SLEEP0 mode starts, the CPU stops and the timing generator stops feeding the clock to the peripheral circuits other than TBT. Then, upon detecting the falling edge of the source clock selected with TBTCCR<TBTCK>, the timing generator starts feeding the clock to all peripheral circuits.

When returned from SLEEP0 mode, the CPU restarts operating, entering SLOW1 mode back again. SLEEP0 mode is entered and returned regardless of how TBTCCR<TBTEN> is set. When IMF = "1", EF₆ (TBT interrupt individual enable flag) = "1", and TBTCCR<TBTEN> = "1", interrupt processing is performed. When SLEEP0 mode is entered while TBTCCR<TBTEN> = "1", the INTTBT interrupt latch is set after returning to SLOW1 mode.

(3) STOP mode

In this mode, the internal oscillation circuit is turned off, causing all system operations to be halted. The internal status immediately prior to the halt is held with a lowest power consumption during STOP mode.

STOP mode is started by the system control register 1 (SYSCR1), and STOP mode is released by a inputting (either level-sensitive or edge-sensitive can be programmably selected) to the $\overline{\text{STOP}}$ pin. After the warming-up period is completed, the execution resumes with the instruction which follows the STOP mode start instruction.

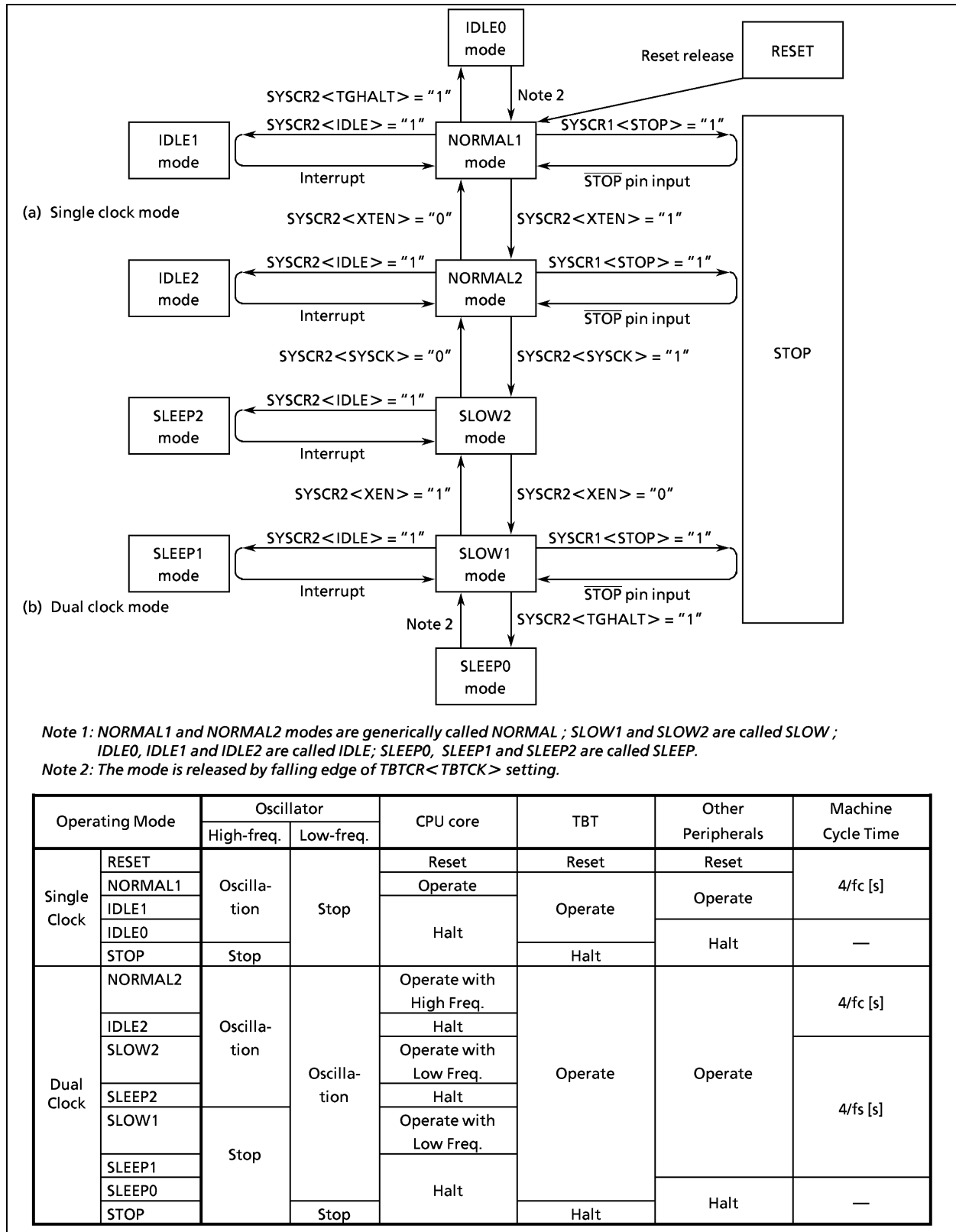


Figure 1-7. Operating Mode Transition Diagram

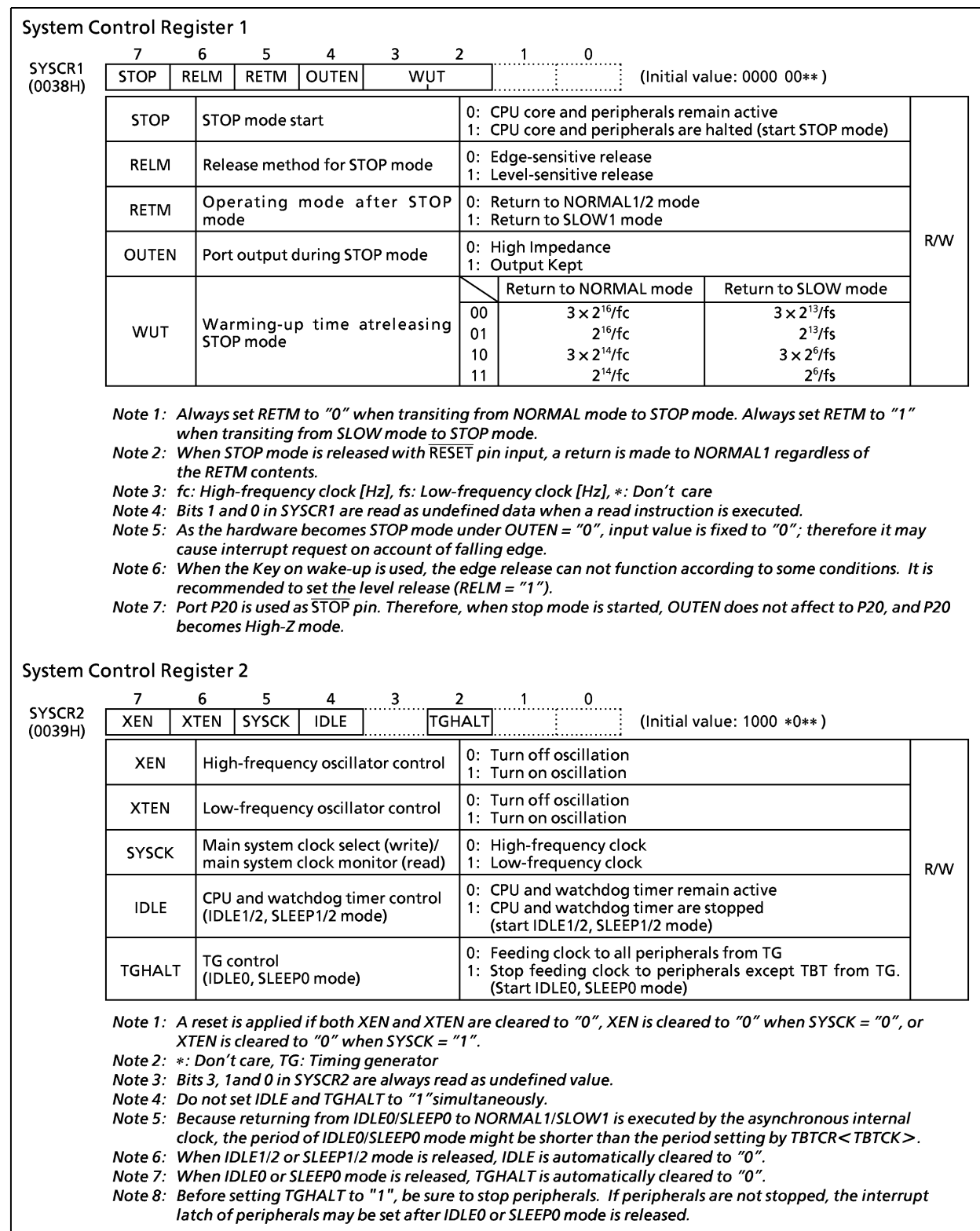


Figure 1-8. System Control Registers

1.4.4 Operating Mode Control

(1) STOP mode

STOP mode is controlled by the system control register 1, the $\overline{\text{STOP}}$ pin input and key wake-up input (STOP2 to STOP5) which is controlled by the STOP mode release control register (STOPCR).

The $\overline{\text{STOP}}$ pin is also used both as a port P20 and an $\overline{\text{INT5}}$ (external interrupt input 5) pin.

STOP mode is started by setting STOP (bit 7 in SYSCR1) to "1". During STOP mode, the following status is maintained.

- ① Oscillations are turned off, and all internal operations are halted.
- ② The data memory, registers, the program status word and port output latches are all held in the status in effect before STOP mode was entered.
- ③ The prescaler and the divider of the timing generator are cleared to "0".
- ④ The program counter holds the address 2 ahead of the instruction (e.g. [SET (SYSCR1).7]) which started STOP mode.

STOP mode includes a level-sensitive mode and an edge-sensitive mode, either of which can be selected with the RELM (bit 6 in SYSCR1). Do not use any STOPx (x: 2 to 5) pin input for releasing STOP mode in edge-sensitive mode.

Note 1: $\overline{\text{STOP}}$ pin doesn't have the control register such as STOPCR, so when STOP mode is released by STOPx (x: 2 to 5), $\overline{\text{STOP}}$ pin should be used as STOP function.

Note 2: During STOP period (from start of STOP mode to end of warming-up), due to changes in the external interrupt pin signal, interrupt latches may be set to "1" and interrupts may be accepted immediately after STOP mode is released. Before starting STOP mode, therefore, disable interrupts. Also, before enabling interrupts after STOP mode is released, clear unnecessary interrupt latches.

a. Level-sensitive release mode (RELM = "1")

In this mode, STOP mode is released by setting the $\overline{\text{STOP}}$ pin high or setting the STOPx (x: 2 to 5) pin input which is enabled by STOPCR. This mode is used for capacitor back-up when the main power supply is cut off and long term battery back-up.

When the $\overline{\text{STOP}}$ pin input is high or STOPx (x: 2 to 5) pin input which is enabled by STOPCR is low, executing an instruction which starts STOP mode will not place in STOP mode but instead will immediately start the release sequence (warm-up). Thus, to start STOP mode in the level-sensitive release mode, it is necessary for the program to first confirm that the $\overline{\text{STOP}}$ pin input is low and STOPx (x: 2 to 5) pin input which is enabled by STOPCR is high. The following two methods can be used for confirmation.

- ① Testing a port P20.
- ② Using an external interrupt input $\overline{\text{INT5}}$ ($\overline{\text{INT5}}$ is a falling edge-sensitive input).

Example 1: Starting STOP mode from NORMAL mode by testing a port P20.

```
LD    (SYSCR1), 01010000B ; Sets up the level-sensitive release mode
SSTOPH: TEST (P2PRD). 0    ; Wait until the  $\overline{\text{STOP}}$  pin input goes low level
JRS  F, SSTOPH
SET  (SYSCR1). 7          ; Starts STOP mode
```

Example 2: Starting STOP mode from NORMAL mode with an INT5 interrupt.

```

PINT5: TEST (P2PRD). 0      ; To reject noise, STOP mode does not start if
      JRS  F, SINT5         port P20 is at high
      LD  (SYSCR1), 01010000B ; Sets up the level-sensitive release mode.
      SET (SYSCR1). 7      ; Starts STOP mode
SINT5:  RETI
    
```

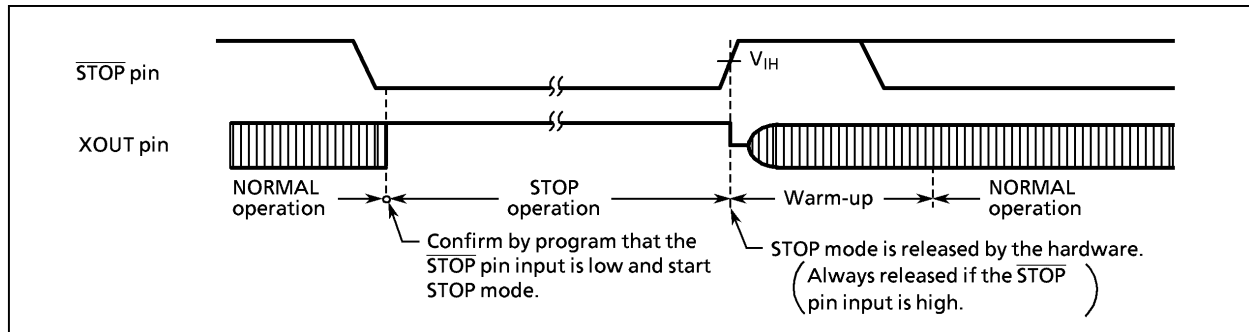


Figure 1-9. Level-sensitive Release Mode

Note 1: Even if the $\overline{\text{STOP}}$ pin input is low or STOP_x ($x: 2$ to 5) pin input which is enabled by STOPCR is high after warming up start, the STOP mode is not restarted.

Note 2: In this case of changing to the level-sensitive mode from the edge-sensitive mode, the release mode is not switched until a rising edge of the $\overline{\text{STOP}}$ pin input is detected.

b. Edge-sensitive release mode (RELM="0")

In this mode, STOP mode is released by a rising edge of the $\overline{\text{STOP}}$ pin input. This is used in applications where a relatively short program is executed repeatedly at periodic intervals. This periodic signal (for example, a clock from a low-power consumption oscillator) is input to the $\overline{\text{STOP}}$ pin. In the edge-sensitive release mode, $\overline{\text{STOP}}$ mode is started even when the $\overline{\text{STOP}}$ pin input is high level. Do not use any STOP_x ($x: 2$ to 5) pin input for releasing STOP mode in edge-sensitive release mode.

Example: Starting STOP mode from NORMAL mode

```
LD (SYSCR1), 10010000B ; Starts after specified to the edge-sensitive release mode
```

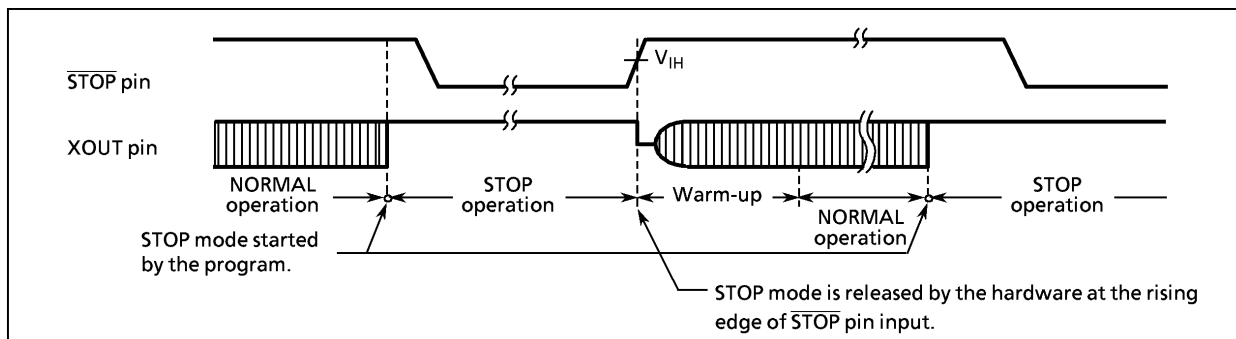


Figure 1-10. Edge-sensitive Release Mode

STOP mode is released by the following sequence.

- ① In the dual-clock mode, when returning to NORMAL2 or SLOW2, both the high-frequency and low-frequency clock oscillators are turned on ; when returning to SLOW1 mode, only the low-frequency clock oscillator is turned on. In the signal-clock mode, only the high-frequency clock oscillator is turned on.
- ② A warm-up period is inserted to allow oscillation time to stabilize. During warm-up, all internal operations remain halted. Four different warm-up times can be selected with the WUT (bits 2 and 3 in SYSCR1) in accordance with the resonator characteristics.
- ③ When the warm-up time has elapsed, normal operation resumes with the instruction following the STOP mode start instruction. The start is made after the prescaler and the divider of the timing generator are cleared to "0".

Table 1-1. Warm-up Time Example (at $f_c = 16.0$ MHz, $f_s = 32.768$ kHz)

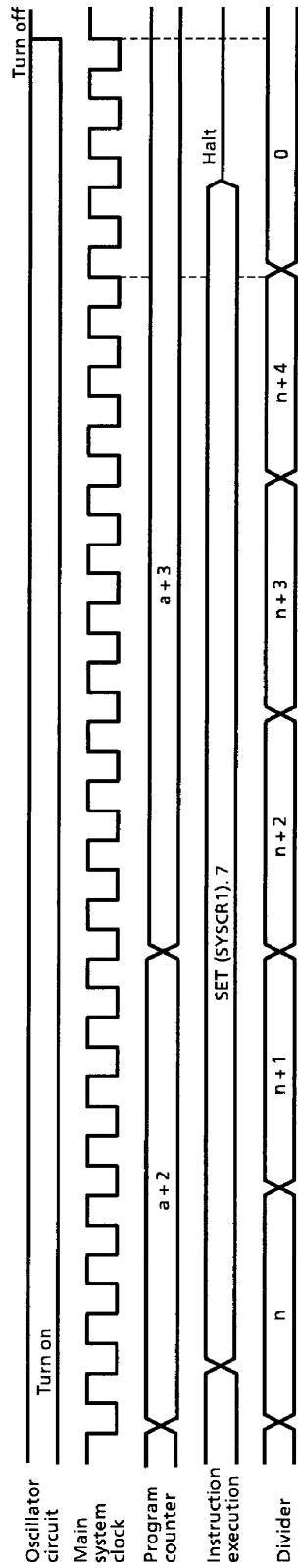
WUT	Warm-up Time [ms]	
	Return to NORMAL mode	Return to SLOW mode
00	12.288	750
01	4.096	250
10	3.072	5.85
11	1.024	1.95

Note: The warm-up time is obtained by dividing the basic clock by the divider; therefore, the warm-up time may include a certain amount of error if there is any fluctuation of the oscillation frequency when STOP mode is released. Thus, the warm-up time must be considered an approximate value.

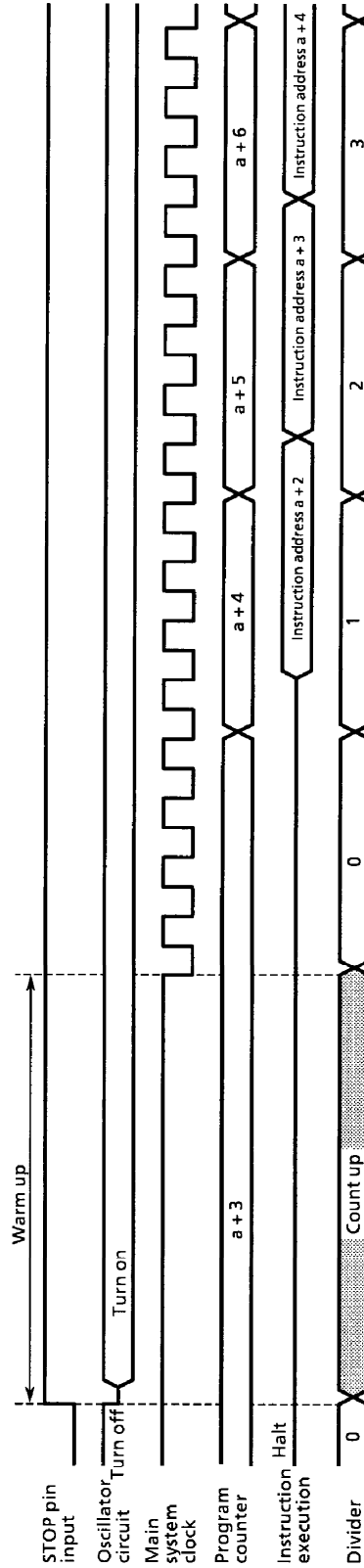
STOP mode can also be released by inputting low level on the $\overline{\text{RESET}}$ pin, which immediately performs the normal reset operation.

Note: When STOP mode is released with a low hold voltage, the following cautions must be observed.

The power supply voltage must be at the operating voltage level before releasing STOP mode. The $\overline{\text{RESET}}$ pin input must also be "H" level, rising together with the power supply voltage. In this case, if an external time constant circuit has been connected, the $\overline{\text{RESET}}$ pin input voltage will increase at a slower pace than the power supply voltage. At this time, there is a danger that a reset may occur if input voltage level of the $\overline{\text{RESET}}$ pin drops below the non-inverting high-level input voltage (hysteresis input).



(a) STOP mode start (Example: Start with SET (SYSCR1).7 instruction located at address a)



(b) STOP mode release

Figure 1-11. STOP Mode Start/Release

- Start the IDLE1/2 and SLEEP1/2 modes

When IDLE1/2 and SLEEP1/2 modes start, set SYSCR2 <IDLE> to “1”.

- Release the IDLE1/2 and SLEEP1/2 modes

IDLE1/2 and SLEEP1/2 modes include a normal release mode and an interrupt release mode. These modes are selected by interrupt master enable flag (IMF).

After releasing IDLE1/2 and SLEEP1/2 modes, the SYSCR2 <IDLE> is automatically cleared to “0” and the operation mode is returned to the mode preceding IDLE1/2 and SLEEP1/2 modes.

IDLE1/2 and SLEEP1/2 modes can also be released by inputting low level on the **RESET** pin. After releasing reset, the operation mode is started from NORMAL1 mode.

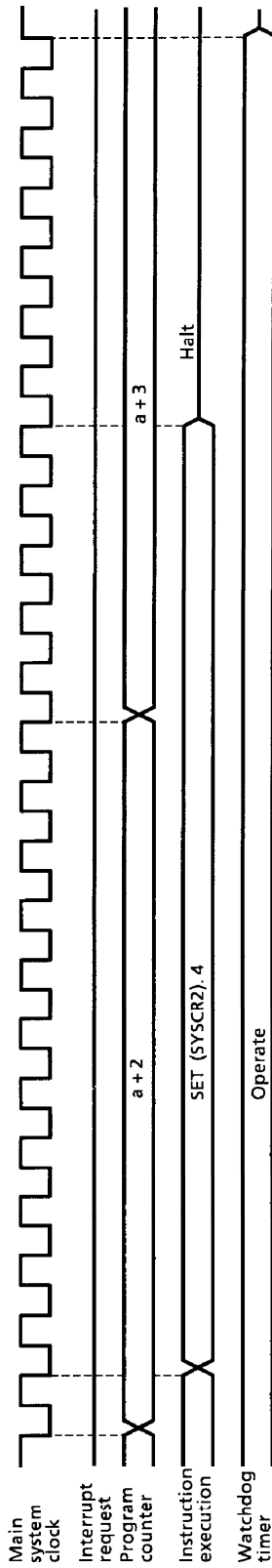
a. Normal release mode (IMF = “0”)

IDLE1/2 and SLEEP1/2 modes are released by any interrupt source enabled by the individual interrupt enable flag (EF). After the interrupt is generated, the program operation is resumed from the instruction following the IDLE1/2 and SLEEP1/2 modes start instruction. Normally, the interrupt latches (IL) of the interrupt source used for releasing must be cleared to “0” by load instructions.

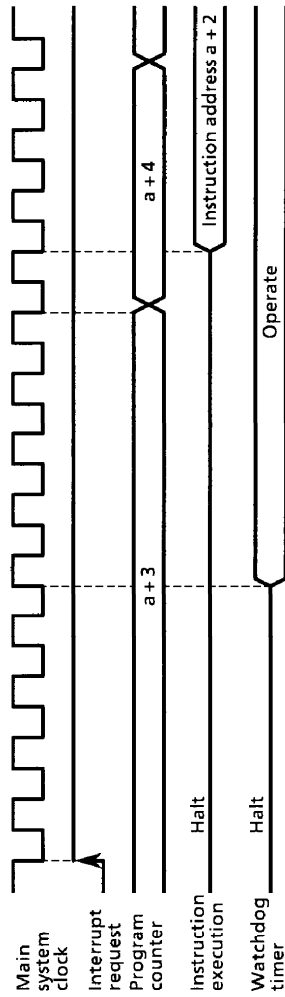
b. Interrupt release mode (IMF = “1”)

IDLE1/2 and SLEEP1/2 modes are released by any interrupt source enabled with the individual interrupt enable flag (EF) and the interrupt processing is started. After the interrupt is processed, the program operation is resumed from the instruction following the instruction, which starts IDLE1/2 and SLEEP1/2 modes.

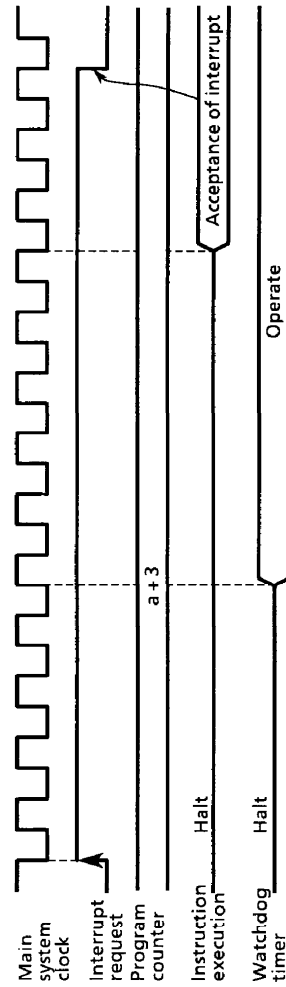
Note: When a watchdog timer interrupt is generated immediately before IDLE1/2 and SLEEP1/2 mode are started, the watchdog timer interrupt will be processed but IDLE1/2 and SLEEP1/2 mode will not be started.



(a) IDLE mode start (Example: starting with the SET instruction located at address a)



① Normal release mode



② Interrupt release mode

(b) IDLE mode release

Figure 1-13. IDLE Mode Start/Release

(3) IDLE0, SLEEP0 mode (IDLE0,SLEEP0)

IDLE0 and SLEEP0 modes are controlled by the system control register 2 (SYSCR2) and the time base timer control register (TBTTCR). The following status is maintained during IDLE0 and SLEEP0 modes.

- ① Timing generator stops feeding clock to peripherals except TBT.
- ② The data memory, CPU registers, program status word and port output latches are all held in the status in effect before IDLE0 and SLEEP0 modes were entered.
- ③ The program counter holds the address 2 ahead of the instruction which starts IDLE0 and SLEEP0 modes.

Note: Before starting IDLE0 or SLEEP0 mode, be sure to stop (disable) peripherals.

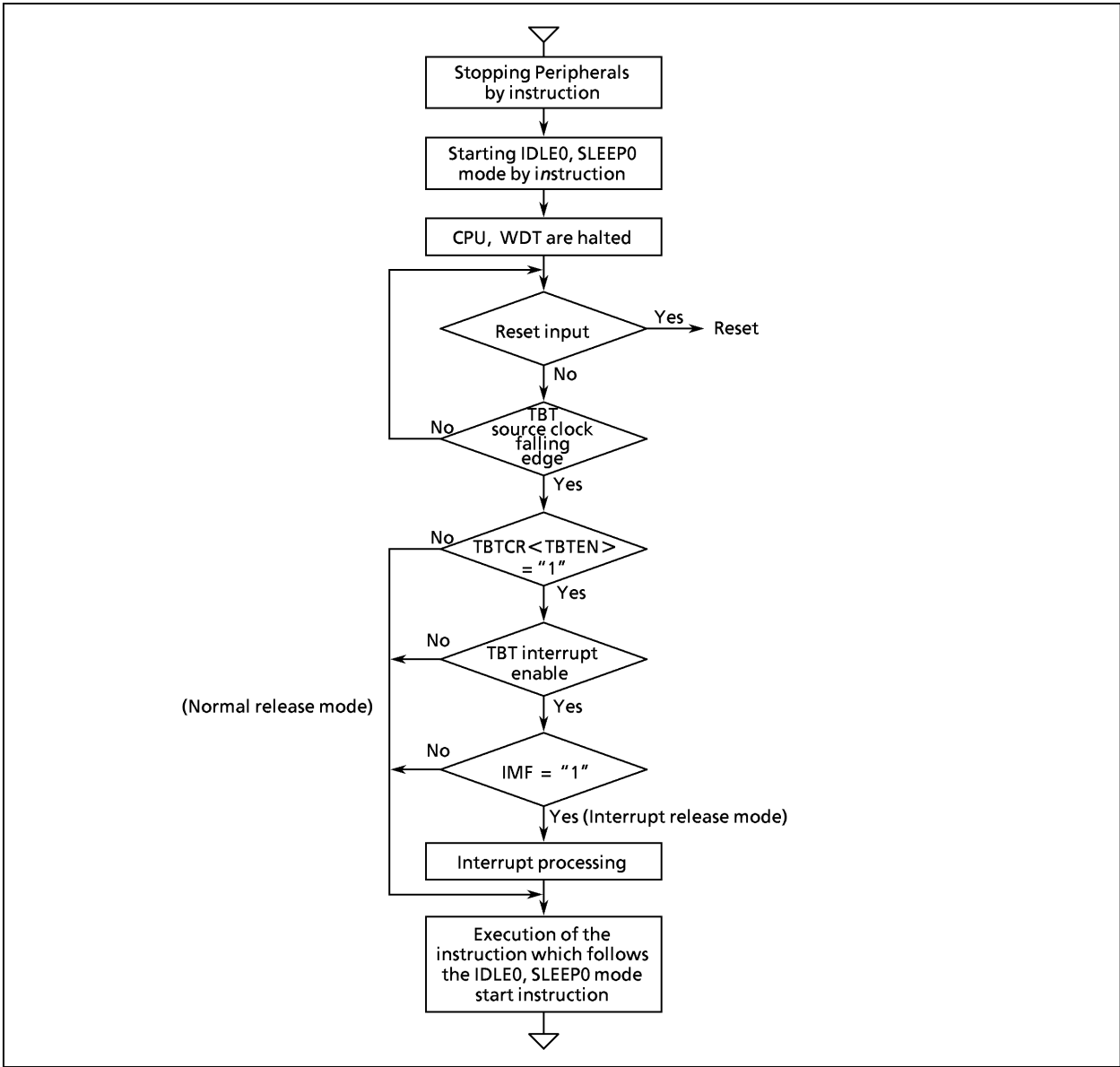


Figure 1-14. IDLE0, SLEEP0 mode

- Start the IDLE0 and SLEEP0 modes

Stop (disable) peripherals such as a timer counter. When IDLE0 and SLEEP0 modes start, set SYSCR2<TGHALT> to “1”.

- Release the IDLE0 and SLEEP modes

IDLE0 and SLEEP0 modes include a normal release mode and an interrupt release mode.

These modes are selected by interrupt master flag (IMF), individual interrupt enable-flag (EF6) for INTTBT and TBTCR<TBTEN>.

After releasing IDLE0 and SLEEP0 modes, the SYSCR2<TGHALT> is automatically cleared to “0” and the operation mode is returned to the mode preceding IDLE0 and SLEEP0 modes. Before starting the IDLE0 or SLEEP0 mode, when the TBTCR<TBTEN> is set to “1”, INTTBT interrupt latch is set to “1”.

IDLE0 and SLEEP0 modes can also be released by inputting low level on the $\overline{\text{RESET}}$ pin. After releasing reset, the operation mode is started from NORMAL1 mode.

Note: IDLE0 and SLEEP0 modes start/release without reference to TBTCR<TBTEN> setting.

- a. Normal release mode (IMF · EF6 · TBTCR<TBTEN> = “0”)

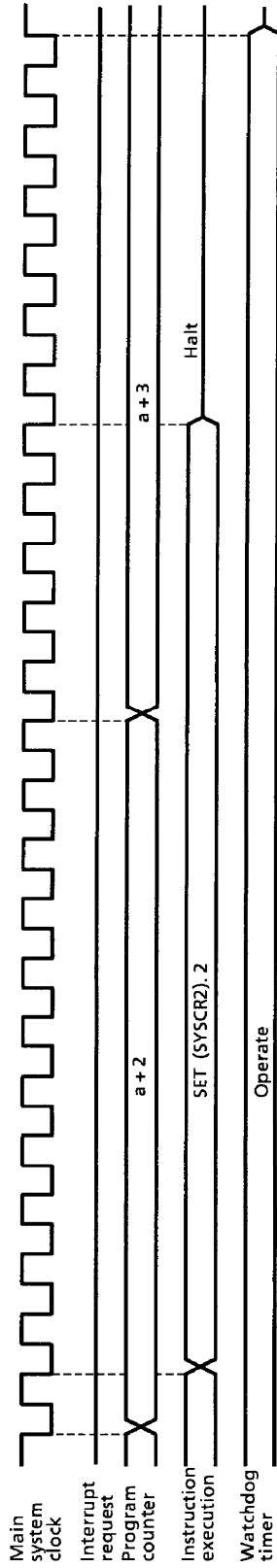
IDLE0 and SLEEP0 modes are released by the source clock falling edge, which is setting by the TBTCR<TBTK>. After the falling edge is detected, the program operation is resumed from the instruction following the IDLE0 and SLEEP0 modes start instruction.

- b. Interrupt release mode (IMF · EF6 · TBTCR<TBTEN> = “1”)

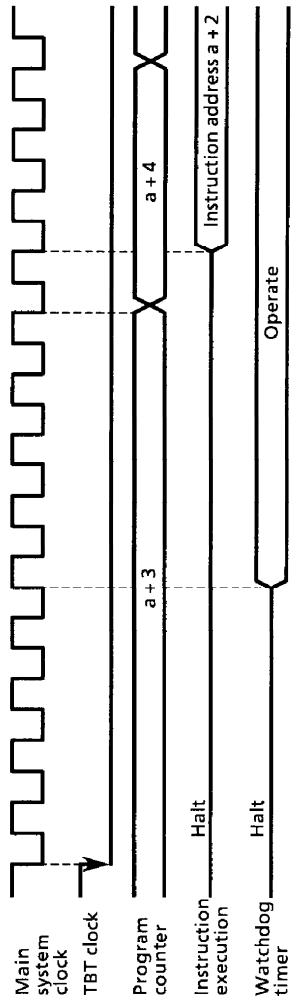
IDLE0 and SLEEP0 modes are released by the source clock falling edge, which is setting by the TBTCR<TBTK> and INTTBT interrupt processing is started.

Note 1: Because returning from IDLE0, SLEEP0 to NORMAL1, SLOW1 is executed by the asynchronous internal clock, the period of IDLE0, SLEEP0 mode might be the shorter than the period setting by TBTCR<TBTK>.

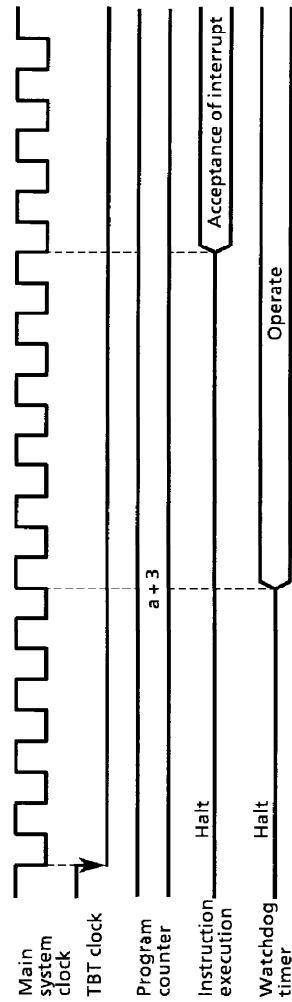
Note 2: When a watchdog timer interrupt is generated immediately before IDLE0 / SLEEP0 mode is started, the watchdog timer interrupt will be processed but IDLE0 / SLEEP0 mode will not be started.



(a) IDLE0, SLEEP0 mode start (Example: starting with the SET instruction located at address a)



① Normal release mode



② Interrupt release mode

(b) IDLE0, SLEEP0 mode release

Figure 1-15. IDLE0, SLEEP0 Mode Start/Release

(4) SLOW mode

SLOW mode is controlled by the system control register 2 (SYSCR2).

The following is the methods to switch the mode with the warming-up counter (TC4, 3).

a. Switching from NORMAL2 mode to SLOW1 mode

First, set SYSCK (bit 5 in SYSCR2) to switch the main system clock to the low-frequency clock for SLOW2 mode.

Next, clear XEN (bit 7 in SYSCR2) to turn off high-frequency oscillation.

Note: The high-frequency clock oscillation can be continued to return quickly to NORMAL2 mode.

When the low-frequency clock oscillation is unstable, wait until oscillation stabilizes before performing the above operations. The timer/counter 4, 3 (TC4, TC3) can conveniently be used to confirm that low-frequency clock oscillation has stabilized.

Example1: Switching from NORMAL2 mode to SLOW1 mode.

```

SET (SYSCR2). 5      ; SYSCR2<SYSCK> ← 1
                    ; (switches the main system clock to the low-frequency
                    ; clock for SLOW2)
CLR (SYSCR2). 7      ; SYSCR2<XEN> ← 0
                    ; (turns off high-frequency oscillation)

```

Example2: Switching to the SLOW1 mode after low-frequency clock has stabilized.

```

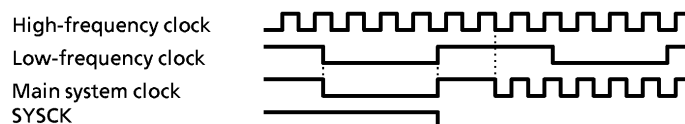
LD (TC3CR), 43H      ; Sets mode for TC4, TC3 (16-bit TC, fs for source)
LD (TC4CR), 05H
LDW (TTREG3), 8000H ; Sets warming-up time
                    ; (depend on oscillator accompanied)
DI                   ; IMF ← 0
SET (EIRH). 3        ; Enables INTTC4
EI                   ; IMF ← 1
SET (TC4CR). 3       ; Starts TC4, 3
:
PINTTC4: CLR (TC4CR). 3 ; Stops TC4, 3
          SET (SYSCR2). 5 ; SYSCR2<SYSCK> ← 1
                    ; (Switches the main system clock to the
                    ; low-frequency clock)
          CLR (SYSCR2). 7 ; SYSCR2<XEN> ← 0
                    ; (Turns off high-frequency oscillation)
          RETI
:
VINTTC4: DW PINTTC4   ; INTTC4 vector table

```

b. Switching from SLOW1 mode to NORMAL2 mode

First, set XEN (bit 7 in SYSCR2) to turn on the high-frequency oscillation. When time for stabilization (warm-up) has been taken by the timer/counter 4, 3 (TC4, 3), clear SYSCK (bit 5 in SYSCR2) to switch the main system clock to the high-frequency clock.

Note 1: After SYSCK is cleared to "0", executing the instructions is continued by the low-frequency clock for the period synchronized with low-frequency and high-frequency clocks.



Note 2: SLOW mode can also be released by inputting low level on the RESET pin, which immediately performs the reset operation. After reset, the TMP86CM25/S25 are placed in NORMAL1 mode.

Example: Switching from the SLOW1 mode to the NORMAL2 mode

($f_c = 16$ MHz, warm-up time is 4.0 ms).

```

SET  (SYSCR2). 7      ; SYSCR2<XEN> ← 1 (Starts high-frequency oscillation)
LD   (TC3CR), 63H    ; Sets mode for TC4, TC3 (16-bit TC, fc for source)
LD   (TC4CR), 05H
LD   (TTREG4), 0F8H  ; Sets warming-up time
DI                                       ; IMF ← 0
SET  (EIRH). 3      ; Enables INTTC4
EI                                       ; IMF ← 1
SET  (TC4CR). 3     ; Starts TC4, 3
    ⋮
PINTTC4: CLR (TC4CR). 3 ; Stops TC4, 3
        CLR (SYSCR2). 5 ; SYSCR2<SYSCK> ← 0
                          ; (Switches the main system clock to the
                          ; high-frequency clock)

        RETI
    ⋮
VINTTC4: DW  PINTTC4    ; INTTC4 vector table

```

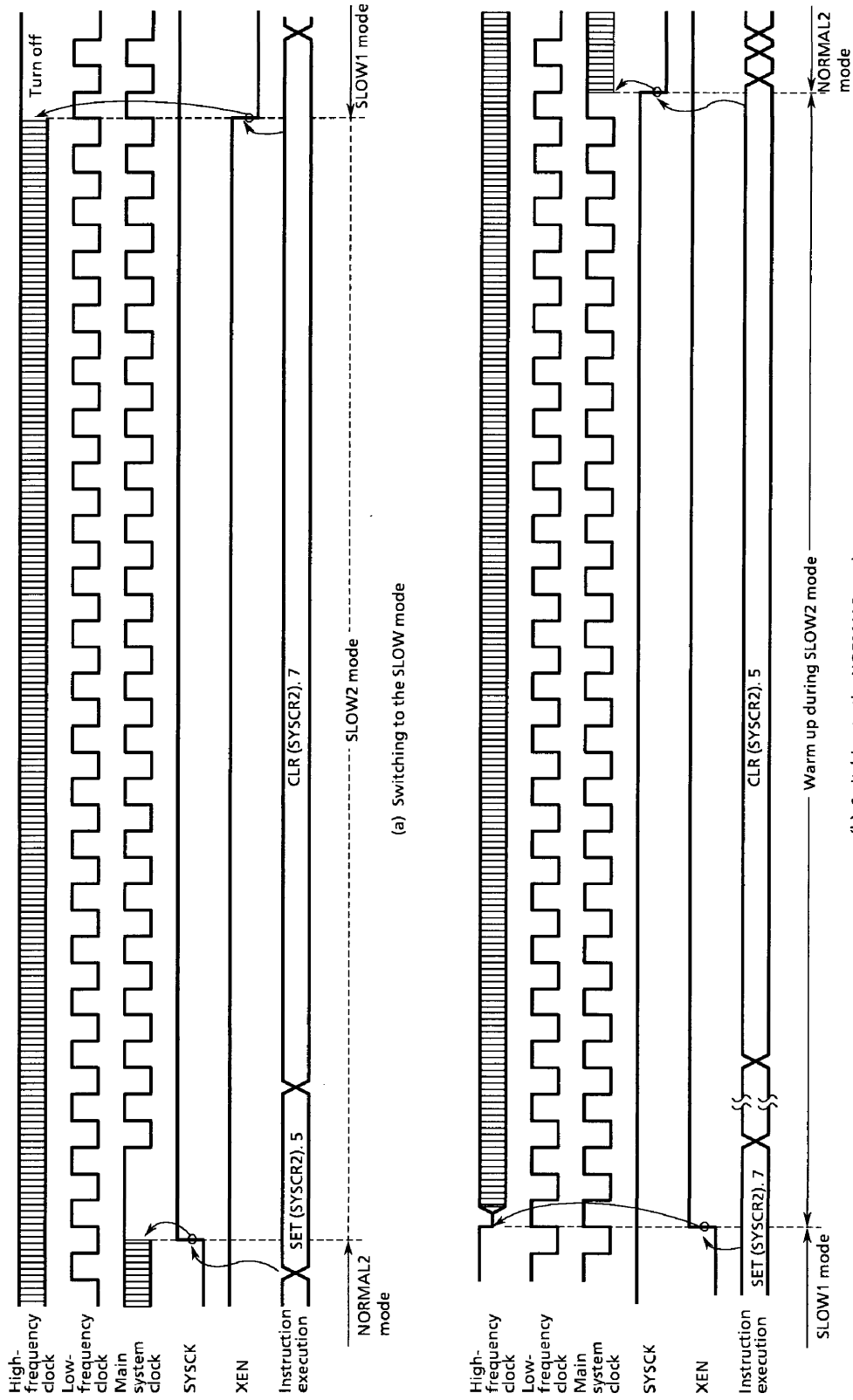


Figure 1-16. Switching between the NORMAL2 and SLOW Modes

1.5 Interrupt Control Circuit

The TMP86CM25/S25 are a total (Reset is excluded) of 15 interrupt sources for 20 interrupt factors; 4 of the sources are multiplexed. Multiple interrupt with priorities is available. 4 of the internal factors are non-maskable interrupts, and the rest of them are maskable interrupts.

Interrupt sources are provided with interrupt latches (IL), which hold interrupt requests, and independent vectors. The interrupt latch is set to “1” by the generation of its interrupt request which requests the CPU to accept its interrupts. Interrupts are enabled or disabled by software using the interrupt master enable flag (IMF) and interrupt enable flag (EF). If more than one interrupts are generated simultaneously, interrupts are accepted in order which is dominated by hardware. However, there are no prioritized interrupt factors among non-maskable interrupts.

Table 1-2. Interrupt Sources

Interrupt Factors		Enable Condition	Interrupt Latch	Vector Address	Priority
Internal/ External	(Reset)	Non-maskable	—	FFFEH	High 1
Internal	INTSWI (Software interrupt)	Non-maskable	—	FFFCH	2
Internal	INTUNDEF (Executed the Undefined Instruction interrupt)	Non-maskable	—	FFFCH	2
Internal	INTATRAP (Address Trap interrupt)	Non-maskable	IL2	FFFAH	2
Internal	INTWDT (Watchdog Timer interrupt)	Non-maskable	IL3	FFF8H	2
External	$\overline{\text{INT0}}$ (External interrupt 0)	IMF = 1, EF4 = 1	IL4	FFF6H	5
External	INT1 (External interrupt 1)	IMF = 1, EF5 = 1	IL5	FFF4H	6
Internal	INTTBT (Time Base Timer interrupt)	IMF = 1, EF6 = 1	IL6	FFF2H	7
External	INT2 (External interrupt2)	IMF = 1, EF7 = 1	IL7	FFF0H	8
Internal	INTTC1 (18-bit TC1 interrupt)	IMF = 1, EF8 = 1	IL8	FFEEH	9
Internal	INTRxD (UART received interrupt)	IMF = 1, EF9 = 1	IL9	FFECH	10
Internal	INTSIO0 (SIO0 interrupt)				
Internal	INTTxD (UART transmitted interrupt)	IMF = 1, EF10 = 1	IL10	FFEAH	11
Internal	INTSIO1 (SIO1 interrupt)				
Internal	INTTC4 (TC4 interrupt)	IMF = 1, EF11 = 1	IL11	FFE8H	12
Internal	INTTC6 (TC6 interrupt)	IMF = 1, EF12 = 1	IL12	FFE6H	13
Internal	INTADC (AD converter interrupt)	IMF = 1, EF13 = 1	IL13	FFE4H	14
External	INT3 (External interrupt 3)	IMF = 1, EF14 = 1	IL14	FFE2H	15
Internal	INTTC3 (TC3 interrupt)				
External	$\overline{\text{INT5}}$ (External interrupt 5)	IMF = 1, EF15 = 1	IL15	FFE0H	Low 16
Internal	INTTC5 (TC5 interrupt)				

Note 1: The following interrupt factors share their interrupt source; the factor is selected on the register INTSEL.

- 1) INTRxD and INTSIO0 share the source whose priority is 10.
- 2) INTTxD and INTSIO1 share the source whose priority is 11.
- 3) INT3 and INTTC3 share the source whose priority is 15.
- 4) $\overline{\text{INT5}}$ and INTTC5 share the source whose priority is 16.

Note 2: To use the address trap interrupt (INTATRAP), clear WDTTCR1<ATOUT> to “0” (it is set for the “reset request” after reset is cancelled). For details, see 2.4.5 Address Trap.

Note 3: To use the watchdog timer interrupt (INTWDT), clear WDTTCR1<WDTOUT> to “0” (it is set for the “reset request” after reset is cancelled). For details, see 2.4 Watchdog Timer.

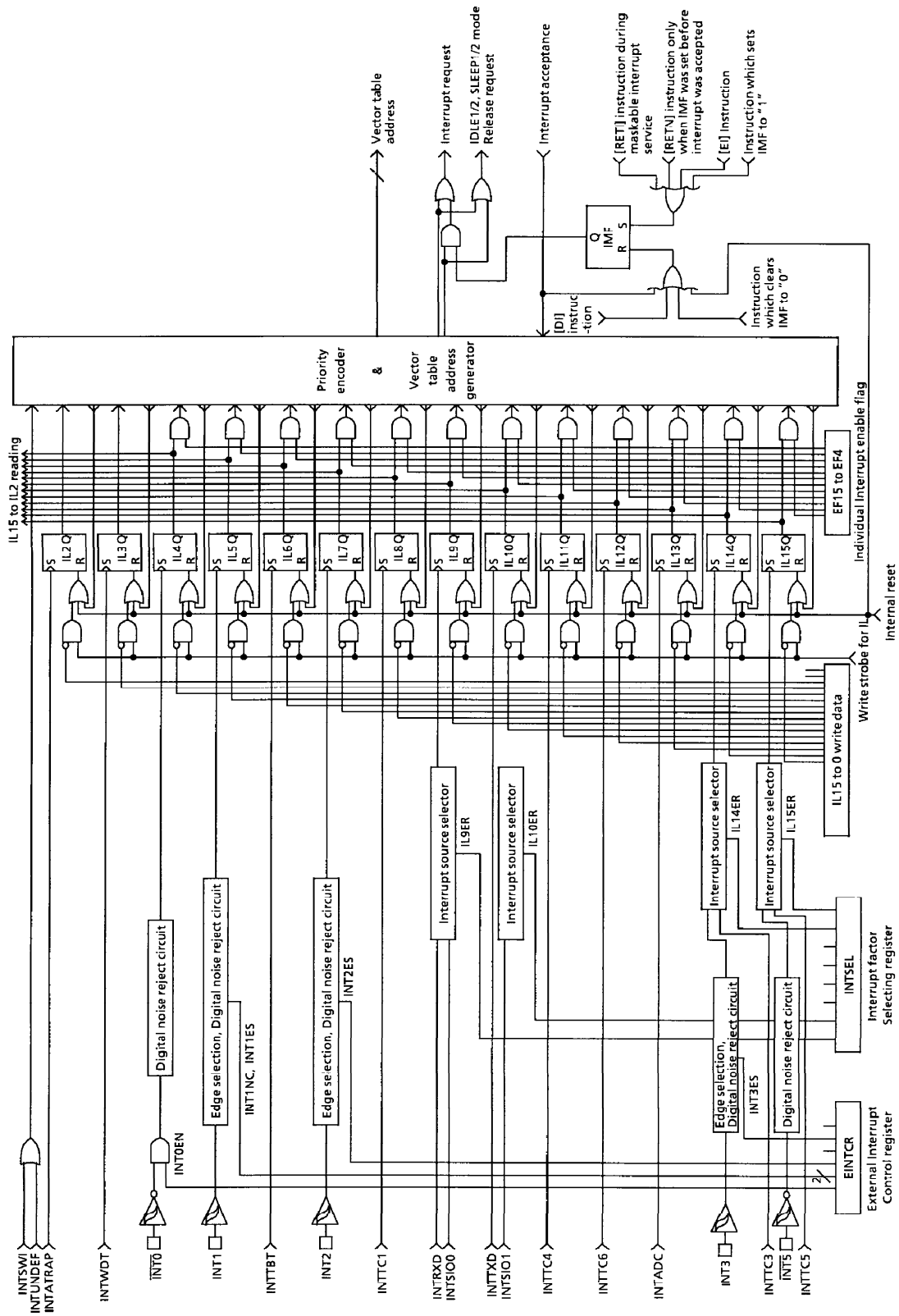


Figure 1-17. Interrupt Controller Block Diagram

(1) Interrupt Latches (IL15 to IL2)

An interrupt latch is provided for each interrupt source, except for a software interrupt. When interrupt request is generated, the latch is set to “1”, and the CPU is requested to accept the interrupt if its interrupt is enabled. All interrupt latches are initialized to “0” during reset.

The interrupt latches are located on address 003CH and 003DH in SFR area. Except for IL3 and IL2, each latch can be cleared to “0” individually by instruction (However, the read-modify-write instructions such as bit manipulation or operation instructions cannot be used. Interrupt request would be cleared inadequately if interrupt is requested while such instructions are executed.). Thus interrupt request can be canceled/initialized by software.

Interrupt latches are not set to “1” by an instruction. Since interrupt latches can be read, the status for interrupt requests can be monitored by software.

Note: When manipulating IL, clear IMF (to disable interrupts) beforehand.

Example 1: Clears interrupt latches

```
DI                ; IMF ← 0
LDW (ILL), 1110100000111111B ; IL12, IL10 to IL6 ← 0
EI                ; IMF ← 1
```

Example 2: Reads interrupt latches

```
LD  WA, (ILL)    ; W ← ILH, A ← ILL
```

Example 3: Tests an interrupt latches

```
TEST (ILL). 7    ; if IL7 = 1 then jump
JR  F, SSET
```

(2) Interrupt Enable Register (EIR)

The interrupt enable register (EIR) enables and disables the acceptance of interrupts, except for the non-maskable interrupts (software interrupt, undefined instruction interrupt, address trap interrupt and watchdog interrupt). Non-maskable interrupt is accepted regardless of the contents of the EIR.

The EIR consists of an interrupt master enable flag (IMF) and the individual interrupt enable flags (EF). These registers are located on address 003C_H and 003D_H in SFR area, and they can be read and written by an instructions (including read-modify-write instructions such as bit manipulation or operation instructions).

a) Interrupt master enable flag (IMF)

The interrupt enable register (IMF) enables and disables the acceptance of the whole maskable-interrupt. While IMF = “0”, all maskable interrupts are not accepted regardless of the status on each individual interrupt enable flag (EF). By setting IMF to “1”, the interrupt becomes acceptable if the individuals are enabled. When an interrupt is accepted, IMF is cleared to “0” after the latest status on IMF is stacked. Thus the maskable interrupts which follow are disabled. By executing return interrupt instruction [RETI/RETN], the stacked data, which was the status before interrupt acceptance, is loaded on IMF again.

The IMF is located on bit0 in EIRL (address: 003AH in SFR), and can be read and written by an instruction. The IMF is normally set and cleared by [EI] and [DI] instruction respectively. During reset, the IMF is initialized to “0”, and maskable interrupts are not accepted until it is set to “1”.

b) Individual interrupt enable flags (EF15 to EF4)

Each of these flags enables and disables the acceptance of its maskable interrupt. Setting the corresponding bit of an individual interrupt enable flag to “1” enables acceptance of its interrupt, and setting the bit to “0” disables acceptance. The individual interrupt enable flags (EF15 to EF4) are located on EIRL to EIRH (address: 003AH to 003BH in SFR), and can be read and written by an instruction. During reset, all the individual interrupt enable flags (EF15 to EF4) are initialized to “0” and all maskable interrupts are not accepted until they are set to “1”.

Note: Before manipulating EF, be sure to clear IMF (interrupt disabled). Then set IMF newly again after operating on the interrupt enables flag (EF). Normally, IMF is clear to “0” automatically on service routine. When IMF is set to “1” for using a multiple interrupt on service routine, be sure to process as is the case with EF.

Example 1: Enables interrupts individually and sets IMF

```

DI                               ; IMF ← 0
LDW (EIRL), 1110100010100000B   ; EF15 to EF13, EF11, EF7, EF5 ← 1
:                                 Note: IMF is not set.
:
EI                               ; IMF ← 1
    
```

Example 2: C compiler description example

```

unsigned int __io (3AH) EIRL;     ; /* 3AH shows EIRL address */
__DI ();
EIRL = 10100000B;
:
__EI ();
    
```

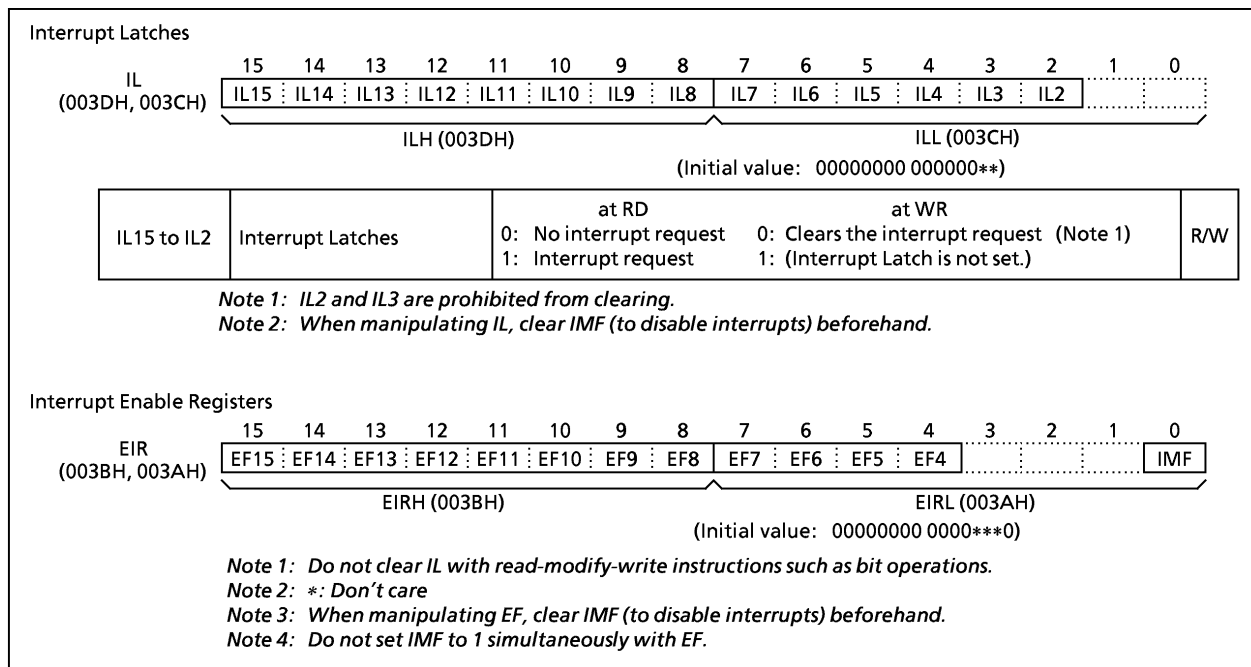


Figure 1-18. Interrupt Latch (IL), Interrupt Enable Registers (EIR)

(3) Selecting interrupt factor (INTSEL)

Each interrupt factor, that shares its interrupt source with other factors, enables its interrupt latch (IL) only if it is selected on INTSEL. The interrupt controller does not hold the interrupt request, while the factor generates the interrupt request is not selected on INTSEL. Therefore, set INTSEL appropriately before interrupt factors arises.

Interrupt source selector								
INTSEL (003EH)	7	6	5	4	3	2	1 0	
	—	IL9ER	IL10ER	—	—	—	IL14ER IL15ER	(Initial value: *00* **00)
	IL9ER	alternative of INTRXD or INTSIO0		0: INTRXD 1: INTSIO0		R/W		
	IL10ER	alternative of INTRXD or INTSIO1		0: INTRXD 1: INTSIO1				
	IL14ER	alternative of INT3 or INTTC3		0: INT3 1: INTTC3				
	IL15ER	alternative of INT5 or INTTC5		0: INT5 1: INTTC5				

Figure 1-19. Interrupt Source Selector

1.5.1 Interrupt Sequence

An interrupt request, which raised interrupt latch, is held, until interrupt is accepted or interrupt latch is cleared to “0” by resetting or an instruction. Interrupt acceptance sequence requires 8 machine cycles (4 μ s at 8.0 MHz) after the completion of the current instruction. The interrupt service task terminates upon execution of an interrupt return instruction [RETI] (for maskable interrupts) or [RETN] (for non-maskable interrupts). Figure 1-18 shows the timing chart of interrupt acceptance processing.

- (1) Interrupt acceptance processing is packaged as follows.
 - a) The interrupt master enable flag (IMF) is cleared to “0” in order to disable the acceptance of any following interrupt.
 - b) The interrupt latch (IL) for the interrupt source accepted is cleared to “0”.
 - c) The contents of the program counter (PC) and the program status word, including the interrupt master enable flag (IMF), are saved (pushed) on the stack in sequence of PSW + IMF, PCH, PCL. Meanwhile, the stack pointer (SP) is decremented by 3.
 - d) The entry address (interrupt vector) of the corresponding interrupt service program, loaded on the vector table, is transferred to the program counter.
 - e) The instruction stored at the entry address of the interrupt service program is executed.

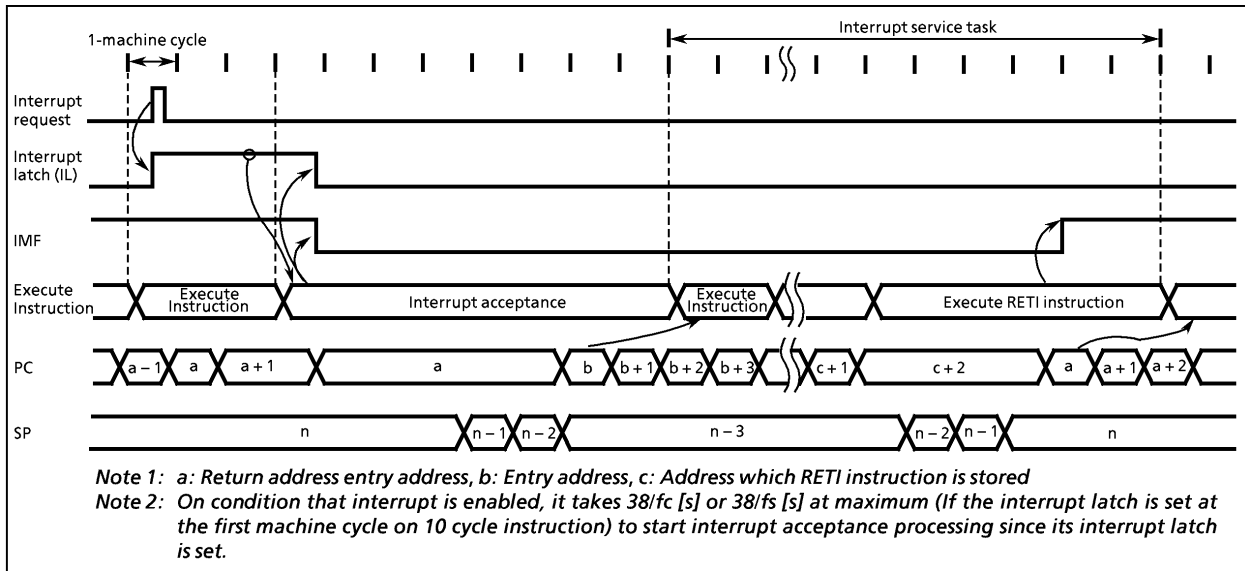
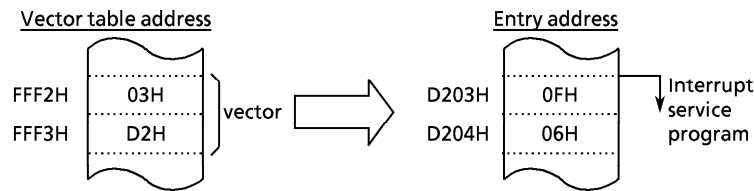


Figure 1-20. Timing Chart of Interrupt Acceptance/Return Interrupt Instruction

Example: Correspondence between vector table address for INTTBT and the entry address of the interrupt service program



A maskable interrupt is not accepted until the IMF is set to “1” even if the maskable interrupt higher than the level of current servicing interrupt is requested.

In order to utilize nested interrupt service, the IMF is set to “1” in the interrupt service program. In this case, acceptable interrupt sources are selectively enabled by the individual interrupt enable flags. To avoid overloaded nesting, clear the individual interrupt enable flag whose interrupt is currently serviced, before setting IMF to “1”. As for non-maskable interrupt, keep interrupt service shorten compared with length between interrupt requests; otherwise the status cannot be recovered as non-maskable interrupt would simply nested.

(2) Saving/restoring general-purpose registers

During interrupt acceptance processing, the program counter (PC) and the program status word (PSW, includes IMF) are automatically saved on the stack, but the accumulator and others are not. These registers are saved by software if necessary. When multiple interrupt services are nested, it is also necessary to avoid using the same data memory area for saving registers. The following methods are used to save/restore the general-purpose registers.

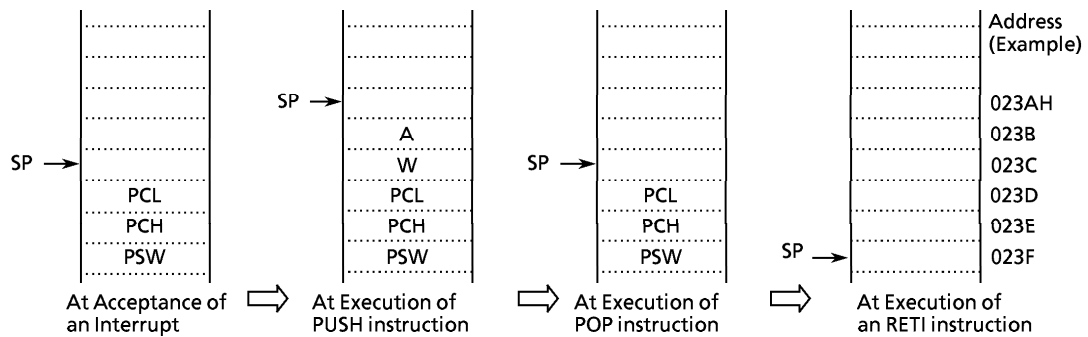
a) Using PUSH and POP instructions

To save only a specific register, PUSH and POP instructions are available.

Example: Save/store register using PUSH and POP instructions

```

PINTxx: PUSH  WA      ; Save WA register
          (interrupt processing)
          POP   WA      ; Restore WA register
          RETI         ; RETURN
    
```



b) Using data transfer instructions

To save only a specific register without nested interrupts, data transfer instructions are available.

Example: Save/store register using data transfer instructions

```

PINTxx: LD  (GSAVA), A      ; Save A register
          (interrupt processing)
          LD  A, (GSAVA)    ; Restore A register
          RETI             ; RETURN
    
```

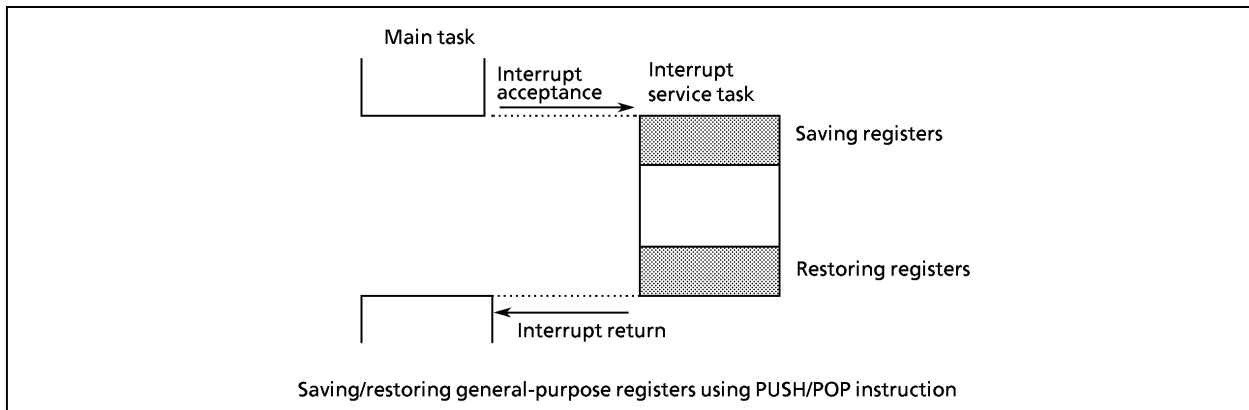


Figure 1-21. Saving/Restoring General-purpose Registers Under Interrupt Processing

(3) Interrupt return

Interrupt return instructions [RETI]/[RETN] perform as follows.

[RETI]/[RETN] Interrupt Return
① Program Counter (PC) and program status word (PSW, includes IMF) are restored from the stack.
② Stack pointer (SP) is incremented by 3.

As for Address Trap interrupt (INTARTAP), it is required to alter stacked data for program counter (PC) to restarting address, during interrupt service program. Otherwise returning interrupt causes INTATRAP again. When interrupt acceptance processing has completed, stacked data for PCL and PCH are located on address (SP + 1) and (SP + 2) respectively.

Example 1: Returning from address trap interrupt (INTATRAP) service program

```
PINTxx: POP   WA           ; Recover SP by 2
        LD    WA, Return Address ;
        PUSH  WA           ; Alter stacked data
        (interrupt processing)
        RETN                ; RETURN
```

Example 2: Restarting without returning interrupt

(In this case, PSW (includes IMF) before interrupt acceptance is discarded.)

```
PINTxx: INC   SP           ; Recover SP by 3
        INC   SP           ;
        INC   SP           ;
        (interrupt processing)
        LD    EIRL, data    ; Set IMF to "1" or clear it to "0"
        JP    Restart Address ; Jump into restarting address
```

It is recommended that stack pointer be return to rate before INTATRAP (increment 3 times), if return interrupt instruction [RETN] is not utilized during interrupt service program under INTATRAP (such as Example 2).

Interrupt requests are sampled during the final cycle of the instruction being executed. Thus, the next interrupt can be accepted immediately after the interrupt return instruction is executed.

Note: When the interrupt processing time is longer than the interrupt request generation time, the interrupt service task is performed but not the main task.

1.5.2 Software Interrupt (INTSW)

Executing the [SWI] instruction generates a software interrupt and immediately starts interrupt processing (INTSW is highest prioritized interrupt).

Use the [SWI] instruction only for detection of the address error or for debugging.

① Address error detection

FF_H is read if for some cause such as noise the CPU attempts to fetch an instruction from a non-existent memory address during single chip mode. Code FF_H is the SWI instruction, so a software interrupt is generated and an address error is detected. The address error detection range can be further expanded by writing FF_H to unused areas of the program memory. Address-trap reset is generated in case that an instruction is fetched from RAM or SFR areas.

② Debugging

Debugging efficiency can be increased by placing the SWI instruction at the software break point setting address.

1.5.3 Undefined Instruction Interrupt (INTUNDEF)

Taking code which is not defined as authorized instruction for instruction causes INTUNDEF. INTUNDEF is generated when the CPU fetches such a code and tries to execute it. INTUNDEF is accepted even if non-maskable interrupt is in process. Contemporary process is broken and INTUNDEF interrupt process starts, soon after it is requested.

Note: The undefined instruction interrupt (INTUNDEF) forces CPU to jump into vector address, as software interrupt (SWI) does.

1.5.4 Address Trap Interrupt (INTATRAP)

Fetching instruction from unauthorized area for instructions (address trapped area) causes reset-output or address trap interrupt (INTATRAP). INTATRAP is accepted even if non-maskable interrupt is in process. Contemporary process is broken and INTATRAP interrupt process starts, soon after it is requested.

The address trapped area is alternative: SFR and RAM, or SFR only.

Note: The operating mode under address trapped, whether to be reset-output or interrupt processing, is selected on watchdog timer control register (WDTCR).

1.5.5 External Interrupts

The TMP86CM25/S25 have five external interrupt inputs. These inputs are equipped with digital noise reject circuits (pulse inputs of less than a certain time are eliminated as noise).

Edge selection is also possible with INT1 to INT3. The $\overline{\text{INT0}}/\text{P63}$ pin can be configured as either an external interrupt input pin or an input/output port, and is configured as an input port during reset.

Edge selection, noise reject control and $\overline{\text{INT0}}/\text{P63}$ pin function selection are performed by the external interrupt control register (EINTCR).

Table 1-3. External Interrupts

Source	Pin	Secondary function pin	Enable conditions	Edge	Digital noise reject
INT0	$\overline{\text{INT0}}$	P63/AIN3	IMF = 1, EF4 = 1, INT0EN = 1	Falling edge	Pulses of less than $2/f_c$ [s] are eliminated as noise. Pulses of $7/f_c$ [s] or more are considered to be signals. In the SLOW or the SLEEP mode, pulses of less than $1/f_s$ [s] are eliminated as noise. Pulses of $3.5/f_s$ [s] or more are considered to be signals.
INT1	MUL4 (Note 4)	P12/SEG54 or P34/COM5	IMF · EF5 = 1	Falling edge or Rising edge	Pulses of less than $15/f_c$ or $63/f_c$ [s] are eliminated as noise. Pulses of $49/f_c$ or $193/f_c$ [s] or more are considered to be signals. In the SLOW or the SLEEP mode, pulses of less than $1/f_s$ [s] are eliminated as noise. Pulses of $3.5/f_s$ [s] or more are considered to be signals.
INT2	MUL5 (Note 4)	P13/SEG55 or P35/COM6	IMF · EF7 = 1		
INT3	MUL6 (Note 4)	P14/SEG56 or P36/COM7	IMF · EF14 = 1 IL14ER = 0		
INT5	$\overline{\text{INT5}}$	P20/ $\overline{\text{STOP}}$	IMF · EF15 = 1 IL15ER = 0	Falling edge	Pulses of less than $2/f_c$ [s] are eliminated as noise. Pulses of $7/f_c$ [s] or more are considered to be signals. In the SLOW or the SLEEP mode, pulses of less than $1/f_s$ [s] are eliminated as noise. Pulses of $3.5/f_s$ [s] or more are considered to be signals.

Note 1: If a noiseless signal is input to the external interrupt pin in the NORMAL 1/2 or IDLE 1/2 mode, the maximum time from the edge of input signal until the IL is set is as follows :

- ① INT1 pin $55/f_c$ [s] (INT1NC = 1), $199/f_c$ [s] (INT1NC = 0)
- ② INT2, INT3 pin $31/f_c$ [s]

Note 2: Even if the falling edge of $\overline{\text{INT0}}$ pin input is detected at INT0EN = 0, the interrupt latch IL₄ is not set.

Note 3: When data changed and did a change of I/O when used external interrupt ports as a normal ports, interrupt request signal occurs incorrectly. Handling of prohibition of interrupt enable register (EIR) is necessary.

Note 4: MUL4 to 6 is interrupt input can be changed by the MULSEL register.

Note 5: The maximum time from modifying INT1NC until a noise reject time is changed is $2^6/f_c$.

External interrupt control register

EINTCR (0037H)	7	6	5	4	3	2	1	0	(Initial value: 00** 000*)
	INT1NC	INT0EN			INT3ES	INT2ES	INT1ES		
INT1NC	Noise reject time select			0: Pulses of less than $63/f_c$ [s] are eliminated as noise 1: Pulses of less than $15/f_c$ [s] are eliminated as noise				R/W	
INT0EN	P63/ $\overline{\text{INT0}}$ pin configuration			0: P63 input/Output port 1: $\overline{\text{INT0}}$ pin (Port P63 should be set to an input mode)					
INT3 ES INT2 ES INT1 ES	INT3 to INT1 edge select			0: Rising edge 1: Falling edge					

Note 1: f_c : High-frequency clock [Hz], *: Don't care

Note 2: When the system clock frequency is switched between high and low or when the external interrupt control register (EINTCR) is overwritten, the noise canceller may not operate normally. It is recommend that external interrupts are disabled using the interrupt enable register (EIR).

Figure 1-22. External Interrupt Control Register

1.6 Reset Circuit

The TMP86CM25/S25 have four types of reset generation procedures: an external reset input, an address trap reset, a watchdog timer reset and a system clock reset. Table 1-4 shows on-chip hardware initialization by reset action.

The malfunction reset circuit such as watchdog timer reset, address trap reset and system clock reset is not initialized when power is turned on. The reset operation might occur for the maximum $24/f_c$ [s] ($1.5 \mu\text{s}$ at 16.0 MHz) when power is turned on.

Table 1-4. Initializing Internal Status by Reset Action

On-chip hardware	Initial value	On-chip hardware	Initial value
Program counter (PC)	(FFEH)	Prescaler and Divider of timing generator	0
Stack pointer (SP)	Not initialized		
General-purpose registers (W, A, B, C, D, E, H, L)	Not initialized		
Jump status flag (JF)	Not initialized	Watchdog timer	Enable
Zero flag (ZF)	Not initialized	Output latches of I/O ports	Refer to I/O port circuitry
Carry flag (CF)	Not initialized		
Half carry flag (HF)	Not initialized		
Sign flag (SF)	Not initialized		
Overflow flag (VF)	Not initialized		
Interrupt master enable flag (IMF)	0		
Interrupt individual enable flags (EF)	0	Control registers	Refer to each of control register
Interrupt latches (IL)	0		
		RAM	Not initialized

1.6.1 External Reset Input

The $\overline{\text{RESET}}$ pin contains a Schmitt trigger (hysteresis) with an internal pull-up resistor.

When the $\overline{\text{RESET}}$ pin is held at “L” level for at least 3 machine cycles ($12/f_c$ [s]) with the power supply voltage within the operating voltage range and oscillation stable, a reset is applied and the internal state is initialized.

When the $\overline{\text{RESET}}$ pin input goes high, the reset operation is released and the program execution starts at the vector address stored at addresses FFFE to FFFFH.

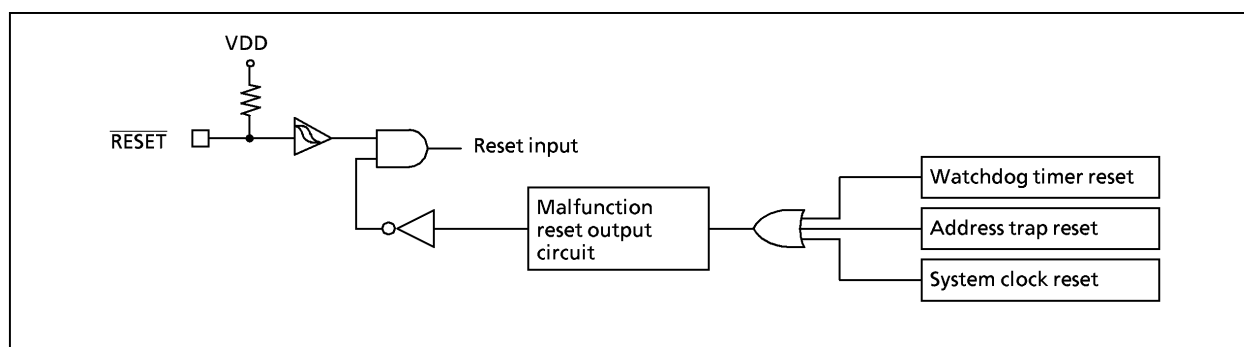


Figure 1-23. Reset Circuit

1.6.2 Address-Trap-Reset

If the CPU should start looping for some cause such as noise and an attempt be made to fetch an instruction from the on-chip RAM (when $WDTCR1<ATAS>$ is set to "1") or the SFR area, address-trap-reset will be generated. The reset time is about $8/f_c$ to $24/f_c$ [s] (0.5 to $1.5 \mu\text{s}$ at 16.0 MHz).

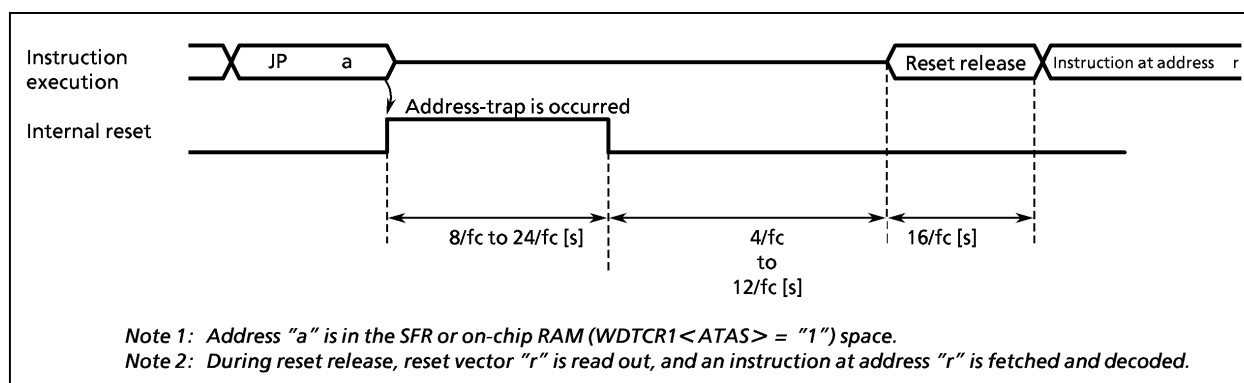


Figure 1-24. Address-Trap-Reset

Note: The operating mode under address trapped is alternative of reset or interrupt. The address trap area is alternative.

1.6.3 Watchdog Timer Reset

Refer to Section "2.4 Watchdog Timer".

1.6.4 System-Clock-Reset

Clearing both XEN and XTEN (bits 7 and 6 in SYSCR2) to "0", clearing XEN to "0" when $SYSCK=0$, or clearing XTEN to "0" when $SYSCK="1"$ stops system clock, and causes the microcomputer to deadlock. This can be prevented by automatically generating a reset signal whenever $XEN=XTEN="0"$, $XEN=SYSCK="0"$, or $XTEN="0"/SYSCK="1"$ is detected to continue the oscillation. The reset time is about $8/f_c$ to $24/f_c$ [s] (0.5 to $1.5 \mu\text{s}$ at 16.0 MHz).

2. On-Chip Peripherals Functions

2.1 Special Function Registers (SFRs)

The TMP86CM25/S25 adopt the memory mapped I/O system, and all peripheral control and data transfers are performed through the special function register (SFR). The SFR is mapped on address 0000H to 003FH, DBR is mapped on address 0F00H to 0FFFH.

Figure 2-1 (a), (b) indicate the special function register (SFR) and data buffer register (DBR) for TMP86CM25/S25.

Address	Read	Write	Address	Read	Write
0000H	Reserved		0020H	ADCDR1 (AD converter register 1)	-
01	P1DR (P1 Port output latch)	-	21	ADCDR2 (AD converter register 2)	-
02	P2DR (P2 Port output latch)	-	22	Reserved	
03	P3DR (P3 Port output latch)	-	23	Reserved	
04	P3LCR (P3 Port control register)	-	24	Reserved	
05	P5DR (P5 Port output latch)	-	25	UARTSR (UART Status register)	UARTCR1 (UART control register 1)
06	P6DR (P6 Port output latch)	-	26	-	UARTCR2 (UART control register 2)
07	P7DR (P7 Port output latch)	-	27	LCDCTL1 (LCD Control register 1)	
08	P1PRD (P1 Terminal input)	-	28	LCDCTL2 (LCD Control register 2)	
09	P2PRD (P2 Terminal input)	-	29	P1LCR (P1 Control register)	
0A	P3PRD (P3 Terminal input)	-	2A	P5LCR (P5 Control register)	
0B	P5PRD (P5 Terminal input)	-	2B	P7LCR (P7 Control register)	
0C	P6CR (P6 Port input/output control)		2C	PWREG3 (Timer register 3)	
0D	P7PRD (P7 Terminal input)	-	2D	PWREG4 (Timer register 4)	
0E	ADCCR1 (AD control register 1)		2E	PWREG5 (Timer register 5)	
0F	ADCCR2 (AD control register 2)		2F	PWREG6 (Timer register 6)	
10	TREG1AL		30	Reserved	
11	TREG1AM (Timer register 1A)		31	Reserved	
12	TREG1AH		32	Reserved	
13	TREG1B (Timer register 1B)		33	Reserved	
14	TC1CR1 (Timer Counter 1 control 1)		34	-	WDTCR1 (Watchdog timer control)
15	TC1CR2 (Timer Counter 1 control 2)		35	-	WDTCR2 (Watchdog timer control)
16	TC1SR (TC1 Status)	-	36	TBTCR (TBT/TG/DVO control)	
17	Reserved		37	EINTCR (External interrupt control)	
18	TC3CR (Timer Counter 3 control)		38	SYSCR1 (System control 1)	
19	TC4CR (Timer Counter 4 control)		39	SYSCR2 (System control 2)	
1A	TC5CR (Timer Counter 5 control)		3A	EIRL (Interrupt enable register)	
1B	TC6CR (Timer Counter 6 control)		3B	EIRH (Interrupt enable register)	
1C	TTREG3 (Timer register 3)		3C	ILL (Interrupt latch)	
1D	TTREG4 (Timer register 4)		3D	ILH (Interrupt latch)	
1E	TTREG5 (Timer register 5)		3E	INTSEL (Interrupt source selector)	
1F	TTREG6 (Timer register 6)		3F	PSW (Program Status word)	

Figure 2-1 (a). The Special Function Register (SFR) for TMP86CM25/S25

LCD Data Buffer (Write/Read)								
0F00H	0F10H	0F20H	0F30H	0F40H	0F50H	0F60H	0F70H	COM0
0F01H	0F11H	0F21H	0F31H	0F41H	0F51H	0F61H	0F71H	COM1
0F02H	0F12H	0F22H	0F32H	0F42H	0F52H	0F62H	0F72H	COM2
0F03H	0F13H	0F23H	0F33H	0F43H	0F53H	0F63H	0F73H	COM3
0F04H	0F14H	0F24H	0F34H	0F44H	0F54H	0F64H	0F74H	COM4
0F05H	0F15H	0F25H	0F35H	0F45H	0F55H	0F65H	0F75H	COM5
0F06H	0F16H	0F26H	0F36H	0F46H	0F56H	0F66H	0F76H	COM6
0F07H	0F17H	0F27H	0F37H	0F47H	0F57H	0F67H	0F77H	COM7
0F08H	0F18H	0F28H	0F38H	0F48H	0F58H	0F68H	0F78H	COM8
0F09H	0F19H	0F29H	0F39H	0F49H	0F59H	0F69H	0F79H	COM9
0F0AH	0F1AH	0F2AH	0F3AH	0F4AH	0F5AH	0F6AH	0F7AH	COM10
0F0BH	0F1BH	0F2BH	0F3BH	0F4BH	0F5BH	0F6BH	0F7BH	COM11
0F0CH	0F1CH	0F2CH	0F3CH	0F4CH	0F5CH	0F6CH	0F7CH	COM12
0F0DH	0F1DH	0F2DH	0F3DH	0F4DH	0F5DH	0F6DH	0F7DH	COM13
0F0EH	0F1EH	0F2EH	0F3EH	0F4EH	0F5EH	0F6EH	0F7EH	COM14
0F0FH	0F1FH	0F2FH	0F3FH	0F4FH	0F5FH	0F6FH	0F7FH	COM15

SEG7	SEG15	SEG23	SEG31	SEG39	SEG47	SEG55	SEG59
to	to	to	to	to	to	to	to
SEG0	SEG8	SEG16	SEG24	SEG32	SEG40	SEG48	SEG56

Address	Read	Write
0F90H	SIO0BR0 (SIO0 Buffer 0)	
91	SIO0BR1 (SIO0 Buffer 1)	
92	SIO0BR2 (SIO0 Buffer 2)	
93	SIO0BR3 (SIO0 Buffer 3)	
94	SIO0BR4 (SIO0 Buffer 4)	
95	SIO0BR5 (SIO0 Buffer 5)	
96	SIO0BR6 (SIO0 Buffer 6)	
97	SIO0BR7 (SIO0 Buffer 7)	
98	—	SIO0CR1 (SIO0 Control register 1)
99	SIO0SR (SIO0 Status register)	SIO0CR2 (SIO0 Control register 2)
9A	—	STOPCR (Key On Wake Up Control register)
9B	RDBUF (uart received data buffer)	TDBUF (uart transmitted data buffer)
9C	Reserved	
9D	Reserved	
9E	Reserved	
9F	Reserved	
A0	SIO1BR0 (SIO1 Buffer 0)	
A1	SIO1BR1 (SIO1 Buffer 1)	
A2	SIO1BR2 (SIO1 Buffer 2)	
A3	SIO1BR3 (SIO1 Buffer 3)	
A4	SIO1BR4 (SIO1 Buffer 4)	
A5	SIO1BR5 (SIO1 Buffer 5)	
A6	SIO1BR6 (SIO1 Buffer 6)	
A7	SIO1BR7 (SIO1 Buffer 7)	
A8	—	SIO1CR1 (SIO1 Control register 1)
A9	SIO1SR (SIO1 Status register)	SIO1CR2 (SIO1 Control register 2)
AA	Reserved	
⋮	⋮	
BF	Reserved	
C0	MULSEL (Multiplexed Function Select Register)	
C1	Reserved	
⋮	⋮	
OFFF	Reserved	

Figure 2-1 (b). The Data Buffer Register (DBR) for TMP86CM25/S25

2.2 I/O Ports

The TMP86CM25/S25 have 6 parallel input/output ports (42 pins) as follows.

	Primary Function	Secondary Functions
Port P1	8-bit I/O port	External interrupt input, serial interface input/output, UART input/output and segment output.
Port P2	3-bit I/O port	Low-frequency resonator connections, external interrupt input, STOP mode release signal input.
Port P3	7-bit I/O port	Timer/Counter input/output and divider output and segment/common output.
Port P5	8-bit I/O port	Segment output.
Port P6	8-bit I/O port	Analog input, external interrupt input, timer/counter input and STOP mode release signal input.
Port P7	8-bit I/O port	Common output. Timer/Counter input/output and divider output.

Each output port contains a latch, which holds the output data. All input ports do not have latches, so the external input data should be externally held until the input data is read from outside or reading should be performed several timer before processing. Figure 2-2 shows input/output timing examples. External data is read from an I/O port in the S1 state of the read cycle during execution of the read instruction. This timing cannot be recognized from outside, so that transient input such as chattering must be processed by the program.

Output data changes in the S2 state of the write cycle during execution of the instruction which writes to an I/O port.

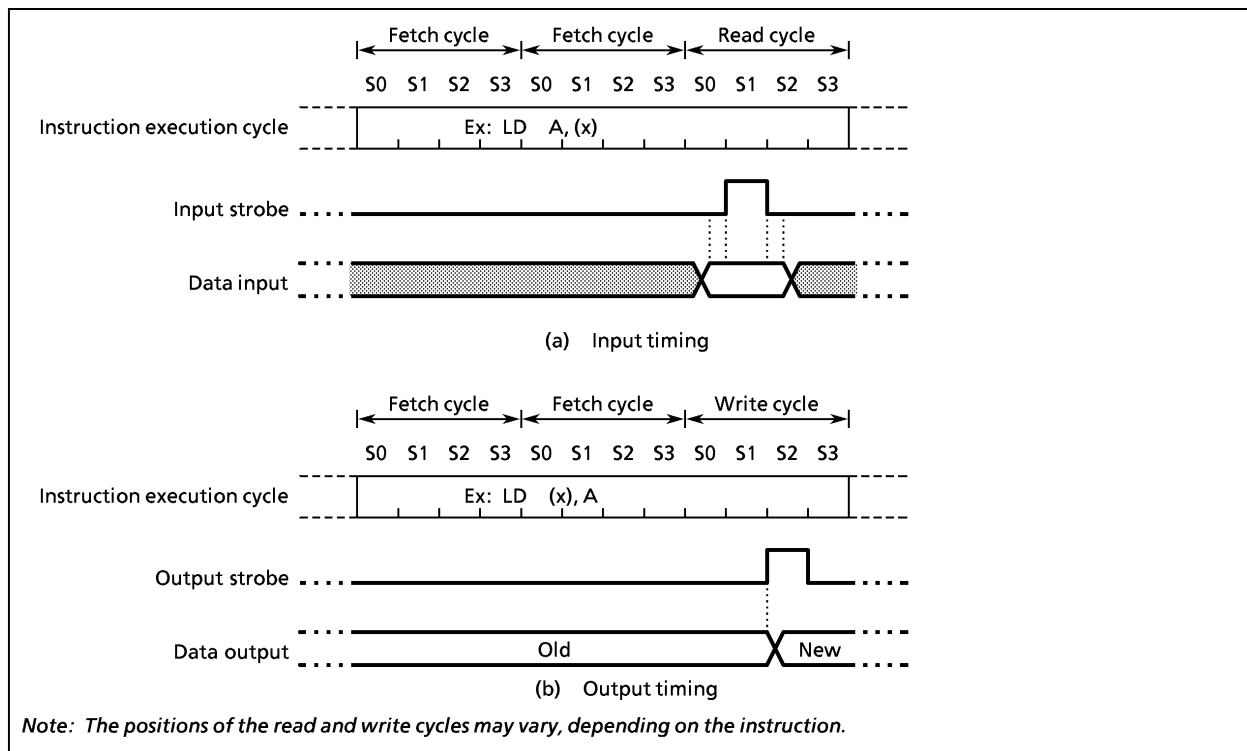


Figure 2-2. Input/Output Timing (Example)

2.2.1 Port P1 (P17 to P10)

Port P1 is an 8-bit input/output port which is also used as an external interrupt input, serial interface input/output, UART input/output and segment output of LCD. When used as a segment pins of LCD, the respective bit of P1LCR should be set to "1".

When used as an input port or a secondary function (except for segment) pins, the respective output latch (P1DR) should be set to "1" and its corresponding P1LCR bit should be set to "0". When used as an output port, the respective P1LCR bit should be set to "0". During reset, the output latch is initialized to "1".

P1 port output latch (P1DR) and P1 port terminal input (P1PRD) are located on their respective address.

When read the output latch data, the P1DR should be read and when read the terminal input data, the P1PRD register should be read.

If the terminal input data which is configured as LCD segment output is read, unstable data is read.

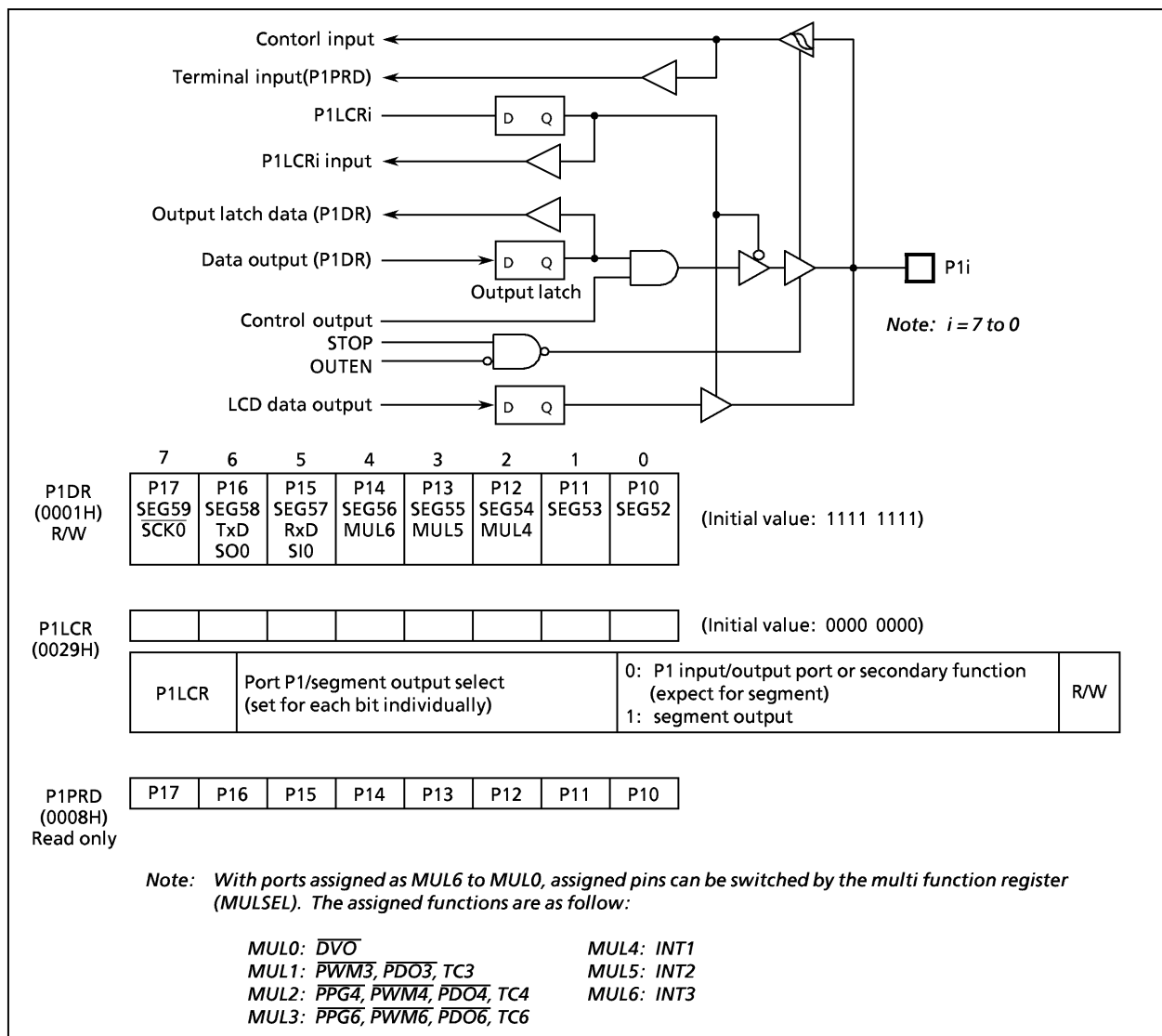


Figure 2-3. Port 1

2.2.2 Port P2 (P22 to P20)

Port P2 is a 3-bit input/output port.

It is also used as an external interrupt, a STOP mode release signal input, and low-frequency crystal oscillator connection pins. When used as an input port or a secondary function pins, respective output latch (P2DR) should be set to "1".

During reset, the P2DR is initialized to "1".

A low-frequency crystal oscillator (32.768 kHz) is connected to pins P21 (XTIN) and P22 (XTOUT) in the dual-clock mode. In the single-clock mode, pins P21 and P22 can be used as normal input/output ports.

It is recommended that pin P20 should be used as an external interrupt input, a STOP mode release signal input, or an input port. If it is used as an output port, the interrupt latch is set on the falling edge of the output pulse.

P2 port output latch (P2DR) and P2 port terminal input (P2PRD) are located on their respective address.

When read the output latch data, the P2DR should be read and when read the terminal input data, the P2PRD register should be read. If a read instruction is executed for port P2, read data of bits 7 to 3 are unstable.

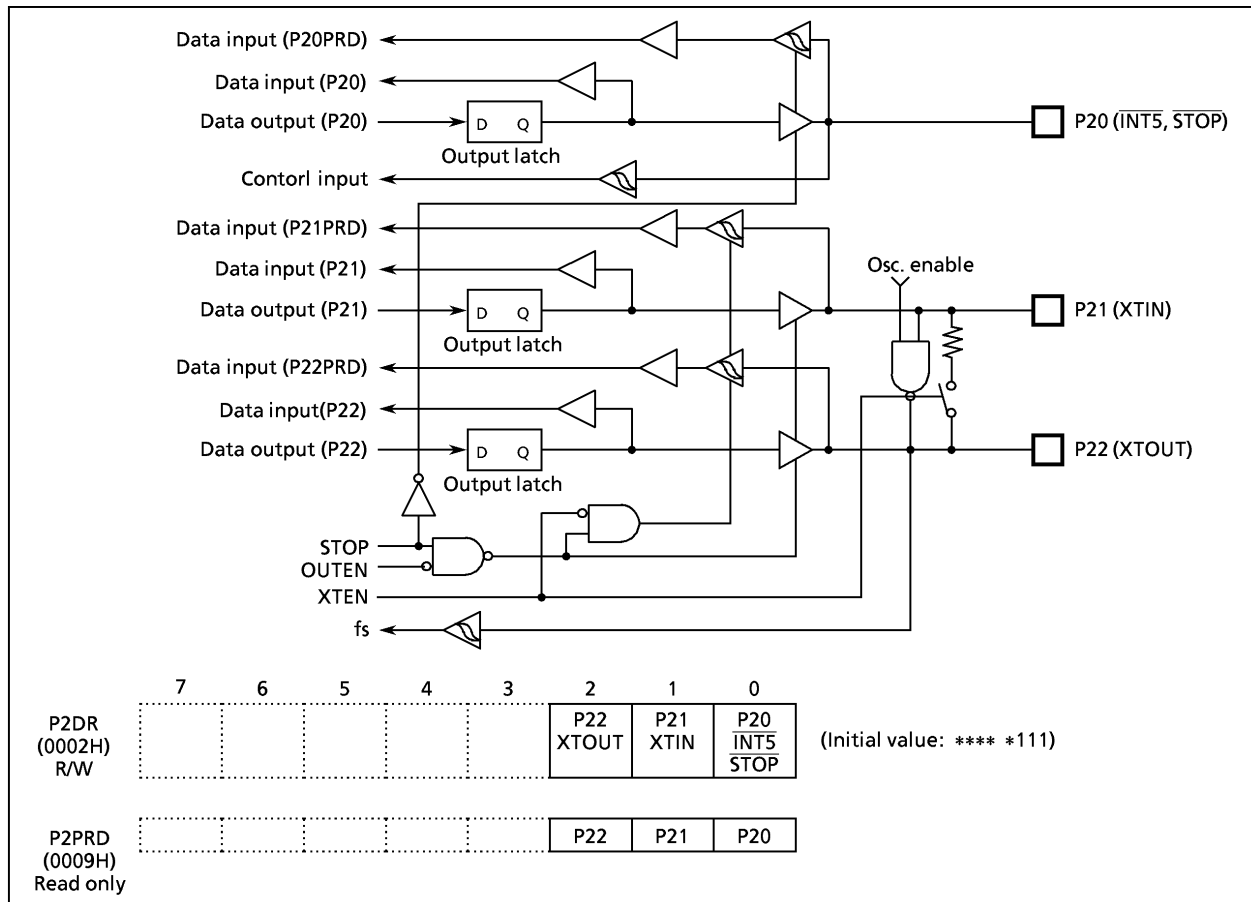


Figure 2-4. Port 2

Note: Port P20 is used as \overline{STOP} pin. Therefore, when stop mode is started, OUTEN does not affect to P20, and P20 becomes High-Z mode.

2.2.3 Port P3 (P36 to P30)

Port P3 is a 7-bit input/output port.

It is also used as a External interrupt input, timer/counter input/output, divider output and LCD common/segment output.

When used as an input port or a secondary function pins, after setting segment/common output control (P3LCR) to “0” respective output latch (P3DR) should be set to “1”. During reset, the P3DR is initialized to “1”, and segment output control (P3LCR) is initialized by “0”. In using it as LCD segment/common output, it sets the bit to which P3LCR corresponds to “1”.

P3 port output latch (P3DR) and P3 port terminal input (P3PRD) are located on their respective address.

When read the output latch data, the P3DR should be read and when read the terminal input data, the P3PRD register should be read. If a read instruction is executed for port P3, read data of bit 7 is unstable.

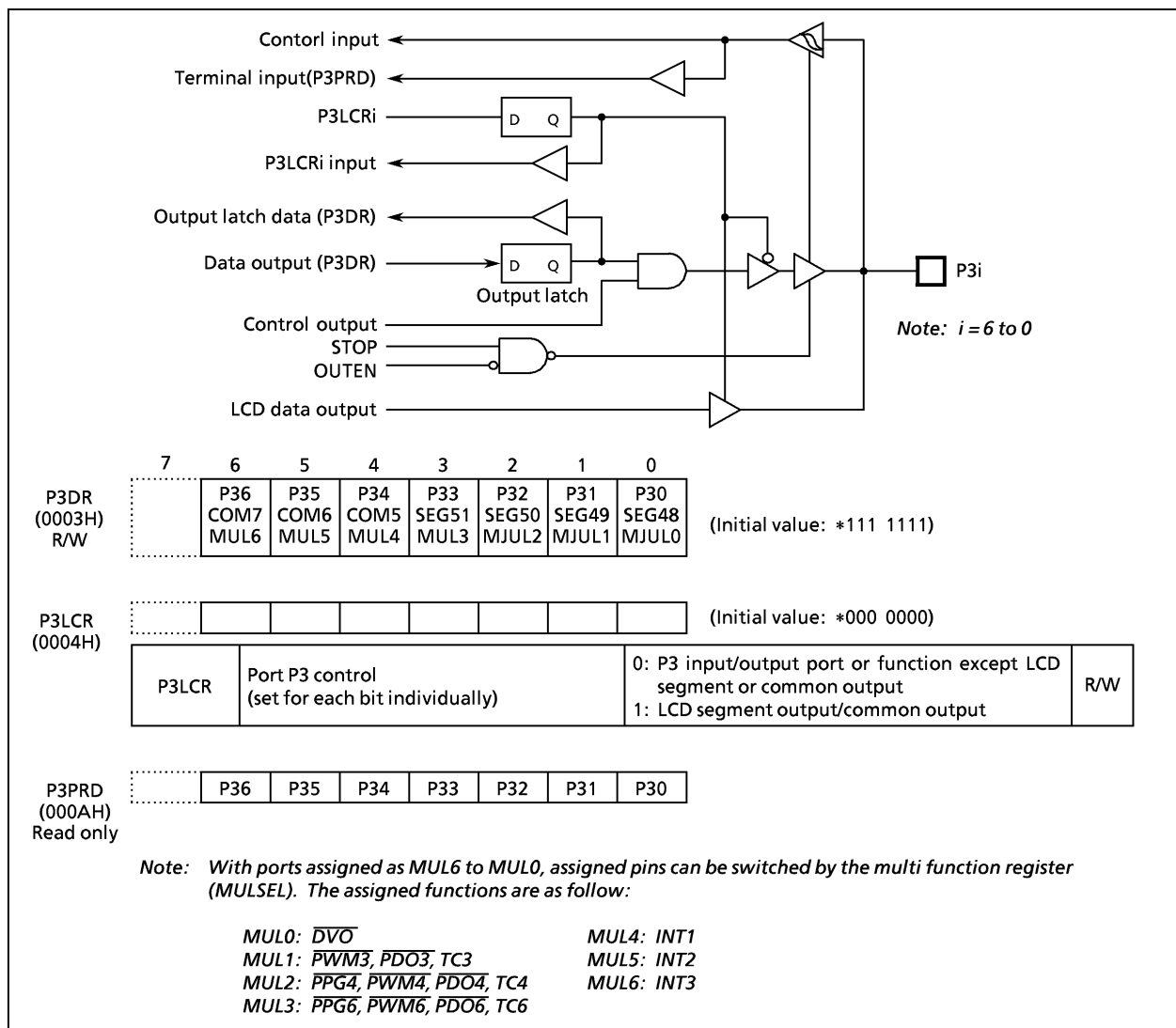


Figure 2-5. Port 3

2.2.4 Port P5 (P57 to P50)

Port P5 is an 8-bit input/output port which is also used as a segment pins of LCD.

When used as input port, the respective output latch (P5DR) should be set to “1”.

During reset, the P5DR is initialized to “1”.

When used as a segment pins of LCD, the respective bit of P5LCR should be set to “1”. When used as an output port, the respective P5LCR bit should be set to “0”.

P5 port output latch (P5DR) and P5 port terminal input (P5PRD) are located on their respective address.

When read the output latch data, the P5DR should be read and when read the terminal input data, the P5PRD register should be read. If the terminal input data which is configured as LCD segment output is read, unstable data is read.

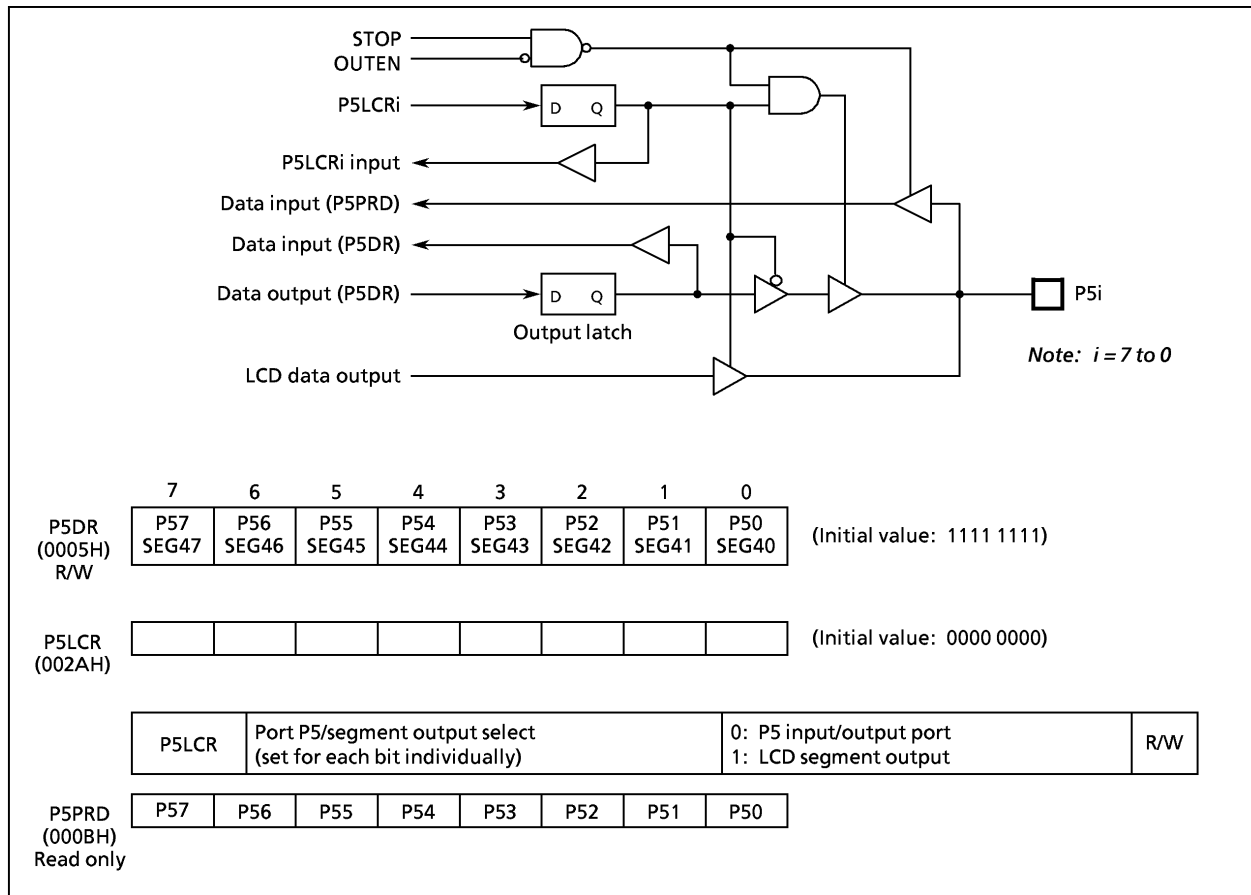


Figure 2-6. Port 5

2.2.5 Port P6 (P67 to P60)

Port P6 is an 8-bit input/output port which can be configured as an input or an output in one-bit unit. Port P6 is also used as an analog input, Key on Wake up input, timer/counter input and external interrupt input. Input/output mode is specified by the P6 control register (P6CR), the P6 output latch (P6DR), and AINDS (bit 4 in ADCCR1). During reset, P6CR and P6DR are initialized to "0" and AINDS is set to "1". At the same time, the input data of pins P67 to P60 are fixed to "0". To use port P6 as an input port, external interrupt input, timer/counter input or key on wake up input, set data of P6DR to "1" and P6CR to "0". To use it as an output port, set data of P6CR to "1". To use it as an analog input, set data of P6DR to "0" and P6CR to "0", and start the AD. It is the penetration electric current measures by the analog voltage.

Pins not used for analog input can be used as I/O ports. During AD conversion, output instructions should not be executed to keep a precision. In addition, a variable signal should not be input to a port adjacent to the analog input during AD conversion.

When the AD converter is in use (P6DR=0), bits mentioned above are read as "0" by executing input instructions.

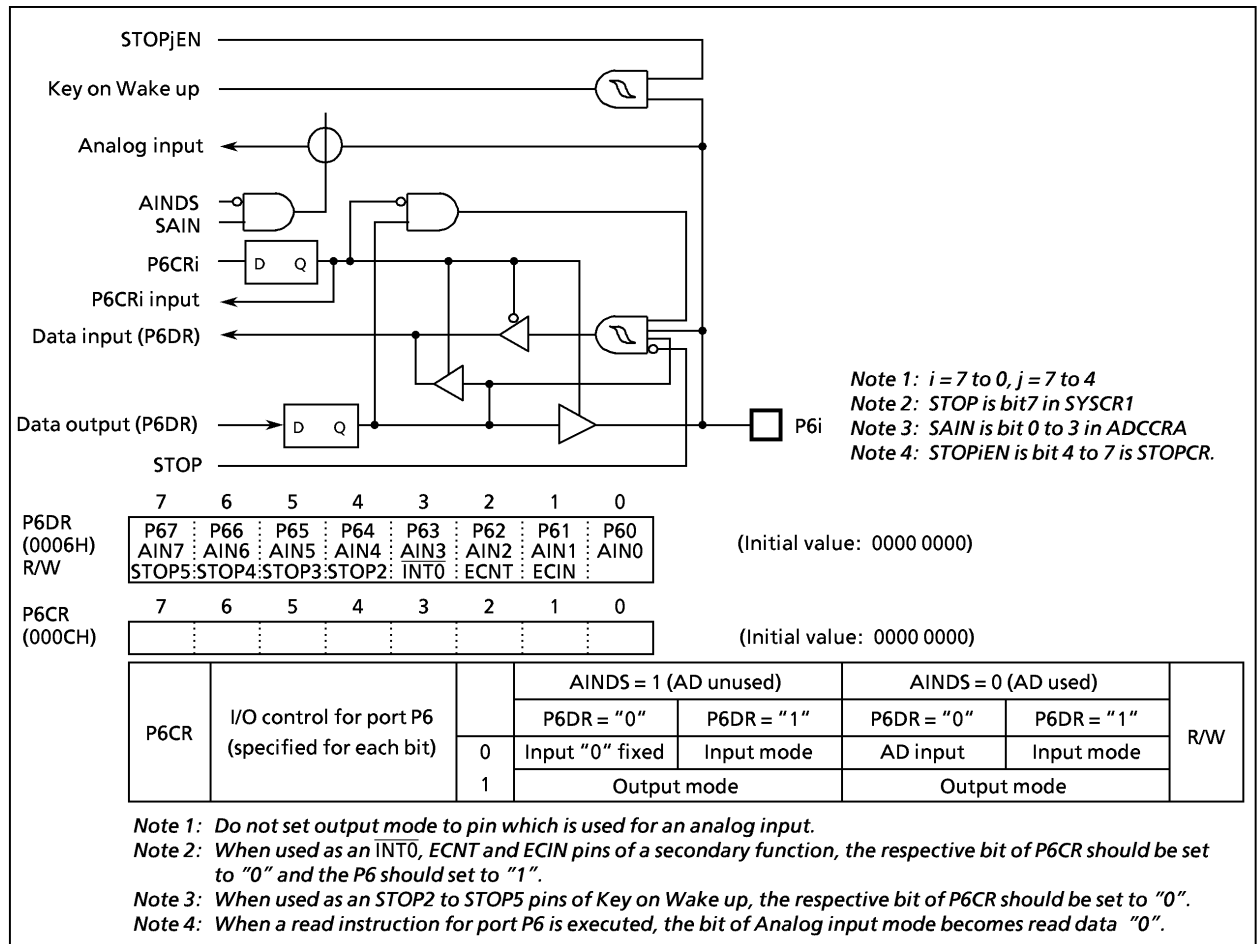


Figure 2-7. Port 6 and P6CR

Note: Although P6DR is a read/writer register, because it is also used as an input mode control function, read-modify-write instructions such as bit manipulate instructions cannot be used. Read-modify-write instruction writes the all data of 8-bit after data is read and modified. Because a bit setting Input mode read data of terminal, the output latch is changed by these instruction. So P6 port can not input data.

2.2.6 Port P7 (P77 to P70)

Port P7 is an 8-bit input/output port which is also used as an external interrupt input, a divider output a segment pins of LCD.

When used as input port or a secondary function pins, the respective output latch (P7DR) should be set to “1”.

During reset, the P7DR is initialized to “1”.

When used as a segment pins of LCD, the respective bit of P7LCR should be set to “1”. When used as an output port, the respective P7LCR bit should be set to “0”.

P7 port output latch (P7DR) and P7 port terminal input (P7PRD) are located on their respective address.

When read the output latch data, the P7DR should be read and when read the terminal input data, the P7PRD register should be read. If the terminal input data which is configured as LCD segment output is read, unstable data is read.

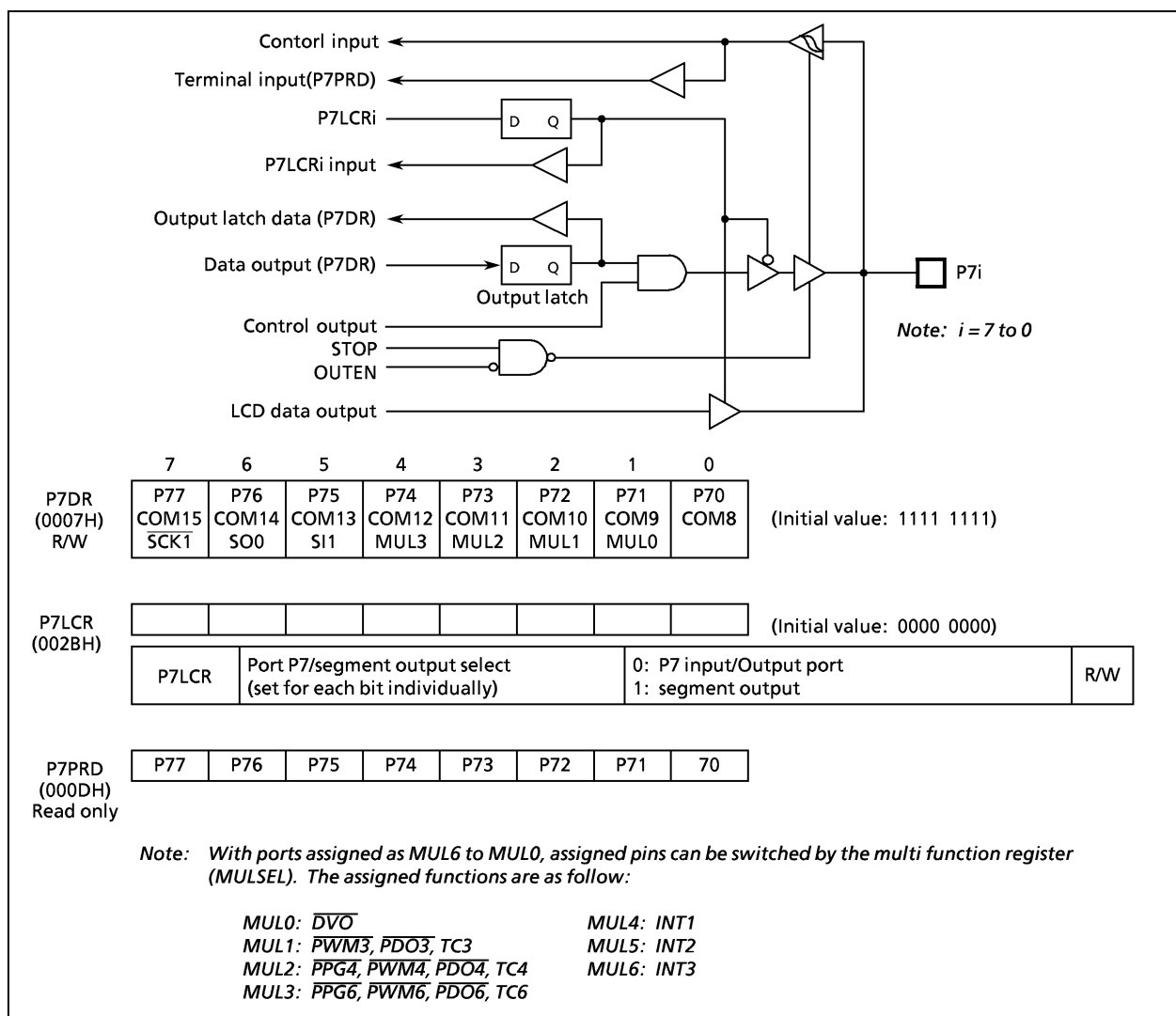


Figure 2-8. Port 7

2.3 Multi Function Register

With function pins assigned as MUL6 to MUL0, the port to be used can be switched by MULSEL.

Multi function register

MULSEL (OFC0H)	7	6	5	4	3	2	1	0	(Initial value: *000 0000)
		MUL6	MUL5	MUL4	MUL3	MUL2	MUL1	MUL0	
	MUL6	INT3 function pin select						0: P14 1: P36	R/W
	MUL5	INT2 function pin select						0: P13 1: P35	
	MUL4	INT1 function pin select						0: P12 1: P34	
	MUL3	PPG6, PWM6, PDO6 TC6 function pin select						0: P33 1: P74	
	MUL2	PPG4, PWM4, PDO4, TC4 function pin select						0: P32 1: P73	
	MUL1	PWM3, PDO3, TC3 function pin select						0: P31 1: P72	
	MUL0	DVO function pin select						0: P30 1: P71	

Figure 2-9. Multi Function Register

2.4 Time Base Timer (TBT)

The time base timer generates time base for key scanning, dynamic displaying, etc. It also provides a time base timer interrupt (INTTBT).

An INTTBT is generated on the first falling edge of source clock (the divider output of the timing generator) after the time base timer has been enabled. The divider is not cleared by the program ; therefore, only the first interrupt may be generated ahead of the set interrupt period (Figure 2-10 (b)).

The interrupt frequency (TBTCK) must be selected with the time base timer disabled (the interrupt frequency must not be changed with the disable from the enable state). Both frequency selection and enabling can be performed simultaneously.

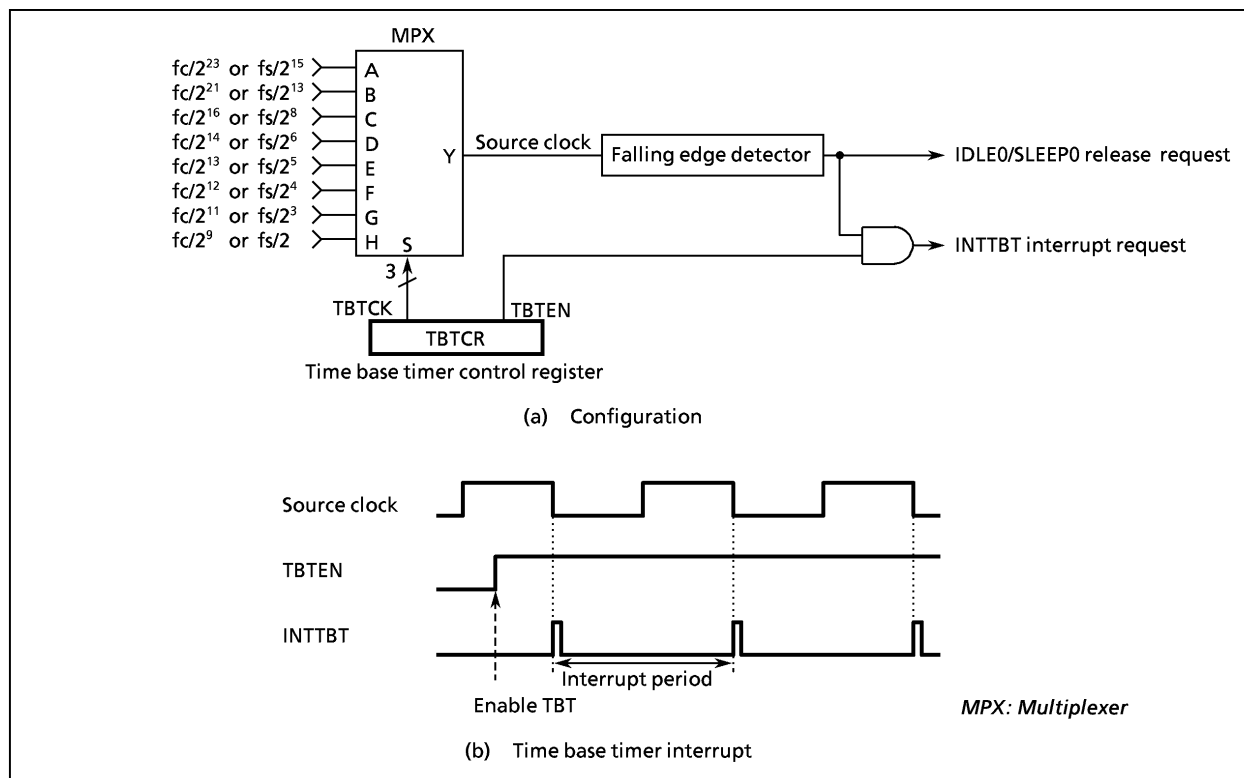


Figure 2-10. Time Base Timer

Example: Sets the time base timer frequency to $fc/2^{16}$ [Hz] and enables an INTTBT interrupt.

```
LD    (TBTCTR), 00000010B    ; TBTCK ← 010
LD    (TBTCTR), 00001010B    ; TBTEN ← 1
SET   (EIRL). 6
```

TBTCR (0036H)	7	6	5	4	3	2	1	0	(Initial value: 0**0 0***)
	(DVOEN)	(DVQCK)	(DV7CK)	TBTEN	TBTC				
	Time base timer enable/disable		0: Disable 1: Enable						
TBTEN	Time base timer interrupt frequency select [Hz]		NORMAL1/2, IDLE1/2 mode		SLOW, SLEEP mode				R/W
TBTC			DV7CK = 0		DV7CK = 1				
			000	$fc/2^{23}$	$fs/2^{15}$	$fs/2^{15}$			
			001	$fc/2^{21}$	$fs/2^{13}$	$fs/2^{13}$			
			010	$fc/2^{16}$	$fs/2^8$	-			
			011	$fc/2^{14}$	$fs/2^6$	-			
			100	$fc/2^{13}$	$fs/2^5$	-			
			101	$fc/2^{12}$	$fs/2^4$	-			
			110	$fc/2^{11}$	$fs/2^3$	-			
111	$fc/2^9$	$fs/2$	-						

Note: fc : High-frequency clock [Hz], fs : Low-frequency clock [Hz], *: Don't care

Figure 2-11. Time Base Timer Control Register

Table 2-1. Time Base Timer Interrupt Frequency (Example: $fc = 16.0$ MHz, $fs = 32.768$ kHz)

TBTC	Time base timer interrupt frequency [Hz]		
	NORMAL1/2, IDLE1/2 mode		SLOW, SLEEP mode
	DV7CK = 0	DV7CK = 1	
000	1.91	1	1
001	7.63	4	4
010	244.14	128	-
011	976.56	512	-
100	1953.13	1024	-
101	3906.25	2048	-
110	7812.5	4096	-
111	31250	16384	-

2.5 Watchdog Timer (WDT)

The watchdog timer is a fail-safe system to rapidly detect the CPU malfunctions such as endless looping caused by noise or the like, or deadlock and resume the CPU to the normal state.

The watchdog timer signal for detecting malfunction can be selected either a reset generate or a non-maskable interrupt request. However, selection is possible only once after reset. At first the reset generate is selected.

When the watchdog timer is not being used for malfunction detection, it can be used as a timer to generate an interrupt at fixed intervals.

Note: Care must be given in system design so as to protect the Watchdog Timer from disturbing noise. Otherwise the Watchdog Timer may not fully exhibit its functionality.

2.5.1 Watchdog Timer Configuration

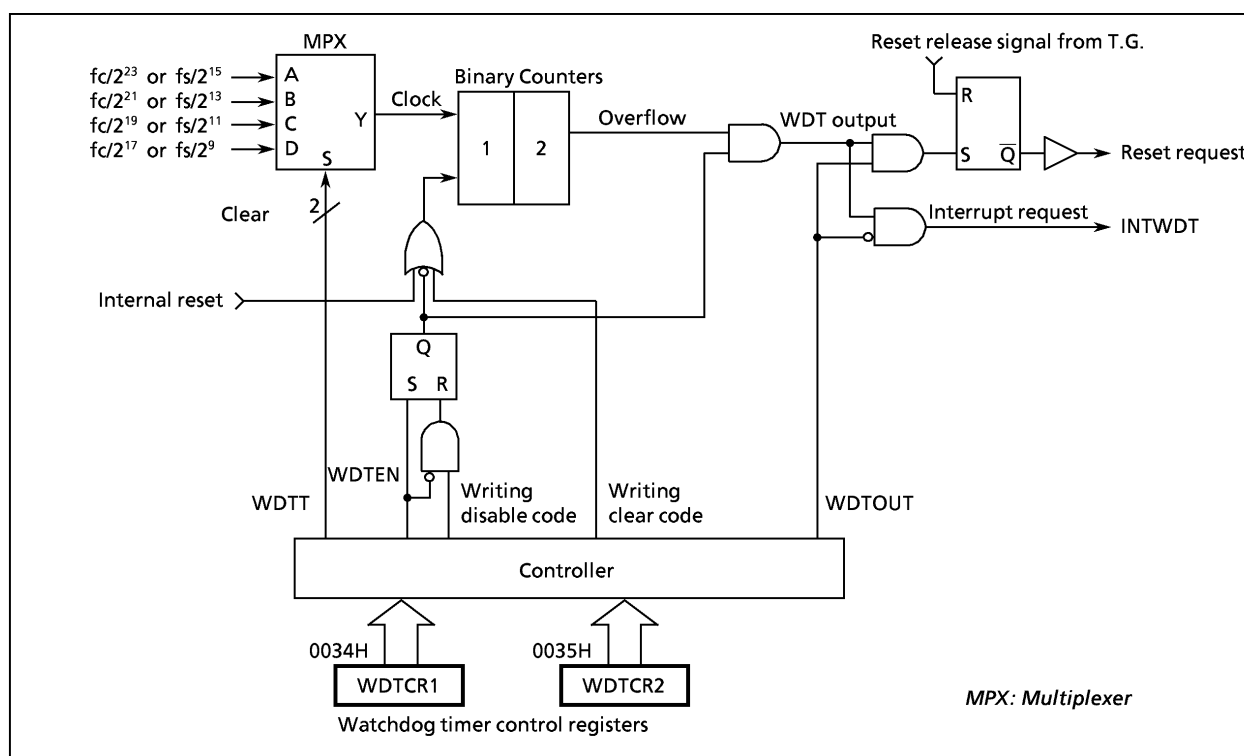


Figure 2-12. Watchdog Timer Configuration

2.5.2 Watchdog Timer Control

Figure 2-13 shows the watchdog timer control registers (WDTCR1, WDTCR2). The watchdog timer is automatically enabled after reset.

(1) Malfunction detection methods using the watchdog timer

The CPU malfunction is detected as follows.

- ① Setting the detection time, selecting output, and clearing the binary counter.
- ② Repeatedly clearing the binary counter within the setting detection time

If the CPU malfunctions such as endless looping or deadlock occur for any cause, the watchdog timer output will become active at the rising of an overflow from the binary counters unless the binary counters are cleared. At this time, when WDTOUT=1 a reset is generated, to reset the internal hardware. When WDTOUT=0, a watchdog timer interrupt (INTWDT) is generated.

The watchdog timer temporarily stops counting in STOP mode including warm-up or IDLE mode, and automatically restarts (continues counting) when the STOP/IDLE mode is released.

Note: The watchdog timer consists of an internal divider and a two-stage binary counter. When clear code 4EH is written, only the binary counter is cleared, not the internal divider. Depending on the timing at which clear code 4EH is written on the WDTCR2 register, the overflow time of the binary counter may be at minimum 3/4 of the time set in WDTCR1 <WDTT>. Thus, write the clear code using a shorter cycle than 3/4 of the time set in WDTCR1 <WDTT>.

Example: Sets the watchdog timer detection time to $2^{21}/f_c$ [s] and resets the CPU malfunction.

	LD(WDTCR2), 4EH	; Clears the binary counters
	LD(WDTCR1), 00001101B	; WDTT ← 10, WDTOUT ← 1
Within 3/4 of WDT detection time	LD(WDTCR2), 4EH	; Clears the binary counters (always clear immediately before and after changing WDTT)
Within 3/4 of WDT detection time	LD(WDTCR2), 4EH	; Clears the binary counters
	LD(WDTCR2), 4EH	; Clears the binary counters

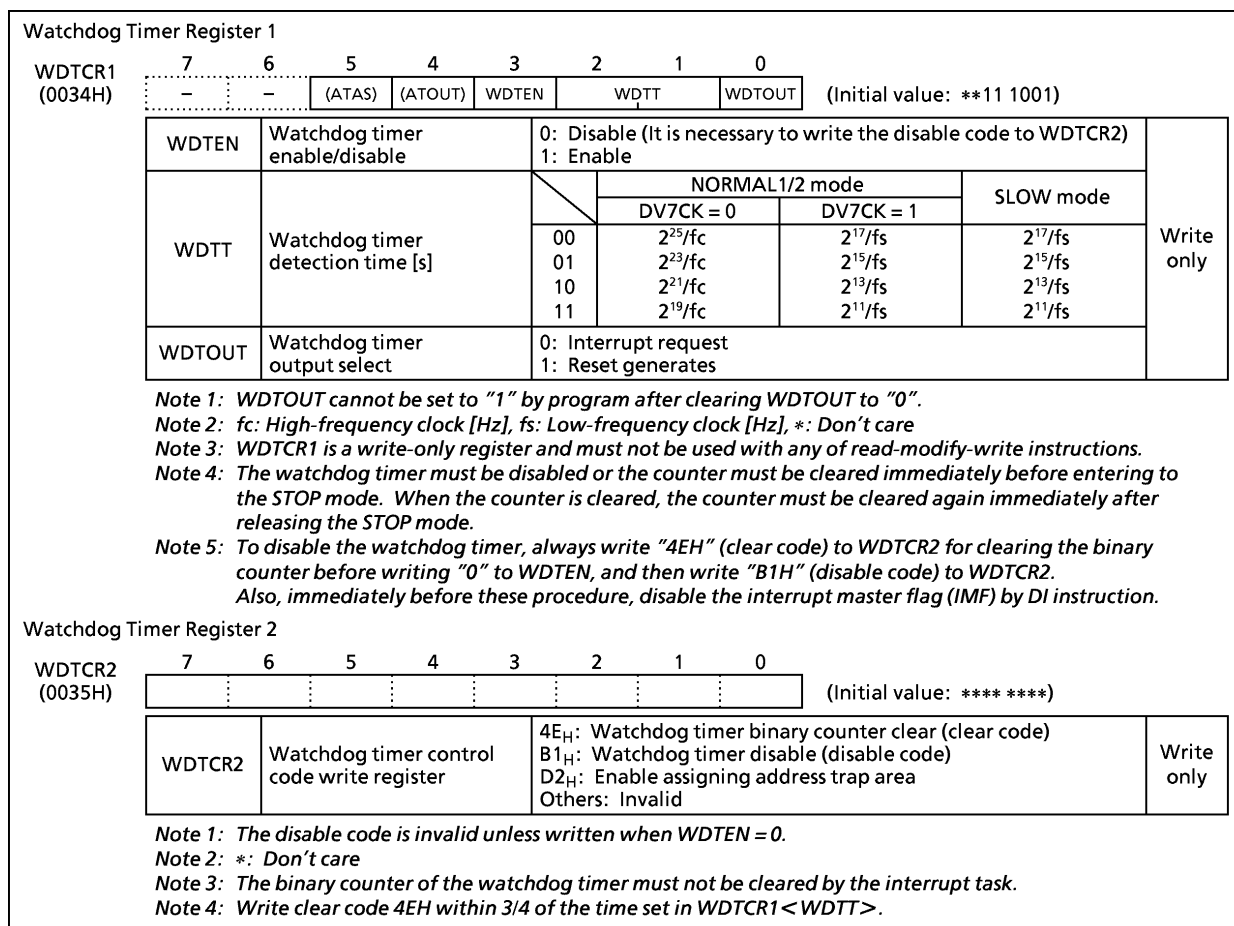


Figure 2-13. Watchdog Timer Control Registers

(2) Watchdog timer enable

The watchdog timer is enabled by setting WDTEN (bit 3 in WDTCR1) to "1". WDTEN is initialized to "1" during reset, so the watchdog timer operates immediately after reset is released.

(3) Watchdog timer disable

To disable the watchdog timer, write "4EH" (clear code) to WDTCR2 for clearing the binary counter before writing "0" to WDTEN, and then write "B1H" (disable code) to WDTCR2. The watchdog timer is not disabled if this procedure is reversed and the disable code is written to WDTCR2 before WDTEN is cleared to "0". Also, immediately before these procedure, disable the interrupt master flag (IMF) by DI instruction. During disabling the watchdog timer, the binary counters are cleared to "0".

Example: Disables watchdog timer

```
DI ; IMF ← 0
LD (WDTCR2), 04EH ; Clear the binary counter
LDW (WDTCR1), 0B101H ; WDTEN ← 0, WDTCR2 ← Disable code
EI ; IMF ← 1
```

Table 2-2. Watchdog Timer Detection Time (Example: fc = 16.0 MHz, fs = 32.768 kHz)

WDTT	Watchdog timer detection time [s]		
	NORMAL1/2 mode		SLOW mode
	DV7CK = 0	DV7CK = 1	
00	2.097	4	4
01	524.288 m	1	1
10	131.072 m	250 m	250 m
11	32.768 m	62.5 m	62.5 m

2.5.3 Watchdog Timer Interrupt (INTWDT)

This is a pseudo non-maskable interrupt which can be accepted regardless of the contents of the EIR. If a watchdog timer interrupt or a software interrupt is already accepted, however, the new watchdog timer interrupt waits until the previous interrupt processing is completed (the end of the [RETN] instruction execution).

The stack pointer (SP) should be initialized before using the watchdog timer output as an interrupt source with WDTOUT.

Example: Watchdog timer interrupt setting up

```
LD SP, 023FH          ; Sets the stack pointer
LD (WDTCR1), 00001000B ; WDTOUT ← 0
```

2.5.4 Watchdog Timer Reset

If the watchdog timer reset request occurs, a reset is generated, to reset the internal hardware. The reset time is about $8/f_c$ to $24/f_c$ [s] (0.5 to $1.5 \mu\text{s}$ at $f_c = 16.0 \text{ MHz}$).

Note: The high-frequency clock oscillator also turns on when a watchdog timer reset is generated in SLOW mode. The reset time is $8/f_c$ to $24/f_c$ [s]. Therefore, the reset time may include a certain amount of error if there is any fluctuation of the oscillation frequency at starting the high-frequency clock oscillation. Thus, the reset time must be considered an approximated value.

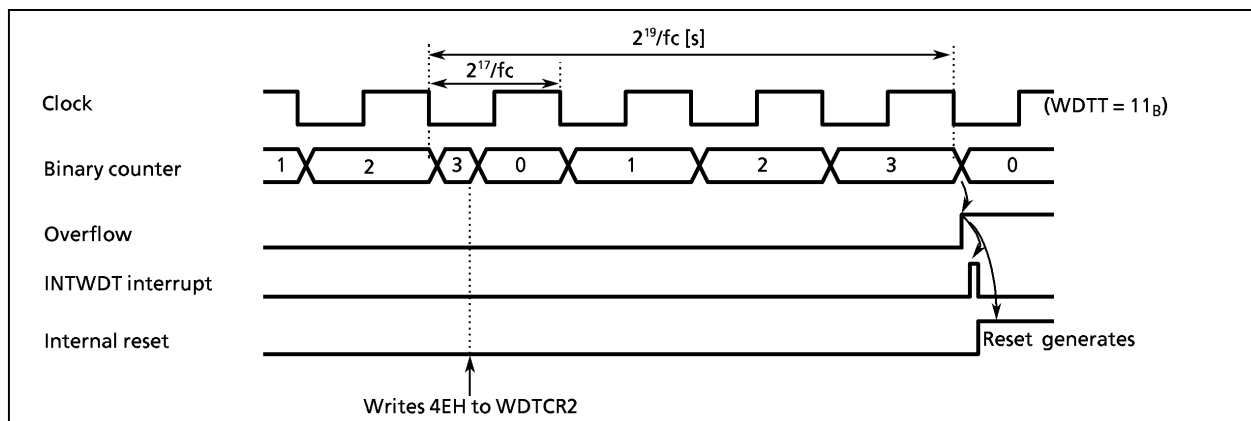


Figure 2-14. Watchdog Timer Interrupt/Reset

2.5.5 Address Trap

The Watchdog Timer Control Register 1, 2 shares its addresses with the control registers in case of address trap. These control registers for address trap are shown on Figure 2-15.

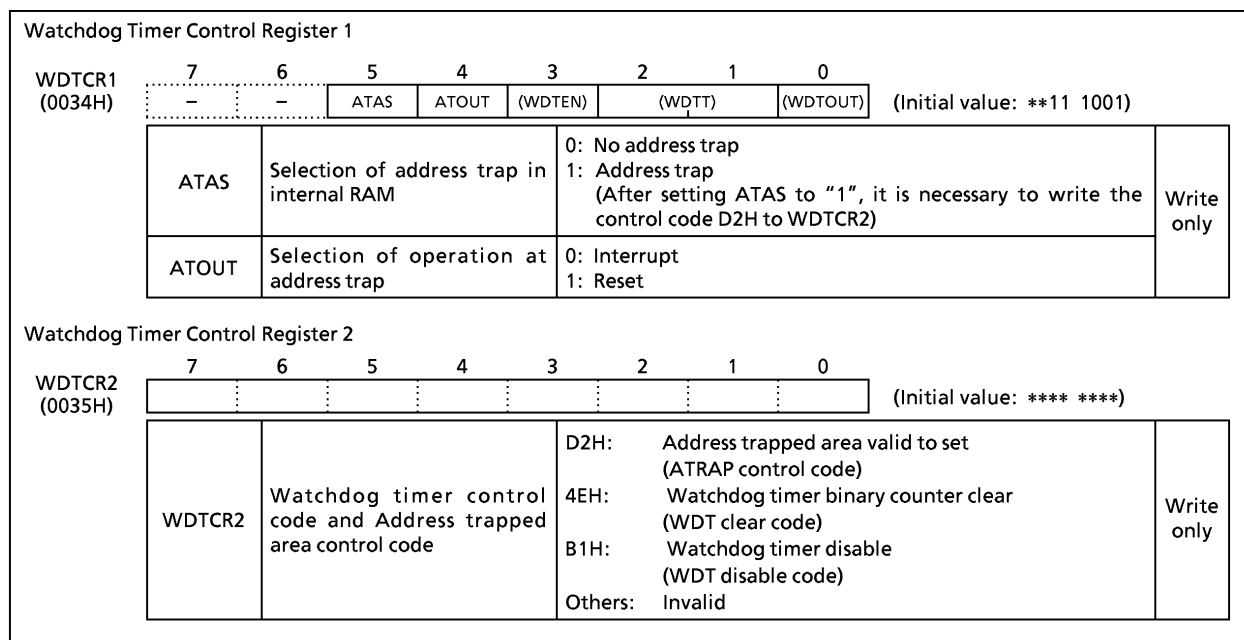


Figure 2-15. Watchdog Timer Control Registers

(1) Selection of address trap in internal RAM (ATAS)

Using WDTCR1<ATAS>, address trap or no address trap can be selected for the internal RAM area. To execute an instruction in the internal RAM area, set "0" in WDTCR1<ATAS>. Setting in WDTCR1<ATAS> becomes valid after control code D2H is written in WDTCR2. Executing an instruction in the SFR/DBR area generates an address trap unconditionally regardless of the setting in WDTCR1<ATAS>.

(2) Selection of operation at address trap (ATOUT)

As the operation at address trap either interrupt generation or reset generation can be selected by WDTCR1<ATOUT>.

2.6 Divider Output (DVO)

Approximately 50% duty pulse can be output using the divider output circuit, which is useful for piezoelectric buzzer drive. Divider output (\overline{DVO}) is output from P30 or P71. The selection of P30 or P71 is controlled by MULSEL<MUL0>. To output \overline{DVO} , the corresponding bit of output latch (P3DR or P7DR) should be set to “1”.

Note: Selection of divider output frequency must be made while divider output is disabled. Also, in other words, when changing the state of the divider output frequency from enabled to disabled, do not change the setting of the divider output frequency.

TBTCR (0036H)	7	6	5	4	3	2	1	0	(Initial value: 0000 0000)
	DVOEN	DVOCK		(DV7CK)	(TBTEN)	(TBTCK)			
	DVOEN	Divider output enable/ disable		0: Disable 1: Enable					
DVOCK	Divider output (\overline{DVO}) frequency selection [Hz]		NORMAL1/2 mode			SLOW, SLEEP mode			
			DV7CK = 0			DV7CK = 1			
			00	$fc/2^{13}$	$fs/2^5$	$fs/2^5$			
			01	$fc/2^{12}$	$fs/2^4$	$fs/2^4$			
			10	$fc/2^{11}$	$fs/2^3$	$fs/2^3$			
11	$fc/2^{10}$	$fs/2^2$	$fs/2^2$						

*Note: fc: High-frequency clock [Hz], fs: Low-frequency clock [Hz], *: Don't care*

Figure 2-16. Divider Output Control Register

Example: 1.95 kHz pulse output (at $fc = 16$ MHz) from P30

```
LD (MULSEL), 00000000B ; DVO output from P30
SET (P3DR), 0 ; P30 output latch ← "1"
LD (TBTCR), 00000000B ; DVOCK ← "00"
LD (TBTCR), 10000000B ; DVOEN ← "1"
```

Table 2-3. Divider Output Frequency (Example: at $fc = 16.0$ MHz, $fs = 32.768$ kHz)

DVOCK	Divider output frequency [Hz]		
	NORMAL1/2, IDLE1/2 mode		SLOW, SLEEP mode
	DV7CK = 0	DV7CK = 1	
00	1.953 k	1.024 k	1.024 k
01	3.906 k	2.048 k	2.048 k
10	7.813 k	4.096 k	4.096 k
11	15.625 k	8.192 k	8.192 k

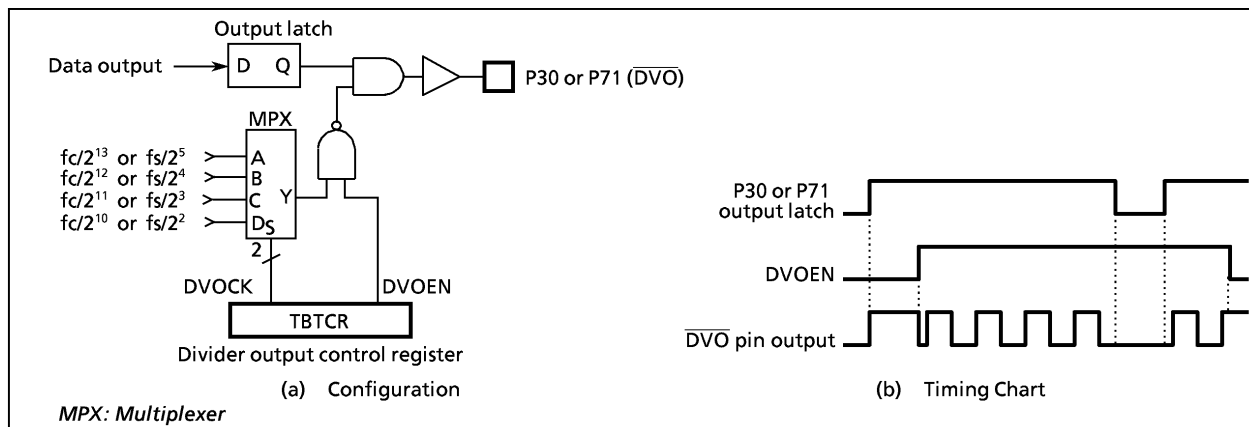


Figure 2-17. Divider Output

2.7 18-Bit Timer/Counter (TC1)

2.7.1 Configuration

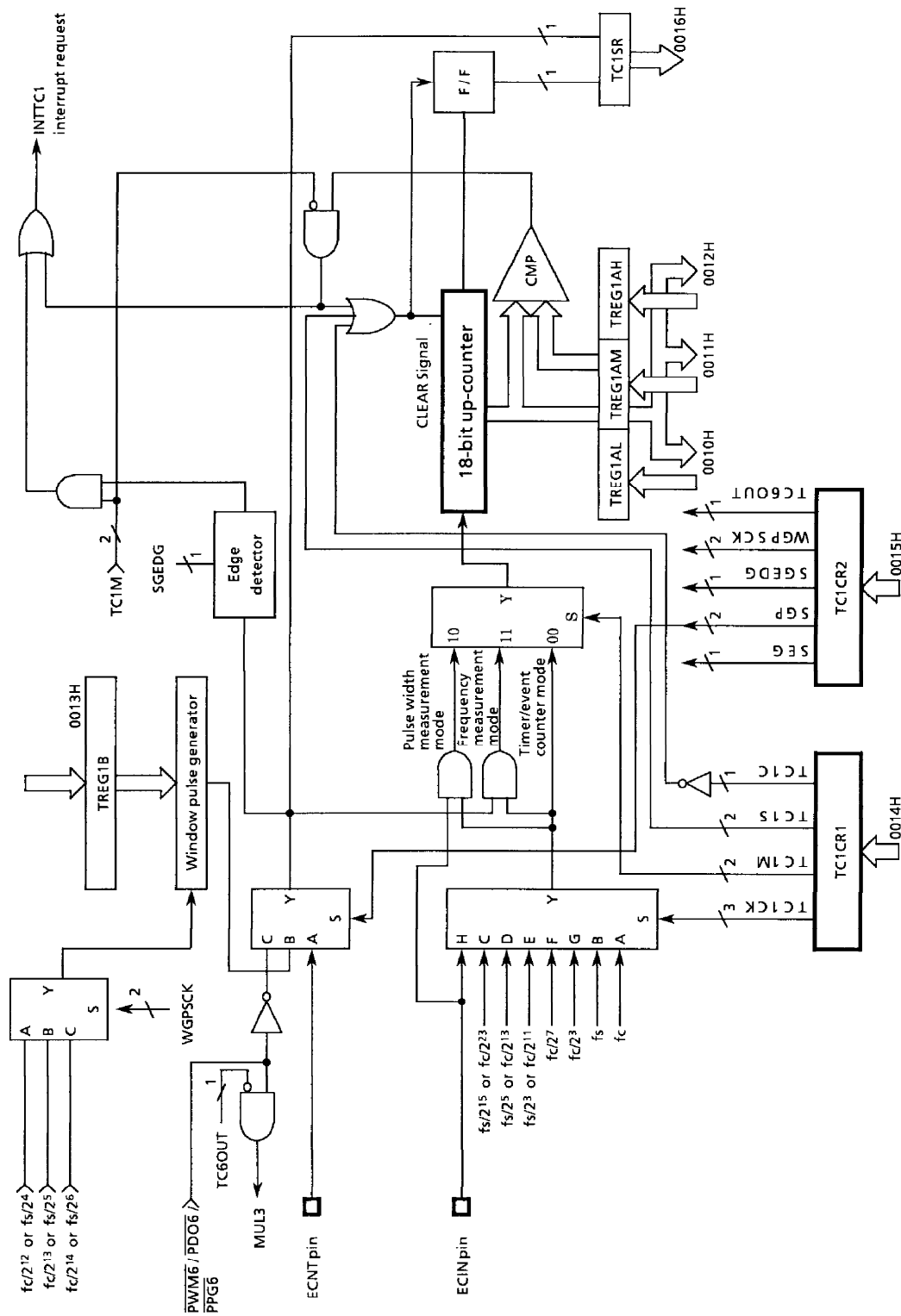


Figure 2-18. Timer/Counter 1

2.7.2 Control

The Timer/Counter 1 is controlled by a timer/counter 1 control registers (TC1CR1/TC1CR2), an 18-bit timer register (TREG1A), and an 8-bit internal window gate pulse setting register (TREG1B).

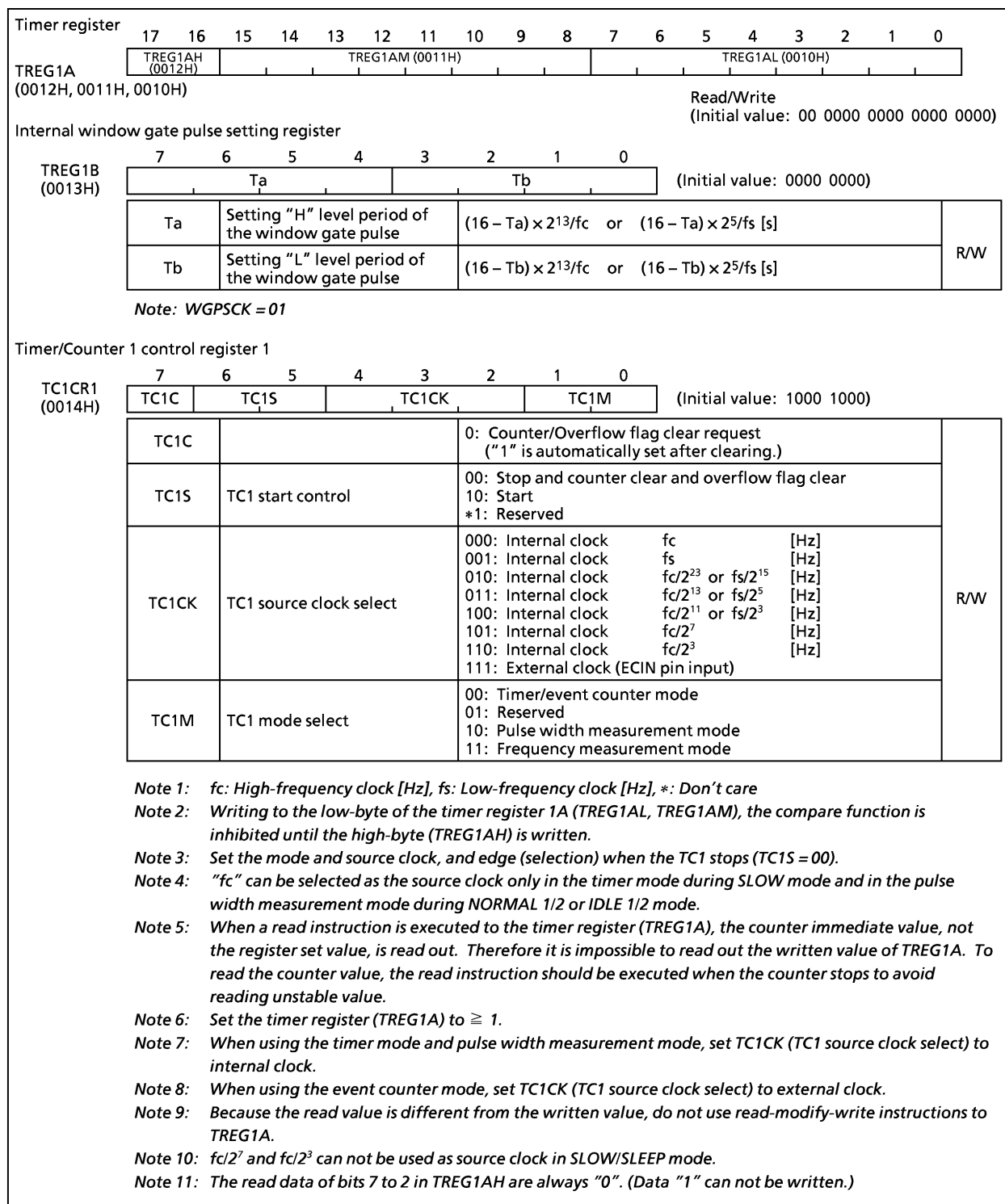


Figure 2-19. Timer Register/Window Gate Pulse Setting Register/Control Register of the TC1

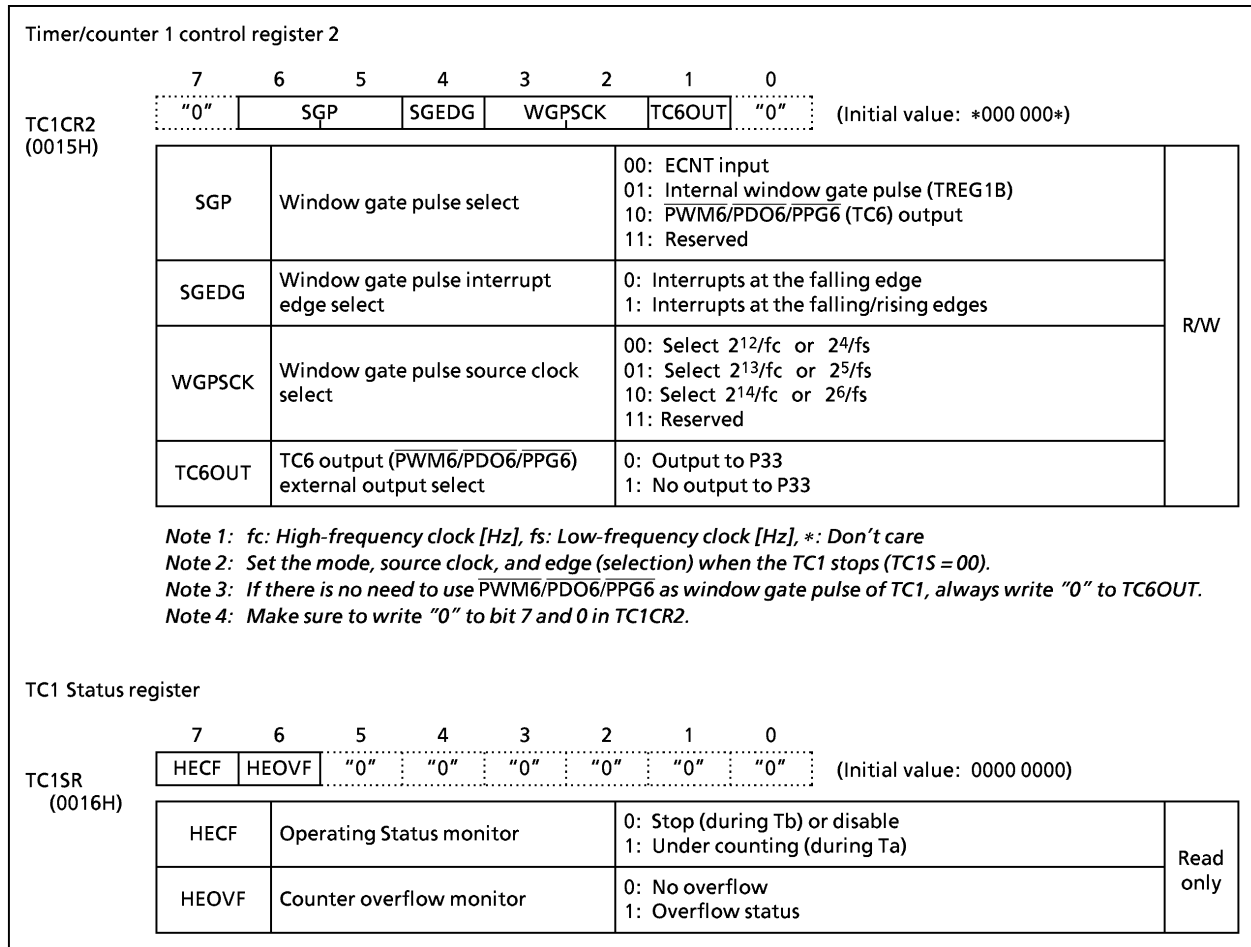


Figure 2-20. Control Register of the TC1/Status Register

2.7.3 Function

TC1 has four operating modes. The timer mode of the TC1 is used at warm-up when switching from SLOW mode to NORMAL2 mode.

(1) Timer mode

In this mode, counting up is performed using the internal clock. The contents of TREG1A are compared with the contents of up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared. Counting up resumes after the counter is cleared.

Table 2-4. Source Clock (internal clock) of Timer/Counter 1

Source clock				Resolution		Maximum time setting	
NORMAL1/2, IDLE1/2 mode		SLOW mode	SLEEP mode	$f_c = 16 \text{ MHz}$	$f_s = 32.768 \text{ kHz}$	$f_c = 16 \text{ MHz}$	$f_s = 32.768 \text{ kHz}$
DV7CK = 0	DV7CK = 1						
$f_c/2^{23} [\text{Hz}]$	$f_s/2^{15} [\text{Hz}]$	$f_s/2^{15} [\text{Hz}]$	$f_s/2^{15} [\text{Hz}]$	0.52 s	1 s	38.2 h	72.8 h
$f_c/2^{13}$	$f_s/2^5$	$f_s/2^5$	$f_s/2^5$	512 μs	0.98 ms	2.2 min	4.3 min
$f_c/2^{11}$	$f_s/2^3$	$f_s/2^3$	$f_s/2^3$	128 μs	244 μs	0.6 min	1.07 min
$f_c/2^7$	$f_c/2^7$	—	—	8 μs	—	2.1 s	—
$f_c/2^3$	$f_c/2^3$	—	—	0.5 μs	—	131.1 ms	—
f_c	f_c	f_c (Note)	—	62.5 ns	—	16.4 ms	—
f_s	f_s	—	—	—	30.5 μs	—	8 s

Note: When f_c is selected for the source clock in SLOW mode, the lower bits 11 of TREG1A is invalid, and a match of the upper bits 7 makes interrupts.

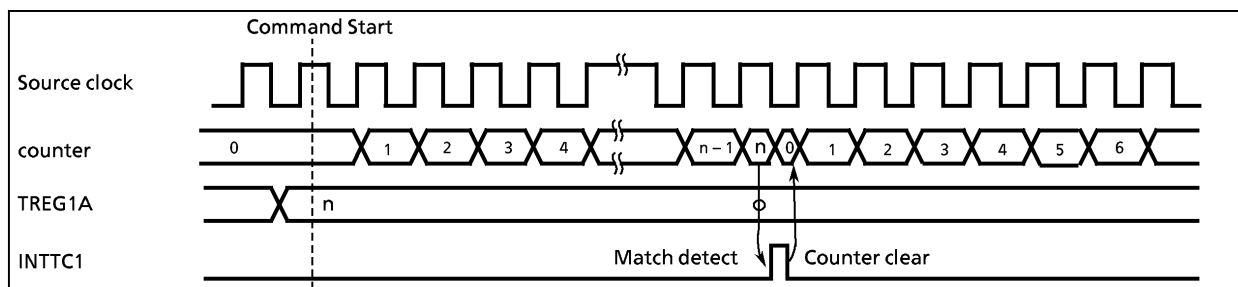


Figure 2-21. Timing Chart for Timer Mode

(2) Event Counter mode

It is a mode to count up at the falling edge of the ECIN pin input. The countents of TREG1A are compared with the contents of up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared. Counting up resumes for ECIN pin input edge each after the counter is cleared. The maximum applied frequency is $f_c/2^4$ [Hz] in NORMAL 1/2 or IDLE 1/2 mode and $f_s/2^4$ [Hz] in SLOW or SLEEP mode. Two or more machine cycles are required for both the “H” and “L” levels of the pulse width.

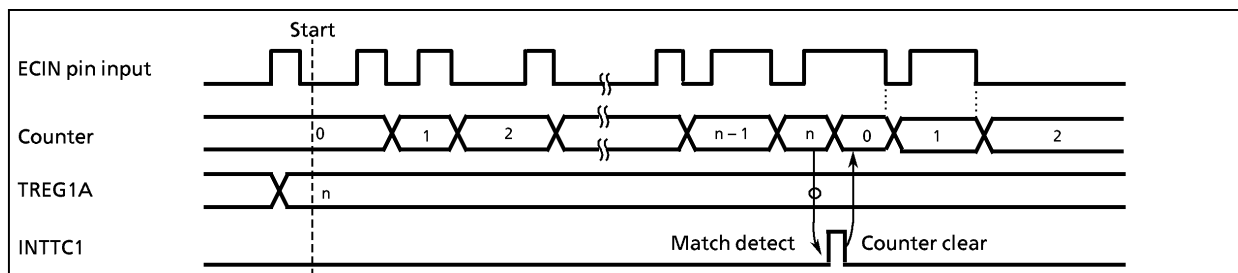


Figure 2-22. Event Counter Mode Timing Chart

(3) Pulse Width Measurement mode

In this mode, pulse widths are counted on the falling edge of logical AND-ed product between ECIN pin input (window pulse) and the internal clock. The internal clock is selected by TC1CK (bit 2, 3 and 4 in TC1CR1). An INTTC1 interrupt is generated at the falling edge of the window pulse or both rising and falling edges of the window pulse, that can be selected by SGEDG (bit 4 in TC1CR2). In the interrupt service program, read the contents of TREG1A while the count is stopped (ECIN pin is low), then clear the counter using TC1C (bit 7 in TC1CR1). When the counter is not cleared, counting up resumes by starting count-up. When TREG1A is counted up from 3FFFF_H to 00000_H, an overflow occurs. HEOVF (bit 6 in TC1SR) of the status register can monitor whether the overflows or not. HEOVF remains the old data until the counter is required to be cleared by TC1C.

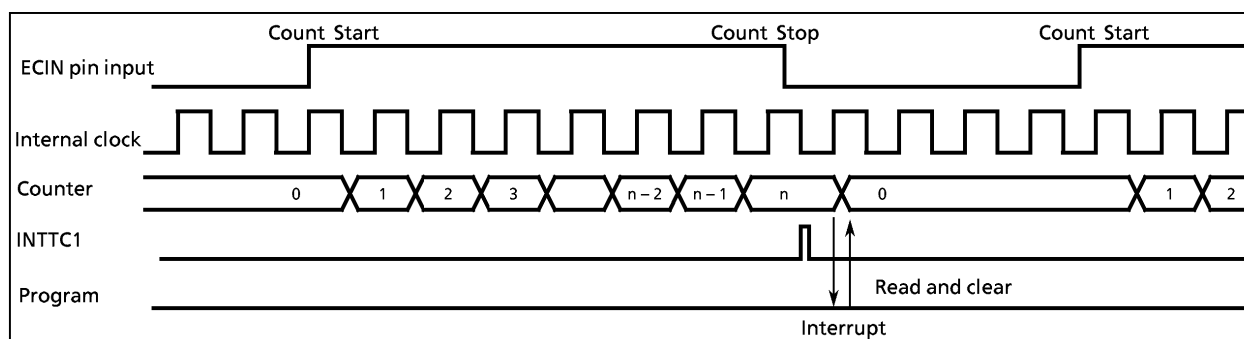


Figure 2-23. Pulse Width Measurement Mode Timing Chart (TC1CR2<SGEDG> = "0")

Note 1: INTTC1 interrupt occurs when ECIN input is "1" and TC1S of TC1CR1 is written to "00". According to the following step, when timer counter is stopped, INTTC1 interrupt latch should be cleared to "0".

```
TC1STOP:    ;
            DI                ; Clear IMF
            CLR (EIRH). EF8    ; Clear EF8
            LD (TC1CR1), 0y00011010 ; Stop timer counter 1
            LD (ILH), 0y11111110    ; Clear IL8
            SET (EIRH). EF8        ; SET EF8
            EI                    ; SET IMF
```

Note 2: When SGEDG (window gate pulse interrupt edge select) is set to both edges and ECIN pin input is "1" in the pulse width measurement mode, an INTTC1 interrupt is generated by setting TC1S (TC1 start control) to "10" (start).

Note 3: In the pulse width measurement mode, HECF (operating status monitor) cannot be used.

(4) Frequency Measurement mode

In this mode, the frequency of ECIN pin input pulse is measured. TC1CK is required to be set to the external clock (TC1CK = "111"). The edge of the input pulse is counted during "H" level of the window gate pulse selected by SGP (bit 5 and 6 in TC1CR2). Whether the input pulse is counted on the falling edge. An INTTC1 interrupt is generated on the falling edge or both the rising/falling edges of the window gate pulse, that can be selected by SGEDG (bit 4 in TC1CR2). To use ECNT terminal input as a window gate pulse, SGP (bit 5 and 6 in TC1CR2) should be set to "00". In the interrupt service program, read the contents of TREG1A while the count is stopped (window gate pulse is low), then clear the counter using TC1C. When the counter is not cleared, counting up resumes by stating count-up. The window pulse status can be monitored by HECF of the status register. HEOVF of the status register can monitor whether the binary counter overflows or not. In the overflow flag status, a new data is not input until the counter clear requests.

- Using TC6 output (PWM6/PDO6/PPG6) for the window gate pulse, external output of PWM6/PDO6/PPG6 to P33 can be controlled using TC6OUT (bit 1 in TC1CR2). Zero-clearing TC6OUT outputs PWM6/PDO6/PPG6 to P33; setting 1 in TC6OUT does not output PWM6/PDO6/PPG6 to P33. (TC6OUT is used to control output to P33 only. Thus, use the timer counter 6 control register to operate/stop PWM6/PDO6/PPG6.)
- When the internal window gate pulse is selected, the window gate pulse is set as follows. The internal window gate pulse consists of "H" level period (Ta) that is counting time and "L" level period (Tb) that is counting stop time. Ta or Tb can be individually set by TREG1B. One cycle contains Ta + Tb.

Note 1: Because the internal window gate pulse is generated in synchronization with the internal divider, it may be delayed for a maximum of one cycle of the source clock (WGPSCK) immediately after start of the timer.

Note 2: Set the internal window gate pulse when the timer counter is not operating or during the Tb period. When Tb is overwritten during the Tb period, the update is valid from the next Tb period.

Tabale 2-5. Setting Ta and Tb
(WGPSCK = 10, fs = 32.768 kHz)

Setting value	Setting time	Setting value	Setting time
0	31.25 ms	8	15.63 ms
1	29.30 ms	9	13.67 ms
2	27.34 ms	A	11.72 ms
3	25.39 ms	B	9.77 ms
4	23.44 ms	C	7.81 ms
5	21.48 ms	D	5.86 ms
6	19.53 ms	E	3.91 ms
7	17.58 ms	F	1.95 ms

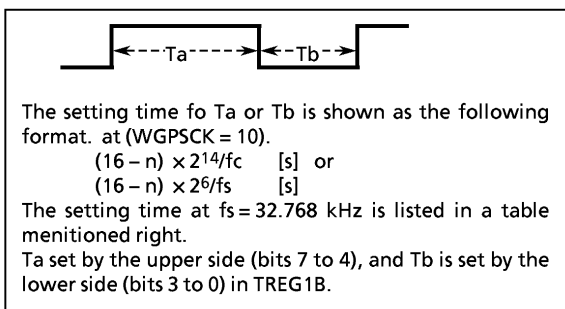


Figure 2-24. Window Gate Pulse Format

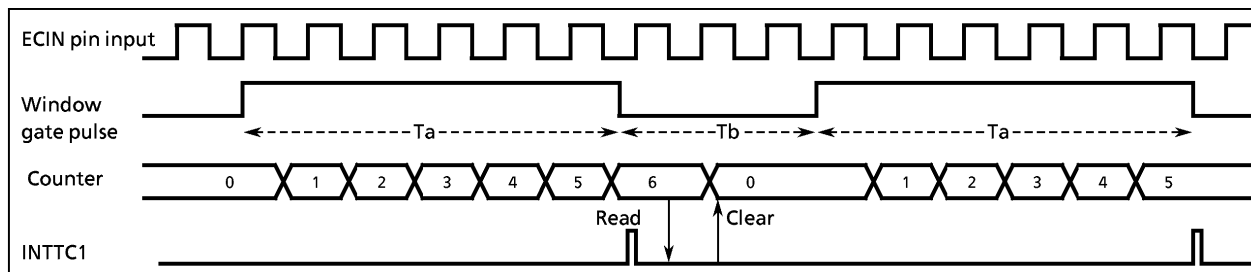


Figure 2-25. Timing Chart for the Frequency Measurement Mode
(window gate pulse falling interrupt at TC1CR2 <SGEDG> = "0")

2.8 8-Bit Timer/Counter (TC3, 4, 5, 6)

The TMP86CM25/S25 have four channels of 8-bit timer/counter (TC3, 4, 5, 6). These timer/counter are used as timer, event counter, PWM, PPG and PDO. These are also available as a 16-bit timer/counter by cascade connection.

2.8.1 Configuration

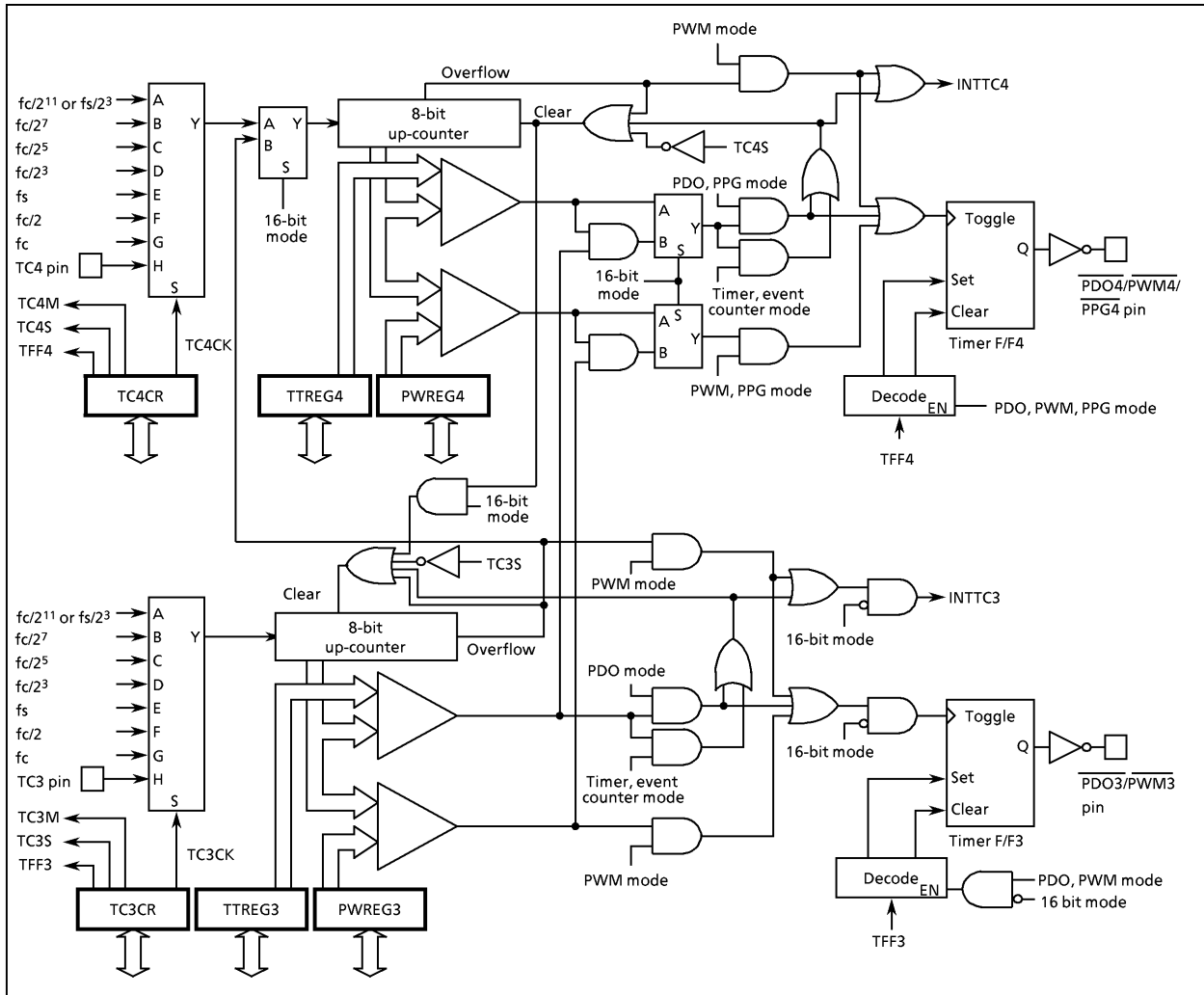


Figure 2-26. 8-Bit Timer 3, 4

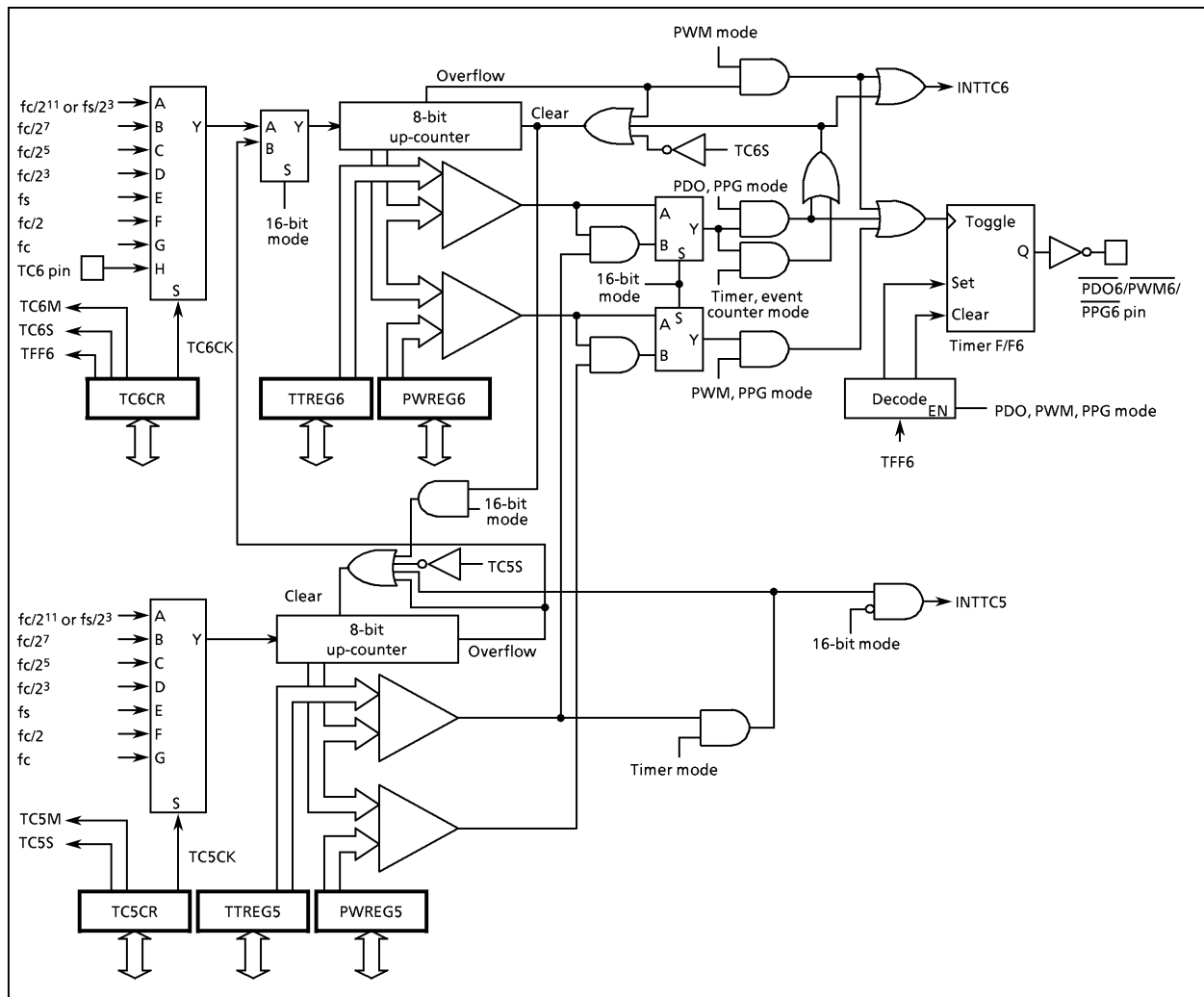


Figure 2-27. 8-Bit Timer 5, 6

2.8.2 Control

The timer/counter 3 is controlled by a timer/counter 3 control register (TC3CR) and two 8-bit timer registers (TTREG3 and PWREG3).

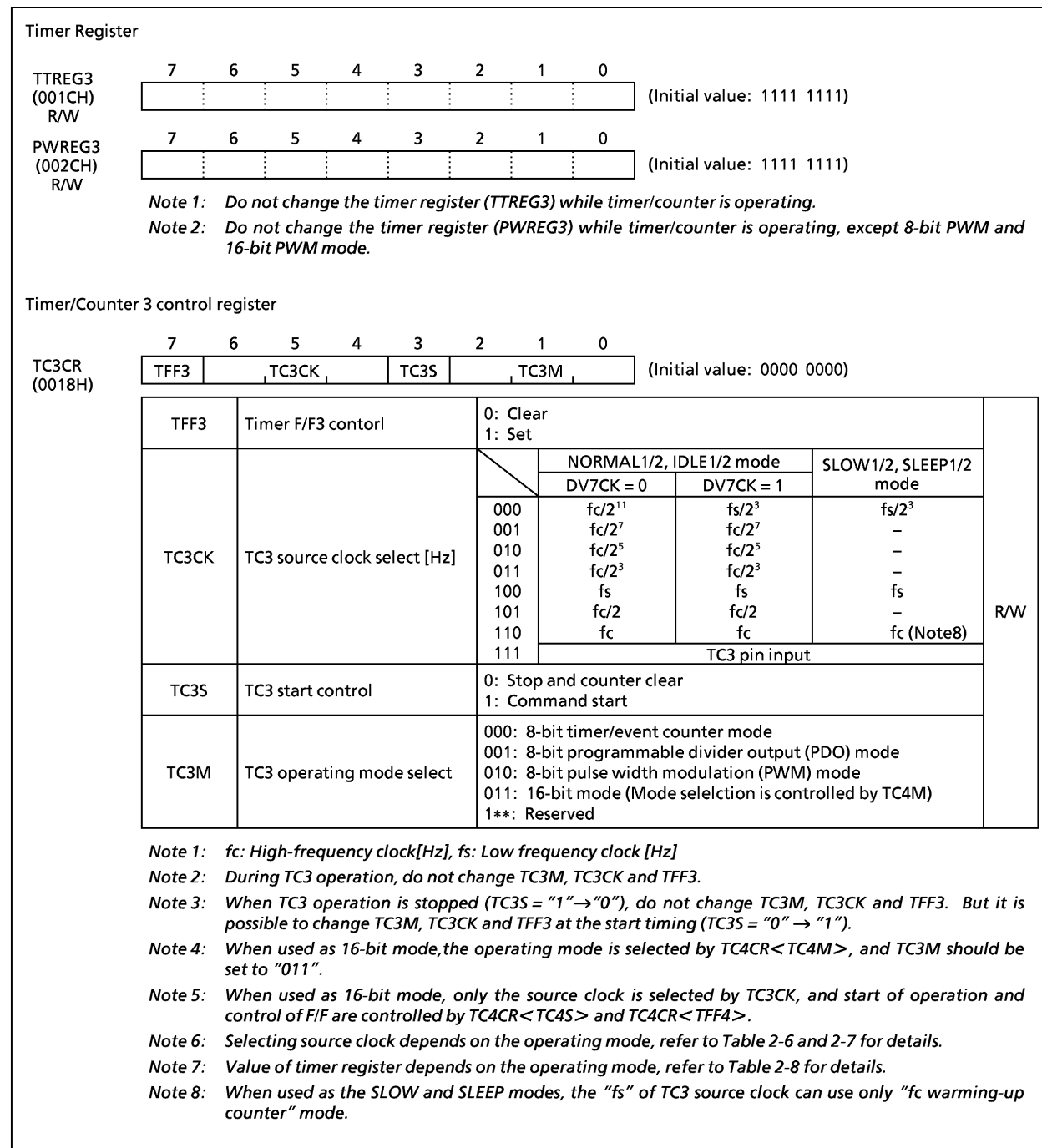


Figure 2-28. Timer 3 Register and Timer/Counter 3 Control Register

The timer/counter 4 is controlled by a timer/counter 4 control register (TC4CR) and two 8-bit timer registers (TTREG4 and PWREG4).

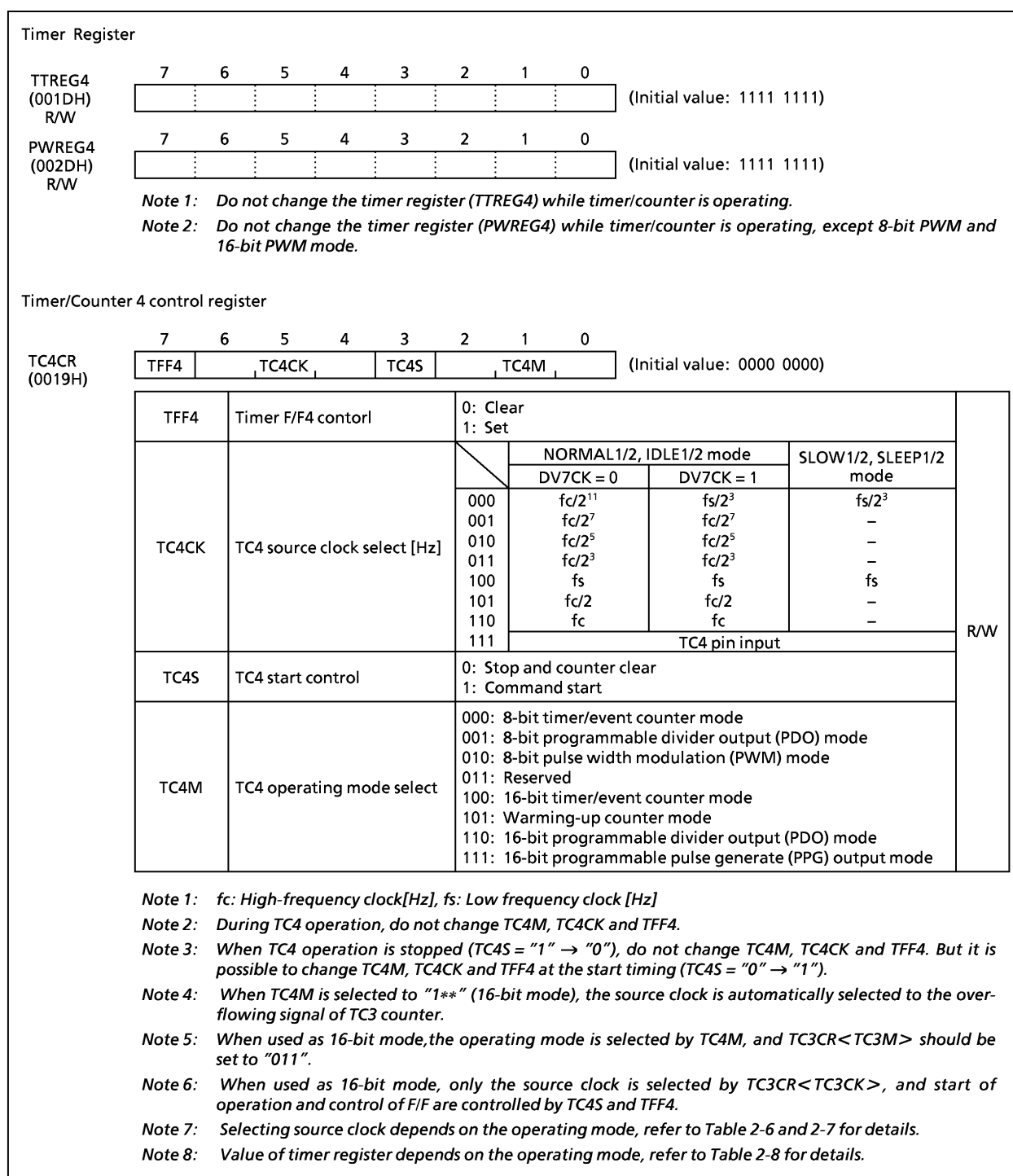


Figure 2-29. Timer 4 Register and Timer/Counter 4 Control Register

The timer/counter 5 is controlled by a timer/counter 5 control register (TC5CR) and two 8-bit timer registers (TTREG5 and PWREG5).

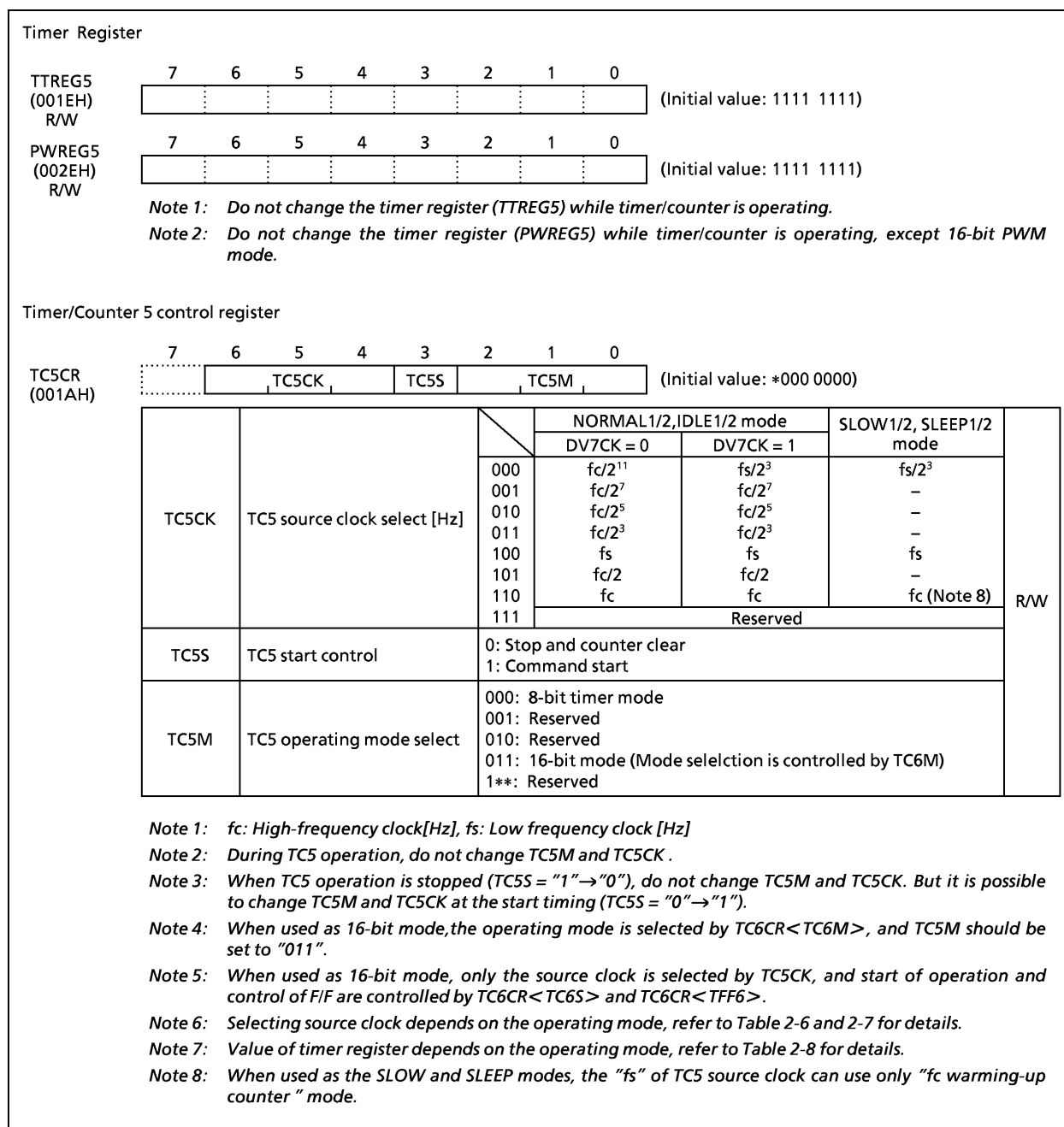


Figure 2-30. Timer 5 Register and Timer/Counter 5 Control Register

The timer/counter 6 is controlled by a timer/counter 6 control register (TC6CR) and two 8-bit timer registers (TTREG6 and PWREG6).

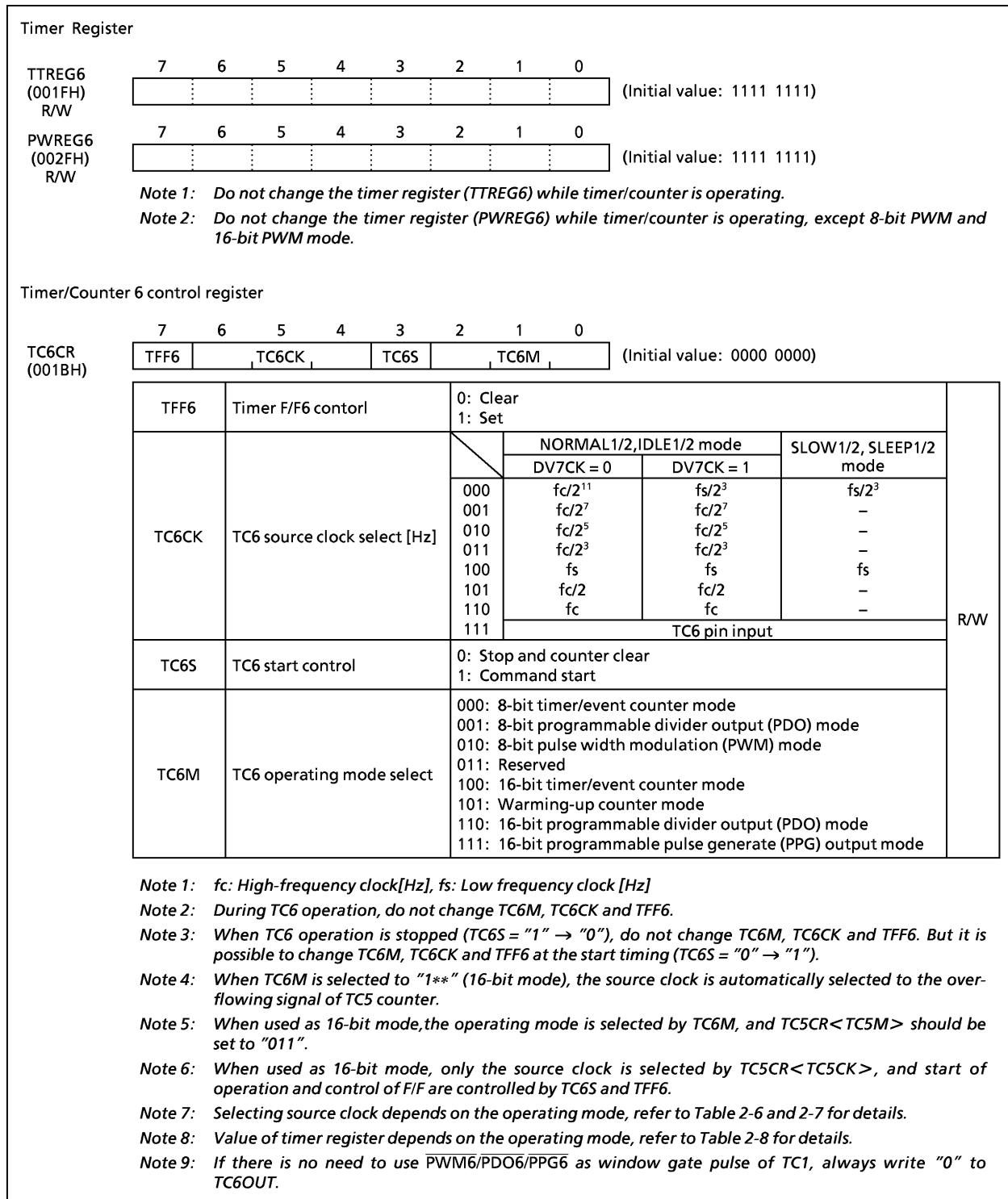


Figure 2-31. Timer 6 Register and Timer/Counter 6 Control Register

Table 2-6. Operating mode and available source clock (NORMAL1/2, IDLE1/2 mode)

Operating Mode	$fc/2^{11}$ or $fs/2^3$	$fc/2^7$	$fc/2^5$	$fc/2^3$	fs	$fc/2$	fc	TCi pin input
8-Bit Timer	○	○	○	○	—	—	—	—
8-Bit Event Counter	—	—	—	—	—	—	—	○
8-Bit PDO	○	○	○	○	—	—	—	—
8-Bit PWM	○	○	○	○	○	○	○	—
16-Bit Timer	○	○	○	○	—	—	—	—
16-Bit Event Counter	—	—	—	—	—	—	—	○
Warming-up Counter	—	—	—	—	○	—	—	—
16-Bit PWM	○	○	○	○	○	○	○	—
16-Bit PPG	○	○	○	○	—	—	—	—

Note 1: For 16-bit operation (16-Bit Timer/Event Counter, Warming-up Counter, 16-Bit PWM and 16-Bit PPG), set its source clock on lower bits (TC3CK, TC5CK).

Note 2: $i = 3, 4, 6$, (8-bit mode)
 $i = 3$ (16-bit mode)

Table 2-7. Operating mode and available source clock (Under SLOW1/2 mode, SLEEP1/2 mode)

Operating Mode	$fc/2^{11}$ or $fs/2^3$	$fc/2^7$	$fc/2^5$	$fc/2^3$	fs	$fc/2$	fc	TCi pin input
8-Bit Timer	○	—	—	—	—	—	—	—
8-Bit Event Counter	—	—	—	—	—	—	—	○
8-Bit PDO	○	—	—	—	—	—	—	—
8-Bit PWM	○	—	—	—	○	—	—	—
16-Bit Timer	○	—	—	—	—	—	—	—
16-Bit Event Counter	—	—	—	—	—	—	—	○
Warming-up Counter	—	—	—	—	—	—	○	—
16-Bit PWM	○	—	—	—	○	—	—	—
16-Bit PPG	○	—	—	—	—	—	—	—

Note 1: For 16-bit operation (16-Bit Timer/Event Counter, Warming-up Counter, 16-Bit PWM and 16-Bit PPG), set its source clock on lower bits (TC3CK, TC5CK).

Note 2: $i = 3, 4, 6$, (8-bit mode)
 $i = 3$ (16-bit mode)

Table 2-8. Restriction against the Rate for Comparing Registers

Operating Mode	Authorized Rate for Register
8-Bit Timer/Event Counter	$1 \leq (TTREGn) \leq 255$
8-Bit PDO	$1 \leq (TTREGn) \leq 255$
8-Bit PWM	$2 \leq (PWREGn) \leq 254$
16-Bit Timer/Event Counter	$1 \leq (TTREG4, 3) \leq 65535, 1 \leq (TTREG6, 5) \leq 65535$
fc Warming-up Counter	$256 \leq (TTREG4, 3) \leq 65535, 256 \leq (TTREG6, 5) \leq 65535$
16-Bit PWM	$2 \leq (PWREG4, 3) \leq 65534, 2 \leq (PWREG6, 5) \leq 65534$
16-Bit PPG	$1 \leq (PWREG4, 3) < (TTREG4, 3) \leq 65535$ and $(PWREG4, 3) + 1 < (TTREG4, 3)$ $1 \leq (PWREG6, 5) < (TTREG6, 5) \leq 65535$ and $(PWREG6, 5) + 1 < (TTREG6, 5)$

Note: $n = 3$ to 6

2.8.3 Function

Timer/Counter 3, 4, 5 and 6 have eight operating modes: 8-bit timer, 8-bit external trigger timer, 8-bit programmable divider output mode, 8-bit pulse width modulation output mode, 16-bit timer, 16-bit external trigger timer, 16-bit pulse width modulation output mode, 16-bit programmable pulse generator output mode.

16-bit timer mode can use Timer counter 3 and 4 (5, 6) by cascade connection.

(1) 8-bit timer mode (Timer/counter 3, 4, 5 and 6)

In this mode, counting up is performed using the internal clock. The contents of TTREGi are compared with the contents of up-counter. If a match is found, an INTTCi interrupt is generated, and the counter is cleared to "0". Counting up resumes after the counter is cleared.

Note 1: In the timer mode, always write TCjCR < TFFj > to "0". If TFFj is set to "1", unexpected pulse may be output from PDOj/PWMj/PPGj pin.

Note 2: In the timer mode, do not change the setting of timer registers (TTREGi) while timer/counter is operating. Since TTREGi is configured as one-stage register, a newly set value is immediately reflected on the timer register.

Note 3: j = 3, 4, 6 i = 3 to 6

Table 2-9. Timer/Counter 1 Source Clock (Internal Clock)

Source clock		SLOW1/2, SLEEP1/2 modes	Resolution		Maximum time setting	
NORMAL1/2, IDLE1/2 modes	fs/2 ³ [Hz]		At fc = 16 MHz	At fs = 32.768 kHz	At fc = 16 MHz	At fs = 32.768 kHz
DV7CK = 0		DV7CK = 1	At fc = 16 MHz	At fs = 32.768 kHz	At fc = 16 MHz	At fs = 32.768 kHz
fc/2 ¹¹ [Hz]	fs/2 ³ [Hz]	fs/2 ³ [Hz]	128 [μs]	244.14 [μs]	32.6 [ms]	62.3 [ms]
fc/2 ⁷	fc/2 ⁷	-	8 [μs]	-	2.0 [ms]	-
fc/2 ⁵	fc/2 ⁵	-	2 [μs]	-	510 [μs]	-
fc/2 ³	fc/2 ³	-	500 [ns]	-	127.5 [μs]	-

Example: Sets the timer mode with source clock fc/2⁷ [Hz] and generates an interrupt 80 μs later (at fc = 16 MHz).

```
LDW (TTREG4), 0AH ; Sets the timer register (80 μs ÷ 27/fc = 0AH)
DI
SET (EIRH).EF11 ; Enables INTTC4 interrupt
EI
LD (TC4CR), 00010000B ; Sets the 8-bit timer mode and source clock (fc/27)
LD (TC4CR), 00011000B ; Starts TC4
```

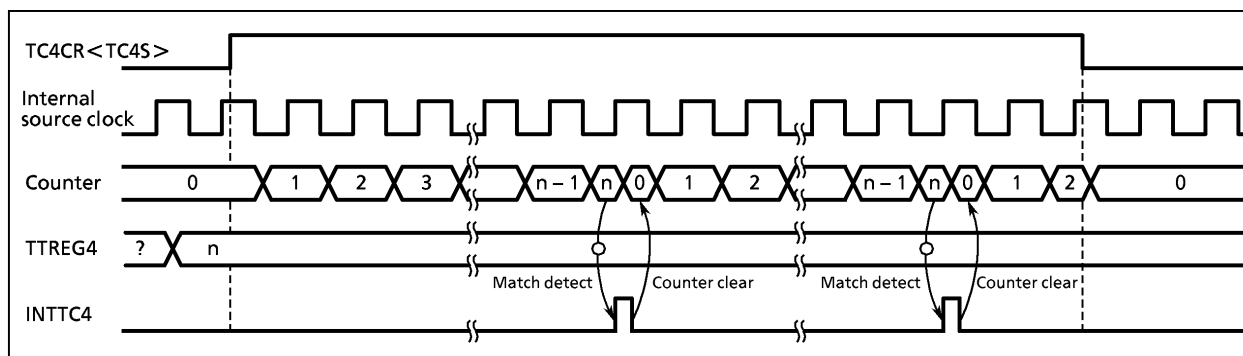


Figure 2-32. 8-bit Timer Mode Timing Chart (In Case of Timer/Counter 4)

(2) 8-bit event counter mode (Timer/counter 3, 4 and 6)

In this mode, events are counted on the falling edge of TC_j pin input. The contents of TTREG_j are compared with the contents of up-counter. If a match is found, an INTTC_j interrupt is generated, and the counter is cleared. The maximum applied frequency is $f_c/2^4$ [Hz] in NORMAL1/2 or IDLE1/2 mode and $f_s/2^4$ [Hz] in SLOW1/2 or SLEEP1/2 mode. Two or more machine cycles are required for both the “H” and “L” levels of the pulse width.

Note 1: In the event counter mode, always write TC_jCR<TFF_j> to “0”. If TFF_j is set to “1”, unexpected pulse may be output from $\overline{PDO_j}/\overline{PWM_j}/\overline{PPG_j}$ pin.

Note 2: In the event counter mode, do not change the setting of timer registers (TTREG_j) while timer/counter is operating. Since TTREG_j is configured as one-stage register, a newly set value is immediately reflected on the timer register.

Note 3: $j = 3, 4, 6$

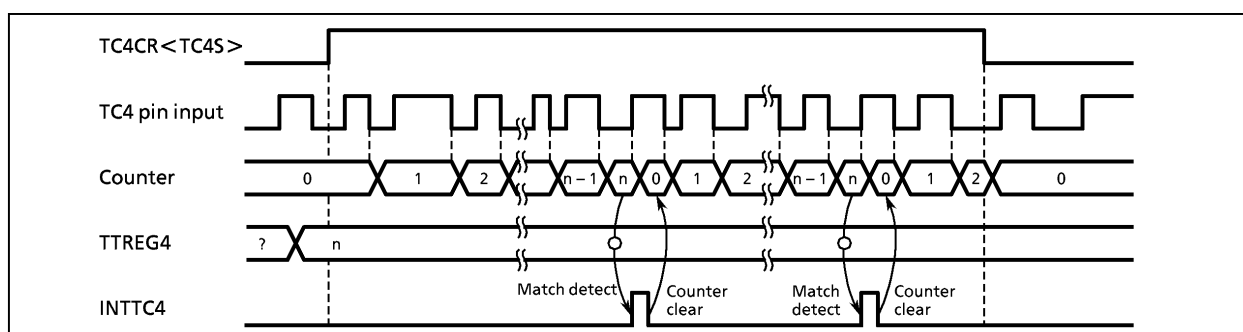


Figure 2-33. Event Counter Mode Timing Chart (In Case of Timer/Counter 4)

(3) 8-bit programmable divider output (PDO) mode (Timer/counter 3, 4 and 6)

The internal clock is used for counting up. The contents of TTREG_j are compared with the contents of the up-counter. Timer F/F_j output is toggled and the counter is cleared each time a match is found. Timer F/F_j output is inverted and output to the $\overline{PDO_j}$ pin. When used as a this mode, respective output latch should be set to “1”. This mode can be used for 50% duty pulse output. Timer F/F_j can be initialized by program, and it is initialized to “0” during reset. An INTTC_j interrupt is generated each time the $\overline{PDO_j}$ output is toggled.

Example: Output a 1024 Hz pulse (at $f_c = 16$ MHz, MULSEL<MUL2> = “0”, in case of TC4)

```
SET (P3DR). 2 ; P32 output latch ← 1
LD (TTREG4), 3DH ; (1/1024 ÷ 27/fc) ÷ 2 = 3DH
LD (TC4CR), 00010001B ; Sets the 8-bit PDO mode and source clock (fc/27)
LD (TC4CR), 00011001B ; Starts TC4
```

Note 1: In the programmable divider output(PDO) mode, do not change the setting of timer registers (TTREG_j) while timer/counter is operating. Since TTREG_j is configured as one-stage register, a newly set value is immediately reflected on the timer register.

Note 2: If PDO output is stopped during output operation, the output state is maintained at the state immediately before timer/counter is stopped. For changing the level of $\overline{PDO_j}$ pin, modify TC_jCR<TFF_j> after timer/counter has been stopped. Do not execute halt of timer/counter and modification of TFF_j simultaneously.

Example: Fixes $\overline{PDO_j}$ output at high level after timer/counter is stopped

```
CLR (TCjCR).3 ; Stops timer/counter.
```

```
CLR (TCjCR).7 ; Sets  $\overline{PDO_j}$  output to high level output
```

Note 3: $j = 3, 4, 6$

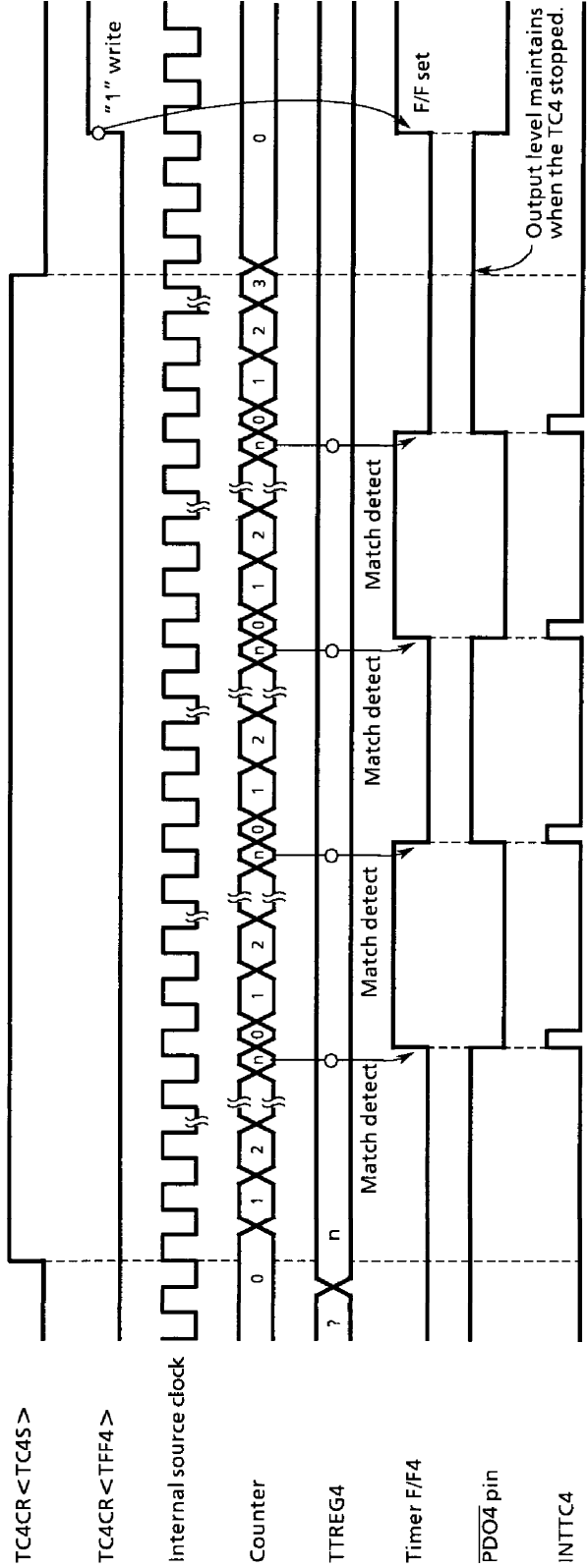


Figure 2-34. 8-Bit PDO Mode Timing Chart (In Case of Tmer/Counter 4)

(4) 8-bit pulse width modulation (PWM) output mode (Timer/counter 3, 4 and 6)

PWM output with a resolution of 8 bits is possible. The internal clock is used for counting up. The contents of PWREG_i are compared with the contents of up-counter. If a match is found, the timer F/F_i output is toggled. The counter continues counting. And, when an overflow occurs, the timer F/F_i output is again toggled and the counter is cleared. Timer F/F_i output is inverted and output to the \overline{PWMi} pin. An INTTC_i interrupt is generated when an overflow occurs.

In PWM mode, because PWREG_i becomes a 2-stage registers with shift register, it is possible to change the setting value of PWREG_i while timer/counter is operating. Therefore, output can be altered continuously. The shift operation of PWREG_i to shift register is executed at the INTTC_i timing. While timer/counter is operating, the data by read instruction is not a setting value of PWREG_i but a value of shift register. Therefore, after writing to PWREG_i, the reading data of PWREG_i is previous value till INTTC_i is generated.

While timer/counter stops, written value to PWREG_i is shifted to shift register immediately.

Note 1: In PWM mode, write to the timer register PWREG_i immediately after an INTTC_i interrupt is generated (normally during the INTTC_i interrupt service routine). If writing to PWREG_i and INTTC_i interrupt occur at the same time, the unstable value being written is shifted. This may cause pulses different from the set value to be output until the next INTTC_i interrupt is generated.

Note 2: If PWM output is stopped during output operation, the output state is maintained at the state immediately before timer/counter is stopped. For changing the level of \overline{PWMi} , modify TCiCR<TFFi> after timer/counter has been stopped. Do not execute halt of timer/counter and modification of TFFi simultaneously.

Example: Fixes \overline{PWMi} output at high level after timer/counter is stopped

CLR (TCiCR).3 ; Stops timer/counter.

CLR (TCiCR).7 ; Sets \overline{PWMi} output to high level output

Note 3: Before starting STOP mode, disable PWM output. When the timer/counter is enabled and fc, fc/2 or fs is selected as the source clock, pulse is output from PWM pin during warming-up after releasing STOP mode.

Note 4: i = 3, 4, 6

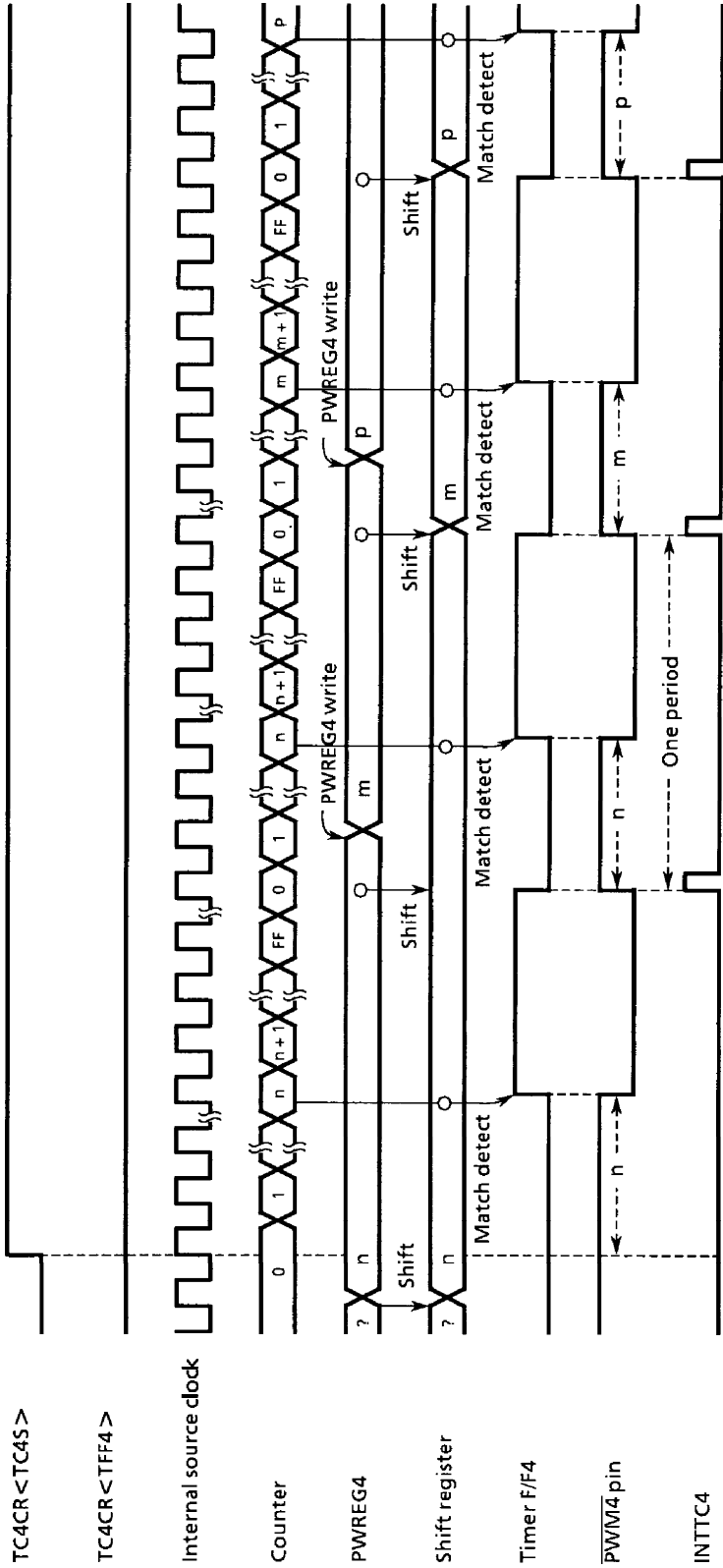


Figure 2-35. 8-Bit PWM Mode Timing Chart (In Case of Timer/Counter 4)

Table 2-10. PWM Output Mode

Source clock			Resolution		Maximum setting time	
NORMAL1/2, IDLE1/2 mode		SLOW1/2, SLEEP1/2 mode	fc = 16 MHz	fs = 32.768 kHz	fc = 16 MHz	fs = 32.768 kHz
DV7CK = 0	DV7CK = 1					
fc/2 ¹¹ [Hz]	fs/2 ³ [Hz]	fs/2 ³ [Hz]	128 [μ s]	244.14 [μ s]	32.8 [ms]	62.5 [ms]
fc/2 ⁷	fc/2 ⁷	–	8 [μ s]	–	2.05 [ms]	–
fc/2 ⁵	fc/2 ⁵	–	2 [μ s]	–	512 [μ s]	–
fc/2 ³	fc/2 ³	–	500 [ns]	–	128 [μ s]	–
fs	fs	fs	30.5 [μ s]	30.5 [μ s]	7.81 [ms]	7.81 [ms]
fc/2	fc/2	–	125 [ns]	–	32 [μ s]	–
fc	fc	–	62.5 [ns]	–	16 [μ s]	–

(5) 16-bit timer mode (Timer/counter 3 and 4, Timer/counter 5 and 6)

In this mode, counting up is performed using the internal clock.

Timer/counter 3 and 4 (5 and 6) are also available as a 16-bit timer mode by cascade connection.

a. 16-bit timer mode of timer/counter 3 and 4

If a match is found, the INTTC4 interrupt is generated and the counter is cleared to “0”. Counting up resumes after the counter is cleared. The timer register should write to the TTREG3 more first than TTREG4. The timer register must not write only either TTREG3 or TTREG4.

b. 16-bit timer mode of timer/counter 5 and 6

If a match is found, the INTTC6 interrupt is generated and the counter is cleared to “0”. Counting up resumes after the counter is cleared. The timer register should write to the TTREG5 more first than TTREG6. The timer register must not write only either TTREG5 or TTREG6.

Note 1: In the timer mode, always write TCjCR<TFFj> to “0”. If TFFj is set to “1”, unexpected pulse may be output from PDOj/PWMj/PPGj pin.

Note 2: In the timer mode, do not change the setting of timer registers (TTREGi) while timer/counter is operating. Since TTREGi is configured as one-stage register, a newly set value is immediately reflected on the timer register.

Note 3: j = 3, 4, 6 i = 3 to 6

Table 2-11. Source Clock of 16-Bit Timer Mode

Source clock			Resolution		Maximum setting time	
NORMAL1/2, IDLE1/2 mode		SLOW1/2, SLEEP1/2 mode	At fc = 16 MHz	At fs = 32.768 kHz	At fc = 16 MHz	At fs = 32.768 kHz
DV7CK = 0	DV7CK = 1					
fc/2 ¹¹	fs/2 ³ [Hz]	fs/2 ³	128 [μ s]	244.14 [μ s]	8.39 [s]	16 [s]
fc/2 ⁷	fc/2 ⁷	–	8 [μ s]	–	524.3 [ms]	–
fc/2 ⁵	fc/2 ⁵	–	2 [μ s]	–	131.1 [ms]	–
fc/2 ³	fc/2 ³	–	500 [ns]	–	32.8 [ms]	–

Example: Set the 16-bit timer mode with source clock fc/2⁷ [Hz] and generates an interrupt 300 [ms] later (at fc = 16 [MHz])

```
LDW (TTREG3), 927CH ; Sets the timer register (300 ms ÷ 27/fc = 927CH)
DI
SET (EIRH). EF11 ; Enable INTTC4 interrupt
EI
LD (TC3CR), 13H ; Sets the 16-bit timer mode (lower) and source clock
LD (TC4CR), 04H ; Sets the 16-bit timer mode (upper)
LD (TC4CR), 0CH ; Starts timer/counter
```

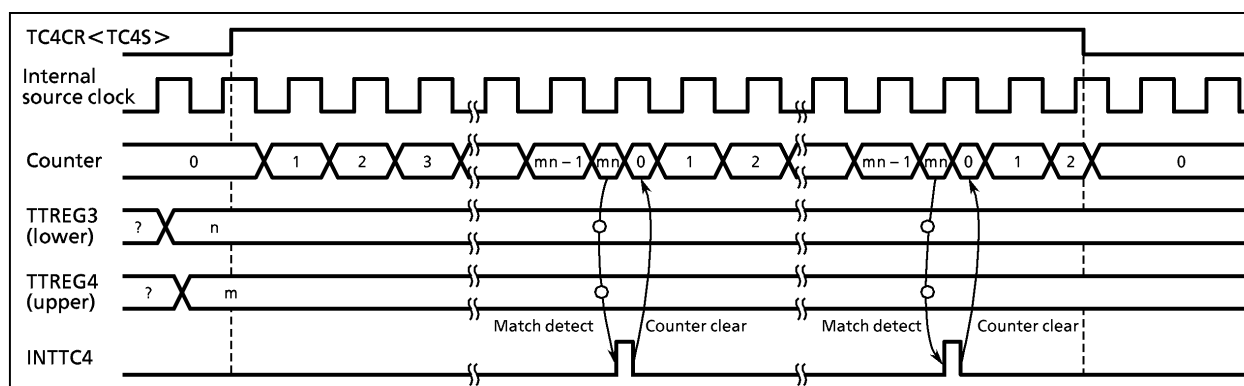


Figure 2-36. 16-Bit Timer Mode Timing Chart (In Case of Timer/Counter 3 and 4)

(6) 16-bit event counter mode (Timer/counter 3 and 4)

In this mode, events are counted on the falling edge of the TC3 pin input. Timer/Counter 5 and 6 are cannot use a 16-bit event counter mode. Timer/Counter 3 and 4 are also available as a 16-bit event counter mode by cascade connection.

a. 16-bit event counter mode of timer/counter 3 and 4

If a match is found, the INTTC4 interrupt is generated and the counter is cleared to "0". After the counter is cleared, counting up resumes every falling edge of TC3 input. The maximum applied frequency is $f_c/2^4$ [Hz] in NORMAL1/2 or IDLE1/2 mode and $f_s/2^4$ [Hz] in SLOW1/2 or SLEEP1/2 mode. Two or more machine cycles are required for both the "H" and "L" levels of the pulse width. The timer register should write to the TTREG3 more first than TTREG4. The timer register must not write only either TTREG3 or TTREG4.

Note 1: In the event counter mode, always write TCjCR<TFj> to "0". If TFFj is set to "1", unexpected pulse may be output from PDOj/PWMj/PPGj pin.

Note 2: In the event counter mode, do not change the setting of timer registers (TTREGj) while timer/counter is operating. Since TTREGj is configured as one-stage register, a newly set value is immediately reflected on the timer register.

Note 3: j = 3, 4

(7) 16-bit pulse width modulation (PWM) output mode (Timer/counter 3 and 4, Timer/counter 5 and 6)

PWM output with a resolution of 16 bits is possible. Timer/counter 3 and 4 (5 and 6) are also available as a 16-bit PWM output mode by cascade connection.

a. 16-bit PWM output mode of timer/counter 3 and 4

The contents of PWREG3/4 are compared with the contents of up-counter. If a match is found, the timer F/F4 output is toggled. The counter continues counting. And, when an overflow occurs, the timer F/F4 output is again toggled and the counter is cleared. Timer F/F4 output is inverted and output to the PWM4 pin. An INTTC4 interrupt is generated when an overflow occurs. When used as PWM4 pin, respective output latch should be set to "1". In PWM mode, because PWREG4/3 each becomes a 2-stage registers with shift register, it is possible to change the setting value of PWREG4/3 while timer/counter is operating. Therefore, output can be altered continuously. The shift operation of PWREG4/3 to shift register is executed at the INTTC4 timing. While timer/counter is operating, the data by read instruction is not a setting value of PWREG4/3 but a value of shift register. Therefore, after writing to PWREG4/3, the reading data of these registers is previous value till INTTC4 is generated.

While timer/counter stops, written value to PWREG4/3 is shifted to shift register immediately. When writing to PWREG4/3, always write to the lower side (PWREG3) and then the upper side (PWREG4) in that order. Writing to only lower side (PWREG3) or the upper side (PWREG4) has no effect.

b. 16-bit PWM output mode of timer/counter 5 and 6

The contents of PWREG5/6 are compared with the contents of up-counter. If a match is found, the timer F/F6 output is toggled. The counter continues counting. And, when an overflow occurs, the timer F/F6 output is again toggled and the counter is cleared. Timer F/F6 output is inverted and output to the $\overline{\text{PWM6}}$ pin. An INTTC6 interrupt is generated when an overflow occurs. When used as $\overline{\text{PWM6}}$ pin, respective output latch should be set to “1”. In PWM mode, because PWREG6/5 each becomes a 2-stage registers with shift register, it is possible to change the setting value of PWREG6/5 while timer/counter is operating. Therefore, output can be altered continuously. The shift operation of PWREG6/5 to shift register is executed at the INTTC6 timing. While timer/counter is operating, the data by read instruction is not a setting value of PWREG6/5 but a value of shift register. Therefore, after writing to PWREG6/5, the reading data of these registers is previous value till INTTC6 is generated.

While timer/counter stops, written value to PWREG6/5 is shifted to shift register immediately. When writing to PWREG6/5, always write to the lower side (PWREG5) and then the upper side (PWREG6) in that order. Writing to only lower side (PWREG5) or the upper side (PWREG6) has no effect.

Note 1: In PWM mode, write to the timer register PWREG m,n immediately after an INTTC m interrupt is generated (normally during the INTTC m interrupt service routine). If writing to PWREG m, n and INTTC m interrupt occur at the same time, the unstable value being written is shifted. This may cause pulses different from the set value to be output until the next INTTC m interrupt is generated.

Note 2: If PWM output is stopped during output operation, the output state is maintained at the state immediately before timer/counter is stopped. For changing the level of $\overline{\text{PWM}i}$, modify $\text{TCiCR} < \text{TTFi} >$ after timer/counter has been stopped. Do not execute halt of timer/counter and modification of TTFi simultaneously.

Example: Fixes $\overline{\text{PWM}i}$ output at high level after timer/counter is stopped

CLR (TCiCR).3 ; Stops timer/counter

CLR (TCiCR).7 ; Sets $\overline{\text{PWM}i}$ output to high level output

Note 3: Before starting STOP mode, disable PWM output. When the timer/counter is enabled and f_c , $f_c/2$ or f_s is selected as the source clock, pulse is output from PWM pin during warming-up after releasing STOP mode.

Note 4: $m = 4$ and $n = 3$, or $m = 6$ and $n = 5$, or $i = 4, 6$

Table 2-12. 16-Bit PWM Output Mode

Source clock		Resolution	Maximum setting time			
NORMAL1/2, IDLE1/2 mode	SLOW1/2, SLEEP1/2 mode		At $f_c = 16$ MHz	At $f_s = 32.768$ kHz		
DV7CK = 0	DV7CK = 1	At $f_c = 16$ MHz	At $f_s = 32.768$ kHz	At $f_c = 16$ MHz	At $f_s = 32.768$ kHz	
$f_c/2^{11}$	$f_s/2^3$ [Hz]	$f_s/2^3$	128 [μs]	244.14 [μs]	8.39 [s]	16 [s]
$f_c/2^7$	$f_c/2^7$	–	8 [μs]	–	524.3 [ms]	–
$f_c/2^5$	$f_c/2^5$	–	2 [μs]	–	131.1 [ms]	–
$f_c/2^3$	$f_c/2^3$	–	500 [ns]	–	32.8 [ms]	–
f_s	f_s	f_s	30.5 [μs]	30.5 [μs]	2 [s]	2 [s]
$f_c/2$	$f_c/2$	–	125 [ns]	–	8.2 [ms]	–
f_c	f_c	–	62.5 [ns]	–	4.1 [ms]	–

Example: Extract the pulse, whose term and “high” width is 32.768 ms and 1ms respectively, from P32 width 16-bit PWM mode (at $f_c = 16$ MHz, $\text{MULSEL} < \text{MUL2} > = “0”$, $\text{DV7CK} = 0$)

SET (P3DR).2 ; Sets P32 output data latch to “1”

LDW(PWREG3), 07D0H ; Sets pulse width

LD (TC3CR), 33H ; Sets the 16-bit PWM mode (lower) and source clock ($f_c/2^3$)

LD (TC4CR), 056H ; Sets the TFF4 to “1” and sets the 16-bit PWM mode (upper)

LD (TC4CR), 05EH ; Starts timer/counter

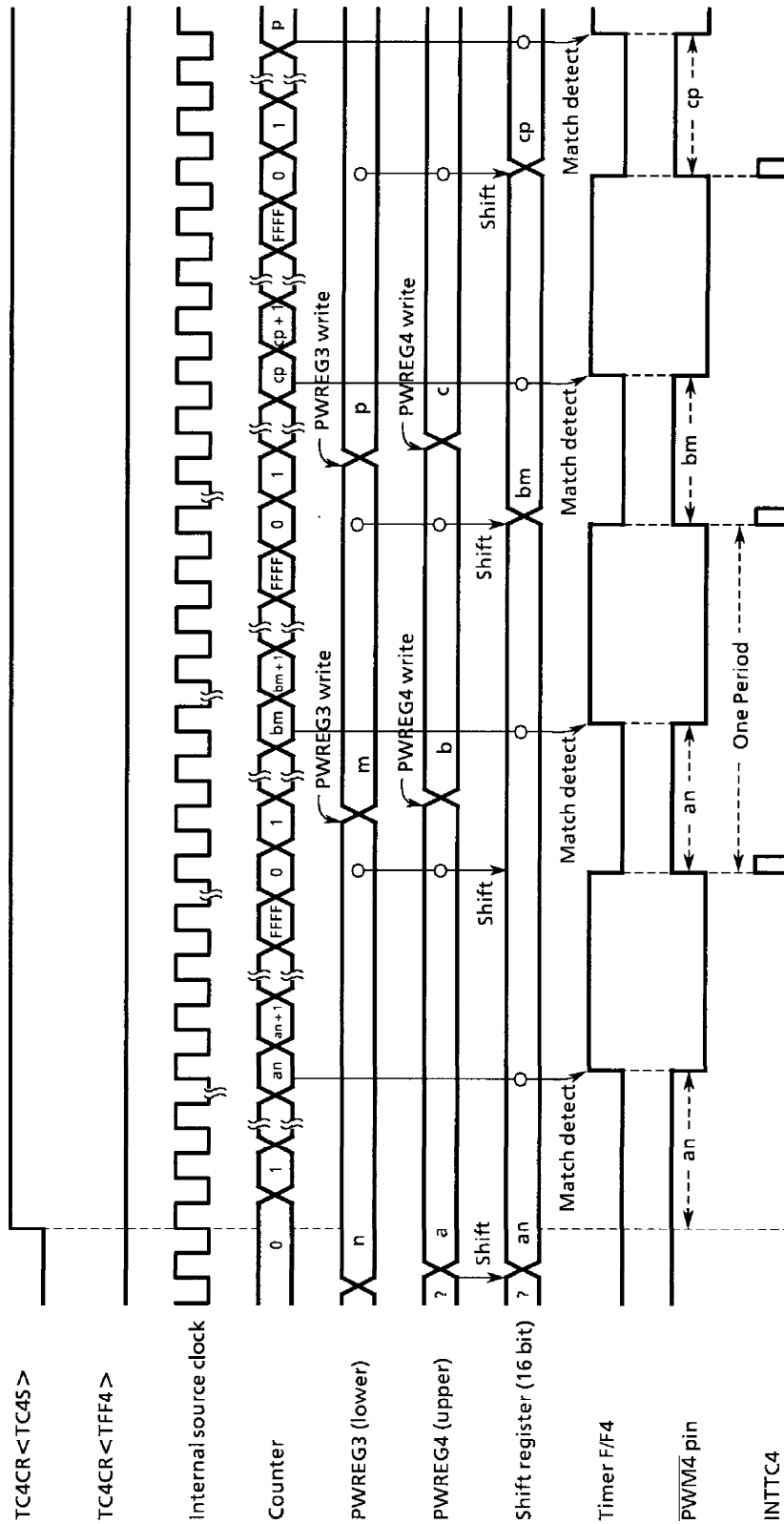


Figure 2-37. 16-Bit PWM Mode Timing Chart (In Case of Timer/Counter 3 and 4)

- (8) 16-bit programmable pulse generate (PPG) output mode
(Timer/Counter 3 and 4, Timer/Counter 5 and 6)

PPG output with a resolution of 16 bits is possible. Timer/Counter 3 and 4 (5 and 6) are also available as a 16-bit PPG output mode by cascade connection.

- a. 16-bit PPG output mode of timer/counter 3 and 4

First, the contents of PWREG3/4 are compared with the contents of up-counter. If a match is found, the timer F/F4 output is toggled. Next, timer F/F4 is again toggled and the counter is cleared by matching with TTREG3/4. The INTTC4 interrupt is generated at this time.

When used as $\overline{PPG4}$ pin, respective output latch should be set to "1". During reset, the F/F4 is initialized to "0".

The F/F4 output is configured by TC4CR<TFF4>. Therefore, the $\overline{PPG4}$ can output either output high or output low at first time. The timer register should write to the PWREG3/TTREG3 more first than PWREG4/TTREG4. The timer register must not write only either PWREG3/TTREG3 or PWREG4/TTREG4.

- b. 16-bit PPG output mode of timer/counter 5 and 6

First, the contents of PWREG5/6 are compared with the contents of up-counter. If a match is found, the timer F/F6 output is toggled. Next, timer F/F6 is again toggled and the counter is cleared by matching with TTREG5/6. The INTTC6 interrupt is generated at this time.

When used as $\overline{PPG6}$ pin, respective output latch should be set to "1". During reset, the F/F6 is initialized to "0".

The F/F6 output is configured by TC6CR<TFF6>. Therefore, the $\overline{PPG6}$ can output either output high or output low at first time. The timer register should write to the PWREG5/TTREG5 more first than PWREG6/TTREG6. The timer register must not write only either PWREG5/TTREG5 or PWREG6/TTREG6.

Example: Extract the pulse, whose term and "high" width is 16.385 ms and 1ms respectively, from P32 with 16-bit PPG mode (at $f_c = 16$ MHz, DV7CK = 0)

```
SET (P3DR).2           ; Sets P32 output data latch to "1"
LDW (PWREG3), 07D0H    ; Sets pulse width
LDW (TTREG3), 8002H    ; Sets pulse term
LD (TC3CR), 33H        ; Sets the 16-bit PPG mode (lower) and source clock ( $f_c/2^3$ )
LD (TC4CR), 057H       ; Sets the TFF4 to "1" and sets the 16-bit PPG mode(upper)
LD (TC4CR), 05FH       ; Starts timer/counter
```

Note 1: In the programmable pulse generate (PPG) mode, do not change the setting of timer registers (PWREG i , TTREG i) while timer/counter is operating. Since PWREG i , TTREG i are configured as one-stage register, a newly set value is immediately reflected on the timer register.

Note 2: If PPG output is stopped during output operation, the output state is maintained at the state immediately before timer/counter is stopped. For changing the level of \overline{PPGj} , modify TC i CR<TFF i > after timer/counter has been stopped. Do not execute halt of timer/counter and modification of TFF j simultaneously.

Example: Fixes \overline{PPGj} output at high level after timer/counter is stopped

CLR (TC j CR).3 ; Stops timer/counter

CLR (TC j CR).7 ; Sets \overline{PPGj} output to high level output

Note 3: $j = 4, 6$ $i = 3$ to 6

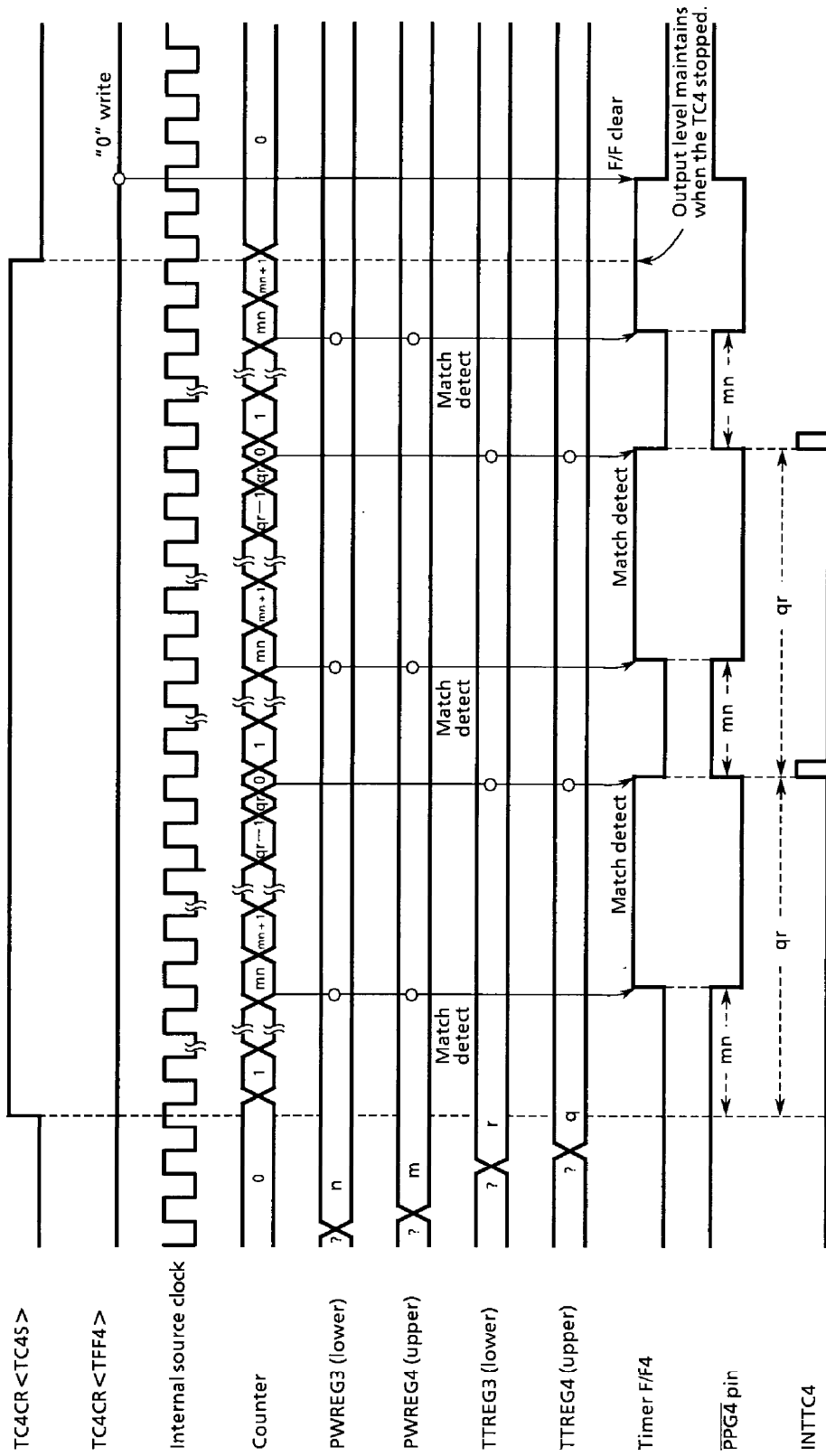


Figure 2-38. 16-Bit PPG Mode Timing Chart (In Case of Timer/Counter 3 and 4)

(9) Warming-up counter mode

In this mode, the warming-up period for switching the main system clock can be generated. Timer/Counter 3 and 4 (5 and 6) are used as a 16-bit timer by cascade connection.

There are 2 modes in warming-up counter mode, one is a mode from NORMAL to SLOW and the other is a mode from SLOW to NORMAL.

Note 1: In the warming-up mode, always write $TCiCR<TFFi>$ to "0". If $TFFi$ is set to "1", unexpected pulse may be output from $\overline{PDOi}/\overline{PWMi}/\overline{PPGi}$ pin.

Note 2: In the warming-up mode, the lower 8 bits of $TTREGm,n$ are ignored and an interrupt is generated by matching the upper 8 bits.

Note 3: $i = 3, 4, 6$ $m = 4$ and $n = 3$, or $m = 6$ and $n = 5$

a. Warming-up counter mode for low-frequency (NORMAL1→NORMAL2→SLOW2→SLOW1)

In this mode, it can obtain the warming-up period till the oscillation for low-frequency (fs) is stabilized.

Before timer/counter is started, turn on low-frequency oscillation by setting $SYSCR2<XTEN>$ to "1".

After timer/counter is started by setting $TCmCR<TCmS>$, the contents of $TTREGm,n$ are compared with the contents of up-counter. If a match is found, an $INTTCm$ interrupt is generated, and the counter is cleared to "0".

In the interrupt service program, stop the timer/counter and change the system clock to low-frequency clock by setting $SYSCR2<SYSCK>$ to "1".

After that, halt the high-frequency oscillation by clearing $SYSCR2<XEN>$ to "0".

Table 2-13. Warming-up Period for Low-frequency Oscillation
(at fs = 32.768 kHz)

Min (at $TTREGm, n = 0100H$)	Max (at $TTREGm, n = FF00H$)
7.81 ms	1.99 s

Example: Switching to the SLOW1 mode after low-frequency clock has stabilized by using TC4, 3.

```

SET   (SYSCR2).6      ; SYSCR2<XTEN> ← "1"
LD    (TC3CR), 43H    ; TFF3="0", fs for source clock, sets 16-bit mode
LD    (TC4CR), 05H    ; TFF4="0", sets warming-up counter mode
LD    (TTREG3), 8000H ; sets warming-up time (depend on oscillator characteristics)
DI    ; IMF ← "0"
SET   (EIRH).3        ; Enables INTTC4
EI    ; IMF ← "1"
SET   (TC4CR).3       ; Starts TC4, 3
:      :
PINTTC4: CLR (TC4CR).3 ; Stops TC4, 3
SET   (SYSCR2).5       ; SYSCR2<SYSCK> ← "1"
                        ; (Switches the main system clock to the low-frequency clock)
CLR   (SYSCR2).7       ; SYSCR2<XEN> ← "0" (Turns off low-frequency oscillation)
RETI
:      :
VINTTC4: DW  PINTTC4   ; INTTC4 vector table

```

b. Warming-up counter mode for high-frequency(SLOW1→SLOW2→NORMAL2→NORMAL1)

In this mode, it can obtain the warming-up period till the oscillation for high-frequency (fc) is stabilized.

Before timer/counter is started, turn on high-frequency oscillation by setting SYSCR2<XEN> to "1".

After timer/counter is started by setting TCmCR<TCmS>, the contents of TTREGm,n are compared with the contents of up-counter. If a match is found, an INTTCm interrupt is generated, and the counter is cleared to "0".

In the interrupt service program, stop the timer/counter and change the system clock to high-frequency clock by clearing SYSCR2<SYSCK> to "0".

After that, halt the low-frequency oscillation by clearing SYSCR2<XTEN> to "0".

Table 2-14. Warming-up Period for High-frequency (at fc = 16 MHz)

Min (at TTREGm, n = 0100H)	Max (at TTREGm, n = FF00H)
16 μ s	4.08 ms

Example: Switching to the NORMAL1 mode after high-frequency clock has stabilized by using TC4, 3.

```

SET   (SYSCR2).7      ; SYSCR2<XEN> ← "1"
LD    (TC3CR), 63H    ; TFF3 = "0", fc for source clock, sets 16-bit mode
LD    (TC4CR), 05H    ; TFF4 = "0", sets warming-up counter mode
LD    (TTREG3), 0F800H ; Sets warming-up time (depend on oscillator characteristics)
DI                      ; IMF ← "0"
SET   (EIRH).3        ; Enables INTTC4
EI                      ; IMF ← "1"
SET   (TC4CR).3       ; Starts TC4, 3
      :                ;
PINTTC4: CLR (TC4CR).3 ; Stops TC4, 3
      CLR (SYSCR2).5   ; SYSCR2<SYSCK> ← "0"
                        ; (Switches the main system clock to the high-frequency clock)
      CLR (SYSCR2).6   ; SYSCR2<XTEN> ← "0"
                        ; (Turns off high-frequency oscillation)
      RETI
      :                ;
VINTTC4: DW  PINTTC4   ; INTTC4 vector table

```


2.9.2 Control

UART is controlled by the UART control registers (UARTCR1, UARTCR2). The operating status can be monitored using the UART status register (UARTSR).

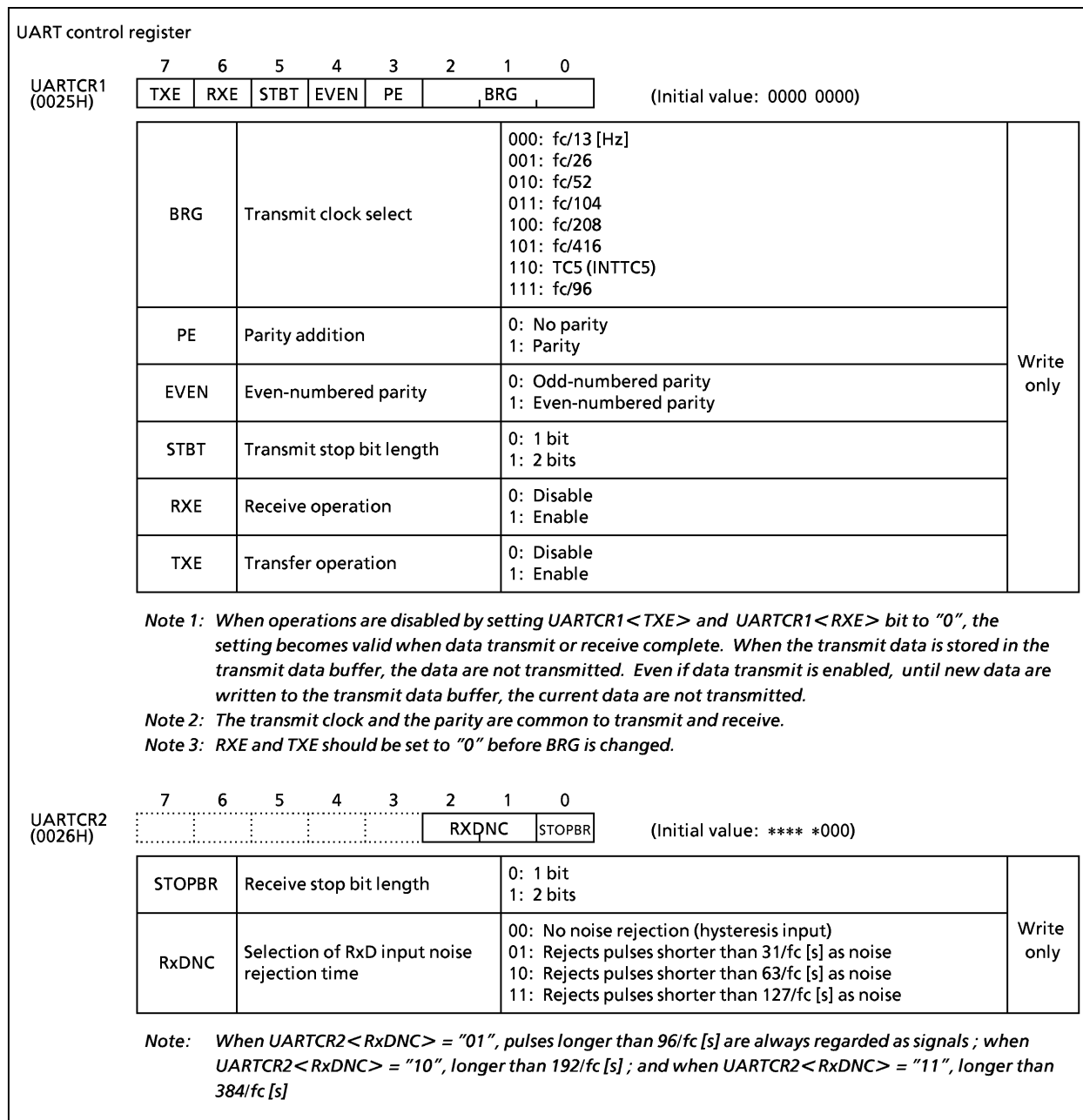


Figure 2-40. UART Control Register

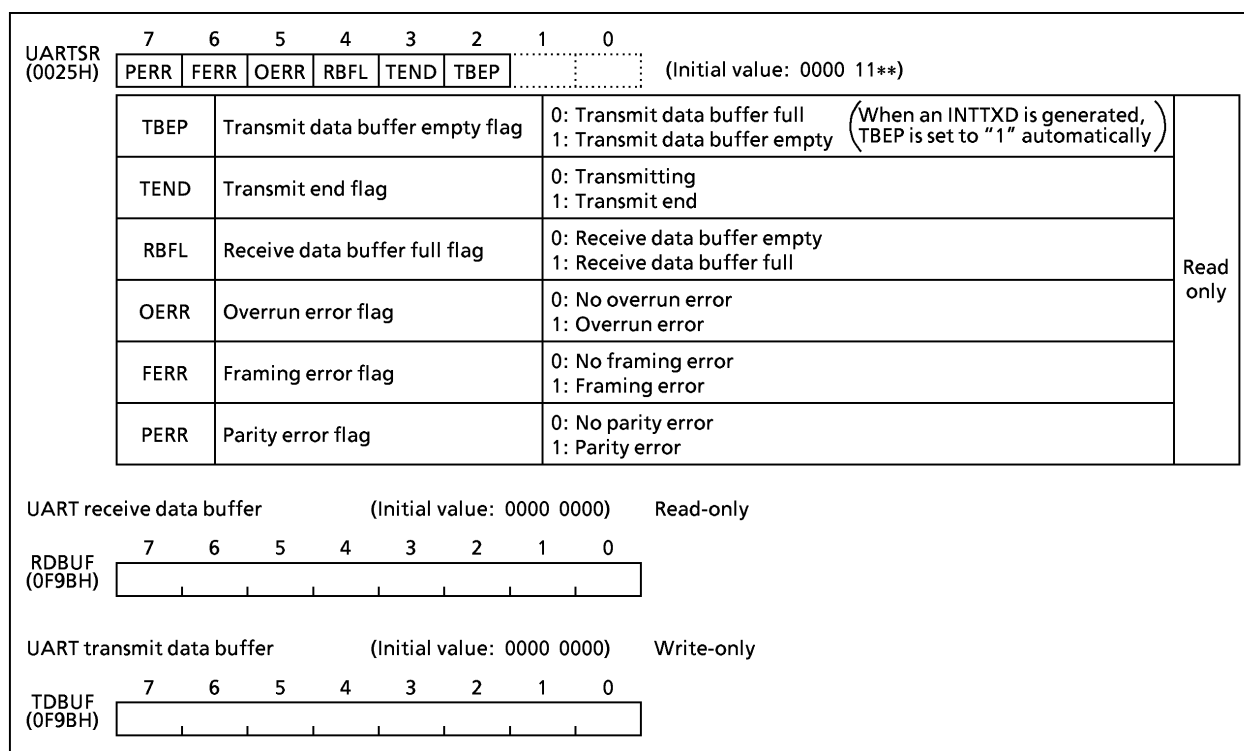


Figure 2-41. UART Status Register and Data Buffer Registers

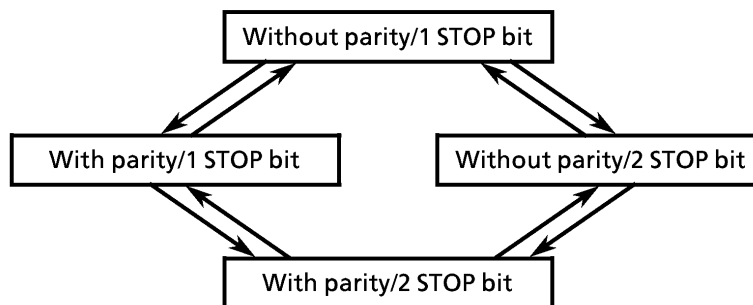
2.9.3 Transfer Data Format

In UART, a one-bit start bit (low level) , stop bit (bit length selectable at high level by UARTCR1<STBT>), and parity (select parity in UARTCR1<PE>; even-or odd-numbered parity by UARTCR1<EVEN>) are added to the transfer data. The transfer data formats are shown as follow.

Table 2-15. Transfer Data Format

PE	STBT	Frame length									
		1	2	3	-----	8	9	10	11	12	
0	0										
0	1										
1	0										
1	1										

Note: In order to switch the transmit data format, perform transmit operations in the following sequence except for the initial setting.



2.9.4 Transfer Rate

The baud rate of UART is set of UARTCR1 <BRG>. The example of the baud rate shown as follows.

Table 2-16. Transfer Rate

BRG	Source clock		
	16 MHz	8 MHz	4 MHz
000	76800 [baud]	38400 [baud]	19200 [baud]
001	38400	19200	9600
010	19200	9600	4800
011	9600	4800	2400
100	4800	2400	1200
101	2400	1200	600

When TC5 is used as the UART transfer rate (when UARTCR1 <BRG> = “110”), the transfer clock and transfer rate are determined as follows:

$$\text{Transfer clock} = \frac{\text{TC5 source clock}}{\text{TTREG5 set value}}$$

$$\text{Transfer rate} = \frac{\text{Transfer clock}}{16}$$

2.9.5 Data Sampling

The UART receiver keeps sampling input using the clock selected by UARTCR1 <BRG> until a start bit is detected in RxD pin input. RT clock starts detecting “L” level of the RxD pin. Once a start bit is detected, the start bit, data bits, stop bit(s), and parity bit are sampled at three times of RT7, RT8, and RT9 during one receiver clock interval (RT clock). (RT0 is the position where the bit supposedly starts). Bit is determined according to majority rule (the data are the same twice or more out of three samplings).

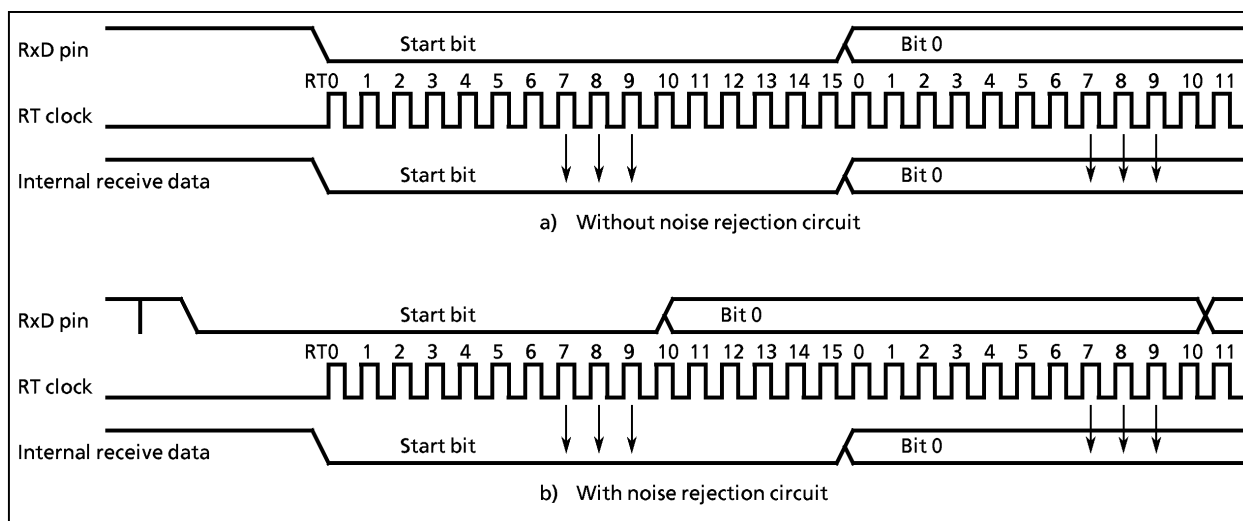


Figure 2-42. Data Sampling

2.9.6 STOP Bit Length

Select a transmit stop bit length (1 or 2 bits) by UARTCR1 <STBT>.

2.9.7 Parity

Set parity/no parity by UARTCR1<PE>; set parity type (odd-or even-numbered) by UARTCR1<EVEN>.

2.9.8 Transmit/Receive

(1) Data transmit

Set UARTCR1<TXE> to “1”. Read UARTSR to check TBEP=“1”, then write data in TDBUF (transmit data buffer). Writing data in TDBUF zero-clears UARTSR<TBEP>, transfers the data to the transmit shift register and the data are sequentially output from the TxD pin. The data output include a one-bit start bit, stop bits whose number is specified in UARTCR1<STBT> and a parity bit if parity addition is specified. Select the data transfer baud rate using bits 0 to 2 in UARTCR1. When data transmit starts, transmit buffer empty flag UARTSR<TBEP> is set to “1” and an INTTXD interrupt is generated.

While UARTCR1<TXE> = “0” and from when “1” is written to UARTCR1<TXE> to when send data are written to TDBUF, the TXD pin is fixed at High level.

When transmitting data, first read UARTSR, then write data in TDBUF. Otherwise, UARTSR<TBEP> is not zero-cleared and transmit does not start.

(2) Data receive

Set UARTCR1<RXE> to “1”. When data are received via the RxD pin, the receive data are transferred to RDBUF (receive data buffer). At this time, the data transmitted include a start bit and stop bit(s) and a parity bit if parity addition is specified. When stop bit(s) are received, data only are extracted and transferred to RDBUF (receive data buffer). Then the receive buffer full flag UARTSR<RBFL> is set and an INTRXD interrupt is generated. Select the data transfer baud rate using bits 0 to 2 in UARTCR1.

If an overrun error (OERR) occurs when data are received, the data are not transferred to RDBUF (receive data buffer) but discarded; data in the RDBUF are not affected.

Note: When a receive operation is disabled by setting UARTCR1<RXE> bit to “0”, the setting becomes valid when data receive is completed. However, if a framing error occurs in data receive, the receive-disabling setting may not become valid. if a framing error occurs, be sure to perform a re-receive operation.

2.9.9 Status Flag/Interrupt Signal

(1) Parity error

When parity determined using the receive data bits differs from the received parity bit, the parity error flag UARTSR<PERR> is set to “1”. The UARTSR<PERR> is cleared to “0” when the RDBUF is read after reading the UARTSR.

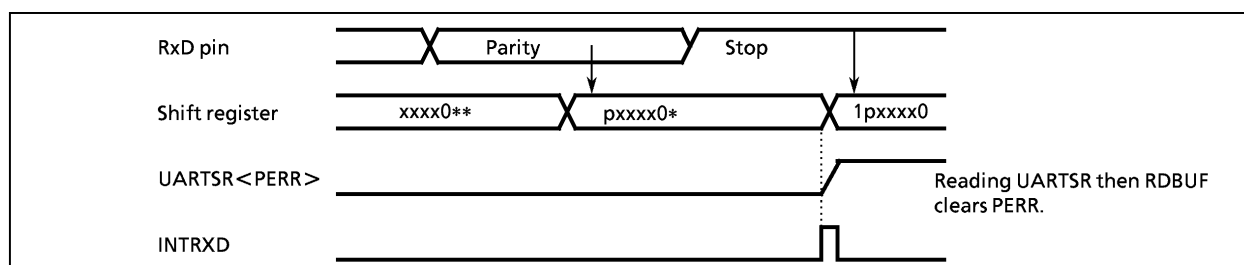


Figure 2-43. Generation of Parity Error

(2) Framing error

When “0” is sampled as the stop bit in the receive data, framing error flag UARTSR<FERR> is set to “1”. The UARTSR<FERR> is cleared to “0” when the RDBUF is read after reading the UARTSR.

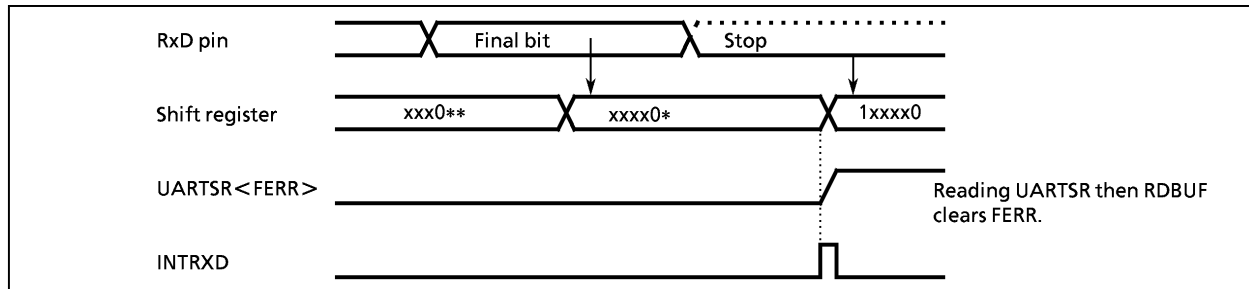


Figure 2-44. Generation of Framing Error

(3) Overrun error

When all bits in the next data are received while unread data are still in RDBUF, overrun error flag UARTSR<OERR> is set to “1”. In this case, the receive data is discarded ; data in RDBUF are not affected. The UARTSR<OERR> is cleared to “0” when the RDBUF is read after reading the UARTSR.

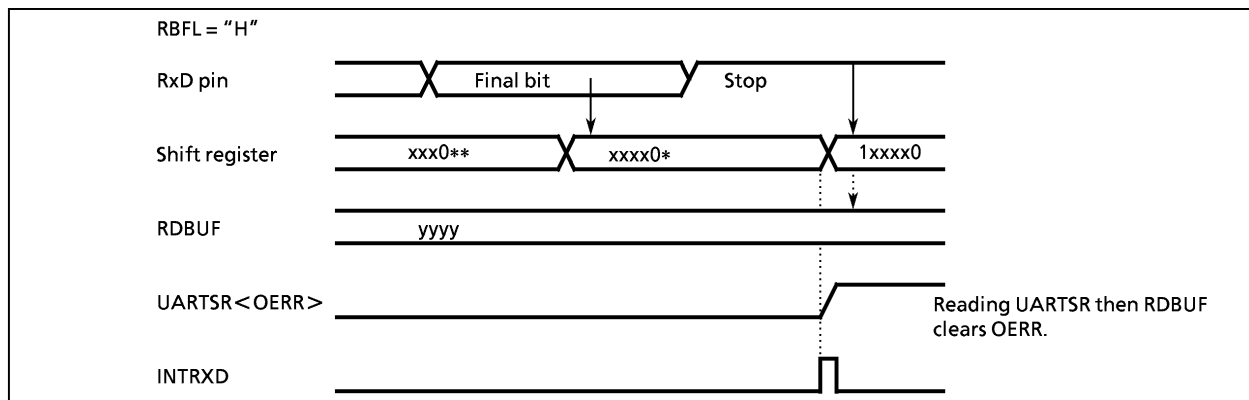


Figure 2-45. Generation of Overrun Error

(4) Receive data buffer full

Loading the received data in RDBUF sets receive data buffer full flag UARTSR<RBFL>. The UARTSR<RBFL> is cleared to “0” when the RDBUF is read after reading the UARTSR.

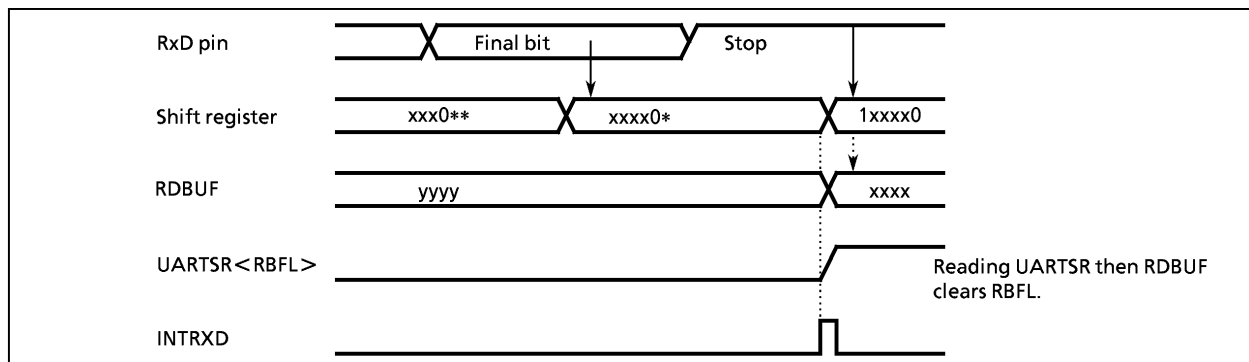


Figure 2-46. Generation of Receive Buffer Full

(5) Transmit data buffer empty

When no data is in the transmit buffer TDBUF, UARTSR<TBEP> is set to “1”, that is, when data in TDBUF are transferred to the transmit shift register and data transmit starts, transmit data buffer empty flag UARTSR<TBEP> is set to “1”. The UARTSR<TBEP> is cleared to “0” when the TDBUF is written after reading the UARTSR.

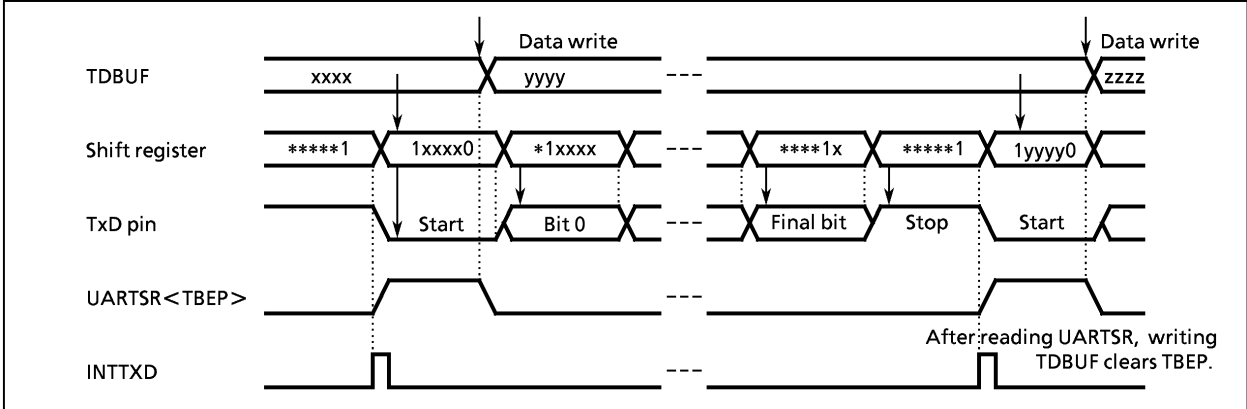


Figure 2-47. Generation of Transmit Buffer Empty

(6) Transmit end flag

When data are transmitted and no data is in TDBUF (UARTSR<TBEP> = “1”), transmit end flag UARTSR<TEND> is set to “1”. The UARTSR<TEND> is cleared to “0” when the data transmit is stated after writing the TDBUF.

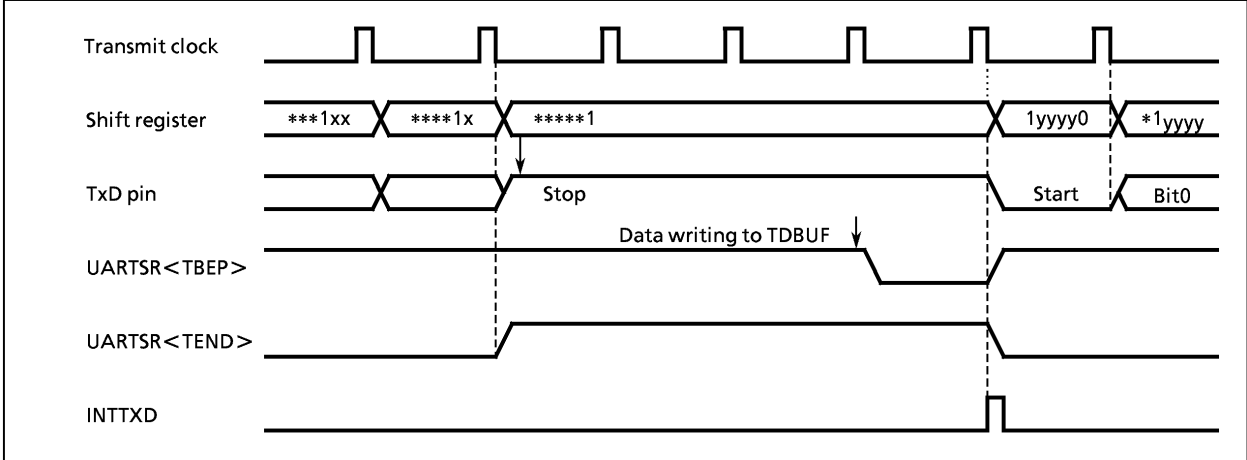


Figure 2-48. Generation of Transmit Buffer Empty

2.10 Serial Interface (SIO0, SIO1)

The TMP86CM25/S25 have two clocked-synchronous 8-bit serial interface. Serial interface has an 8-byte transmit and receive data buffer that can automatically and continuously transfer up to 64 bits of data.

The serial interface is connected to external devices via pins P16 (SO0), P15 (SI0), P17 ($\overline{SCK0}$) and P76 (SO1), P75 (SI1), P73 ($\overline{SCK1}$). The serial interface pins are also used as port P1 port P7. When these pins are used as serial interface pins, the correspondence output latch should be set to "1". In the transmit mode, pin P15, P75 can be used as normal I/O port, and in the receive mode, the pin P16, P76 can be used as normal I/O ports.

2.10.1 Configuration

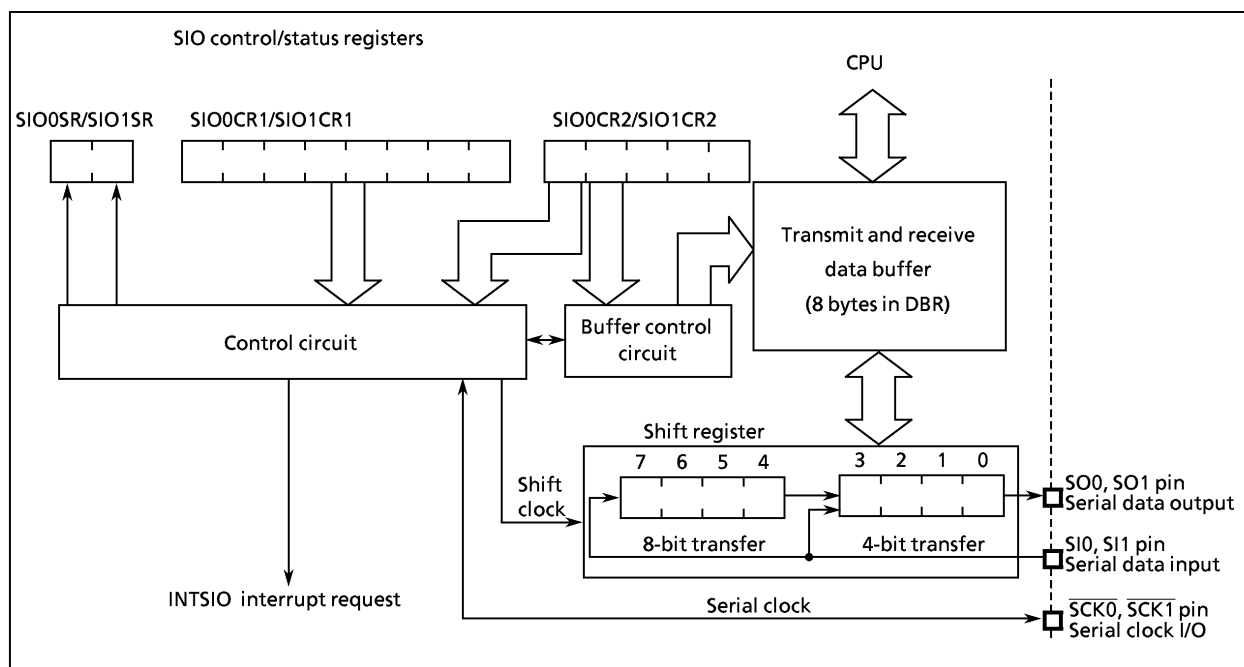


Figure 2-49. Serial Interfaces

2.10.2 Control

The serial interface is controlled by SIO control registers (SIO0CR1/SIO0CR2, SIO1CR1/SIO1CR2). The serial interface status can be determined by reading SIO status register (SIO0SR, SIO1SR). The transmit and receive data buffer is controlled by the BUF (bits 2 to 0 in SIO0CR2, SIO1CR2). The data buffer is assigned to address 0F90H to 0F97H, 0FA0H to 0FA7H for SIO in the DBR area, and can continuously transfer up to 8 words (bytes or nibbles) at one time. When the specified number of words has been transferred, a buffer empty (in the transmit mode) or a buffer full (in the receive mode or transmit/receive mode) interrupt (INTSIO0, INTSIO1) is generated. When the internal clock is used as the serial clock in the 8-bit receive mode and the 8-bit transmit/receive mode, a fixed interval wait can be applied to the serial clock for each word transferred. Four different wait times can be selected with WAIT (bits 4 and 3 in SIO0CR2, SIO1CR2).

SIO Control Register 1		7	6	5	4	3	2	1	0	
SIO0CR1 (0F98H)	SIOS	SIO INH	SIOM			SCK			(Initial value: 0000 0000)	
SIO1CR1 (0FA8H)	SIOS	Indicate transfer start/stop		0: Stop 1: Start		Write only				
	SIOINH	Continue/Abort transfer		0: Continue transfer 1: Abort transfer (automatically cleared after abort)						
	SIOM	Transfer mode select		000: 8-bit transmit mode 010: 4-bit transmit mode 100: 8-bit transmit/receive mode 101: 8-bit receive mode 110: 4-bit receive mode Except the above; reserved						
	SCK	Serial clock select			NORMAL1/2, IDLE1/2 mode					SLOW, SLEEP mode
			DV7CK = 0	DV7CK = 1						
000			$fc/2^{13}$	$fs/2^5$	$fs/2^5$					
001			$fc/2^8$	$fc/2^8$	-					
010			$fc/2^7$	$fc/2^7$	-					
011			$fc/2^6$	$fc/2^6$	-					
100			$fc/2^5$	$fc/2^5$	-					
101	$fc/2^4$	$fc/2^4$	-							
110	Reserved									
111	External clock (input from SCK pin)									

Note 1: fc: High-frequency clock [Hz], fs: Low-frequency clock [Hz]
Note 2: Set SIOS to "0" and SIOINH to "1" when setting the transfer mode or serial clock.
Note 3: SIO0CR1, SIO1CR1 is write-only register, which cannot access any of in read-modify-write instruction such as bit operate, etc.

SIO Status Registers		7	6	5	4	3	2	1	0	
SIO0SR (0F99H)	SIOF	SEF								(Initial value: 00** ****)
SIO1SR (0FA9H)	SIOF	Serial transfer operating status monitor	0: Transfer terminated 1: Transfer in process		(After SIOS is cleared to "0", SIOF is cleared to "0" at the termination of transfer or setting of SIOINH.)				Read only	
	SEF	Shift operating status monitor	0: Shift operation terminated 1: Shift operation in process							

Figure 2-50. SIO Control Register and Status Register (1/2)

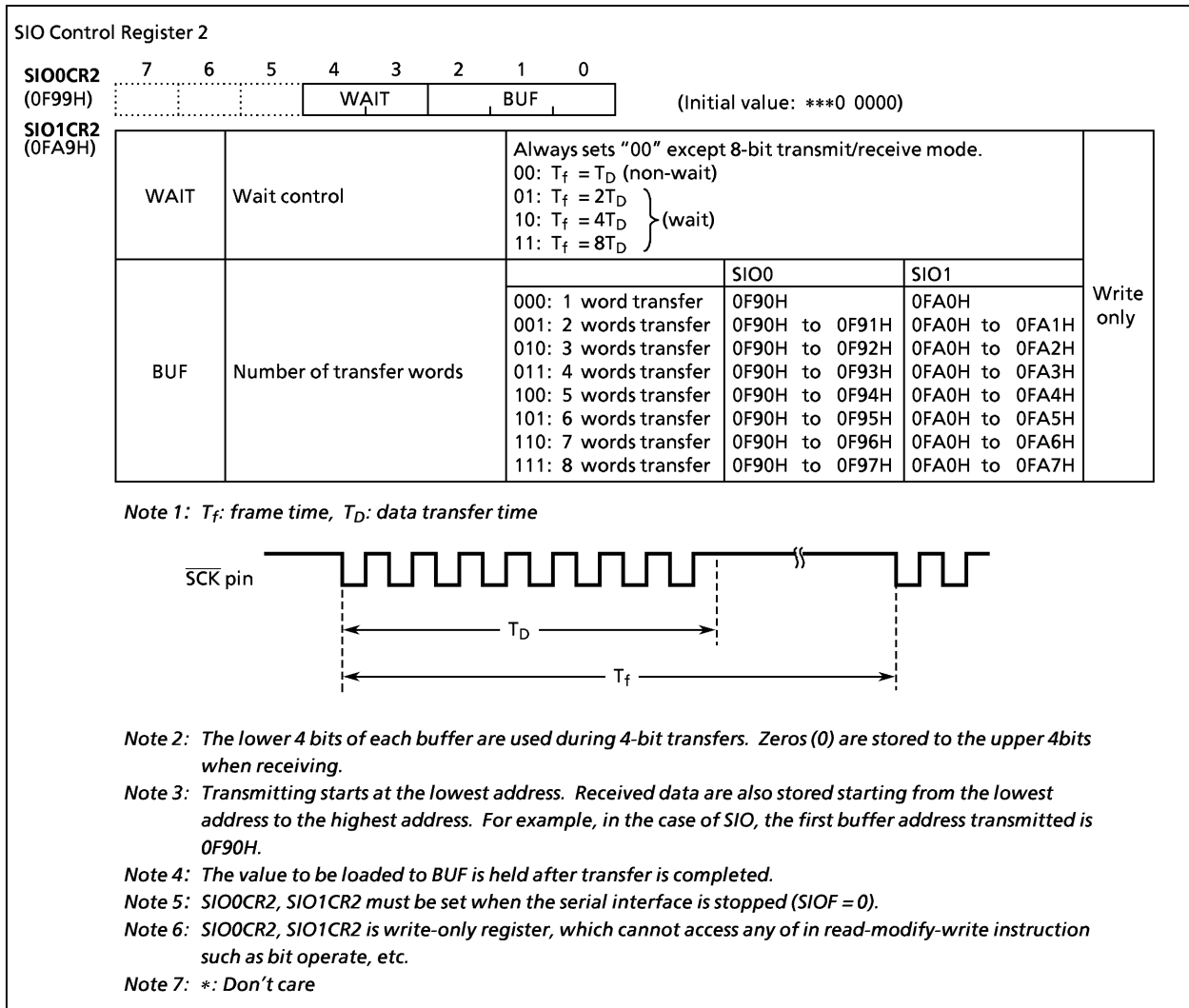


Figure 2-51. SIO Control Register and Status Register (2/2)

(1) Serial clock

a. Clock Source

SCK (bits 2 to 0 in SIO0CR, SIO1CR) is able to select the following:

① Internal clock

Any of four frequencies can be selected. The serial clock is output to the outside on the $\overline{\text{SCK0}}$, $\overline{\text{SCK1}}$ pin. The $\overline{\text{SCK0}}$, $\overline{\text{SCK1}}$ pin goes high when transfer starts.

When data writing (in the transmit mode) or reading (in the receive mode or the transmit/receive mode) cannot keep up with the serial clock rate, there is a wait function that automatically stops the serial clock and holds the next shift operation until the read/write processing is completed.

Table 2-17. Serial Clock Rate

SCK	NORMAL1/2, IDLE1/2 modes				SLOW, SLEEP modes	
	DV7CK = 0		DV7CK = 1		Clock	Baud rate
Clock	Baud rate	Clock	Baud rate	Clock		
000	$fc/2^{13}$	1.91 Kbps	$fs/2^5$	1024 bps	$fs/2^5$	1024 bps
001	$fc/2^8$	61.04 Kbps	$fc/2^8$	61.04 Kbps	-	-
010	$fc/2^7$	122.07 Kbps	$fc/2^7$	122.07 Kbps	-	-
011	$fc/2^6$	244.14 Kbps	$fc/2^6$	244.14 Kbps	-	-
100	$fc/2^5$	488.28 Kbps	$fc/2^5$	488.28 Kbps	-	-
101	$fc/2^4$	976.56 Kbps	$fc/2^4$	976.56 Kbps	-	-
110	-	-	-	-	-	-
111	External	-	External	-	External	-

1 Kbit = 1024 bit
 (fc = 16 MHz, fs = 32.768 kHz)

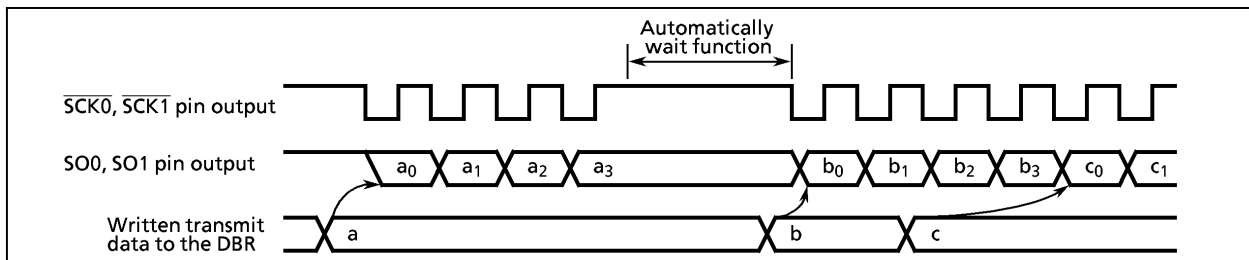
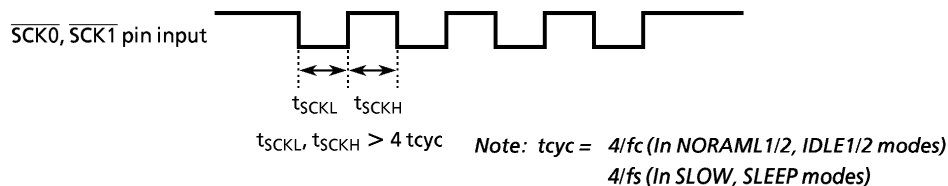


Figure 2-52. Clock Source (Internal Clock)

② External clock

An external clock connected to the $\overline{\text{SCK0}}$, $\overline{\text{SCK1}}$ pin is used as the serial clock. In this case, the P17 ($\overline{\text{SCK0}}$), P73 ($\overline{\text{SCK1}}$) must be set to the input mode. To ensure shifting, a pulse width of at least 4 machine cycles is required. This pulse is needed for the shift operation to execute certainly. Actually, there is necessary processing time for interrupting, writing, and reading. The minimum pulse is determined by setting the mode and the program.



b. Shift edge

The leading edge is used to transmit, and the trailing edge is used to receive.

① Leading edge

Transmitted data are shifted on the leading edge of the serial clock (falling edge of the $\overline{SCK0}$, $\overline{SCK1}$ pin input/output).

② Trailing edge

Received data are shifted on the trailing edge of the serial clock (rising edge of the $\overline{SCK0}$, $\overline{SCK1}$ pin input/output).

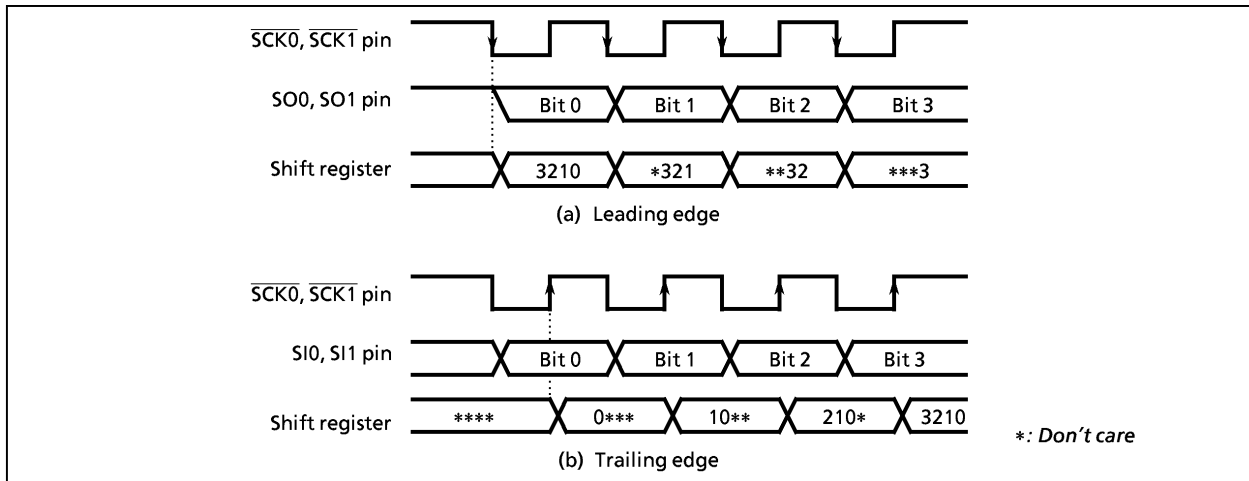


Figure 2-53. Shift Edge

(2) Number of Bits to Transfer

Either 4-bit or 8-bit serial transfer can be selected. When 4-bit serial transfer is selected, only the lower 4 bits of the transmit/receive data buffer register are used. The upper 4 bits are cleared to “0” when receiving.
The data is transferred in sequence starting at the least significant bit (LSB).

(3) Number of Words to Transfer

Up to 8 words consisting of 4 bits of data (4-bit serial transfer) or 8 bits (8-bit serial transfer) of data can be transferred continuously. The number of words to be transferred is loaded to BUF.
An INTSIO0, INTSIO1 interrupt is generated when the specified number of words has been transferred. If the number of words is to be changed during transfer, the serial interface must be stopped before making the change. The number of words can be changed during automatic-wait operation of an internal clock. In this case, the serial interface is not required to be stopped.

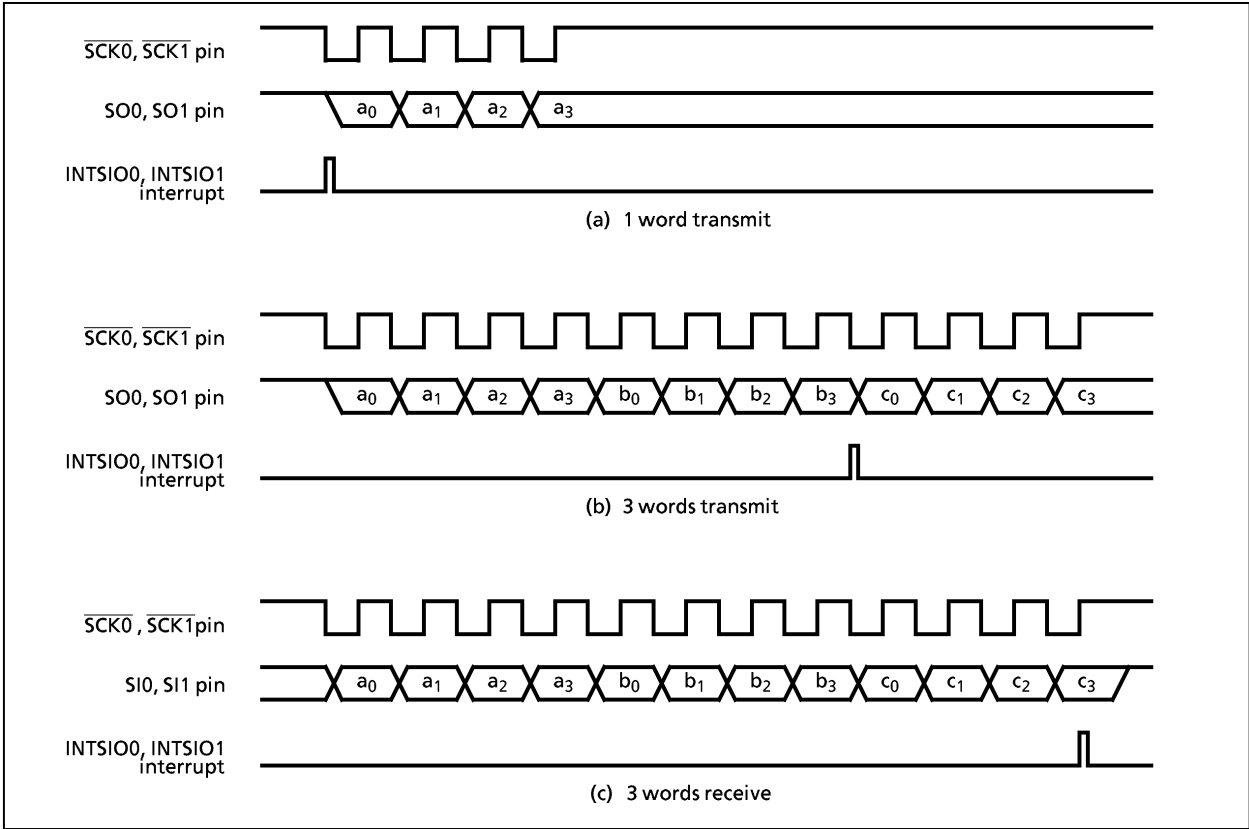


Figure 2-54. Number of Bits to Transfer (Example: 4-bit serial transfer)

2.10.3 Transfer Mode

S10M (bits 5 to 3 in S100CR1, S101CR1) is used to select the transmit, receive, or transmit/receive mode.

(1) 4-bit and 8-bit transmit modes

In these modes, the S100CR1, S101CR1 is set to the transmit mode and then the data to be transmitted first are written to the data buffer registers (DBR). After the data are written, the transmission is started by setting S10S to "1". The data are then output sequentially to the SO0, SO1 pin in synchronous with the serial clock, starting with the least significant bit (LSB). As soon as the LSB has been output, the data are transferred from the data buffer register to the shift register. When the final data bit has been transferred and the data buffer register is empty, an INTS100, INTS101 (buffer empty) interrupt is generated to request the next transmitted data.

When the internal clock is used, the serial clock will stop and an automatic-wait will be initiated if the next transmitted data are not loaded to the data buffer register by the time the number of data words specified with the BUF has been transmitted. Writing even one word of data cancels the automatic-wait; therefore, when transmitting two or more words, always write the next word before transmission of the previous word is completed.

Note: Automatic-waits are also canceled by writing to a DBR not being used as a transmit data buffer register; therefore, during S10 do not use such DBR for other applications. For example, when 3 words are transmitted, do not use the DBR of the remained 5 words.

When an external clock is used, the data must be written to the data buffer register before shifting next data. Thus, the transfer speed is determined by the maximum delay time from the generation of the interrupt request to writing of the data to the data buffer register by the interrupt service program.

The transmission is ended by clearing S10S to "0" or setting S10INH to "1" in buffer empty interrupt service program. That the transmission has ended can be determined from the status of S10F (bit 7 in S100SR, S101SR) because S10F is cleared to "0" when a transfer is completed.

When S10INH is set, the transmission is immediately ended and S10F is cleared to "0".

When an external clock is used, it is also necessary to clear S10S to "0" before shifting the next data; otherwise, dummy data will be transmitted and the operation will end.

If it is necessary to change the number of words, S10S should be cleared to "0", then BUF must be rewritten after confirming that S10F has been cleared to "0".

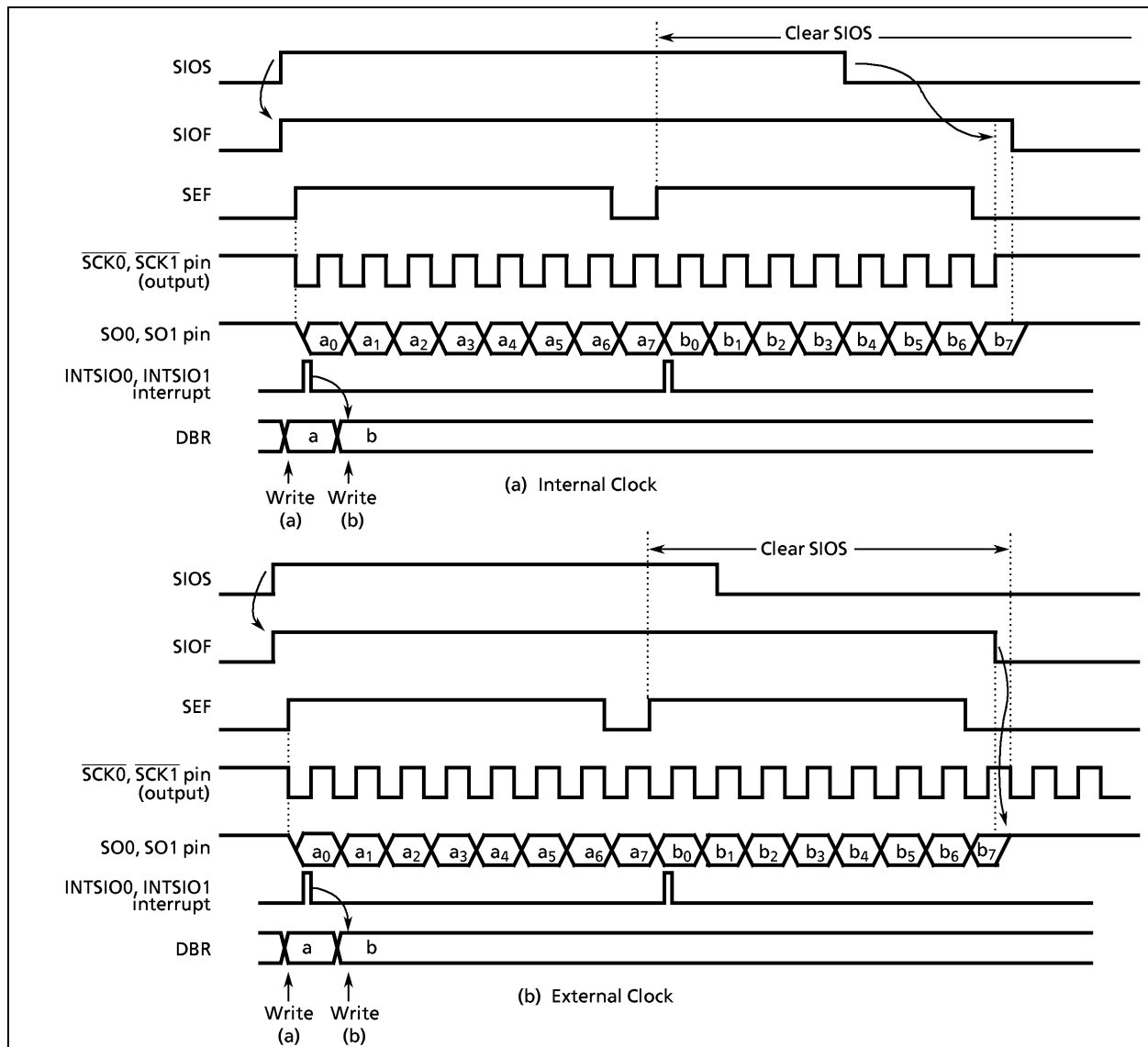


Figure 2-55. Transfer Mode (Example: 8-bit, 1 Word Transfer)

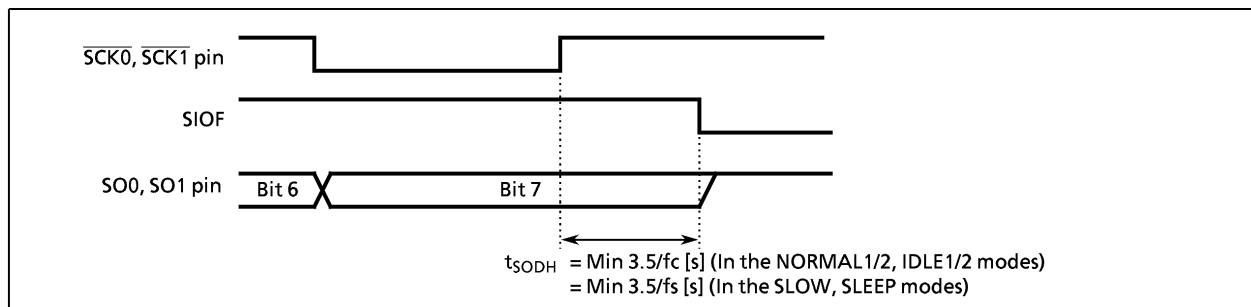


Figure 2-56. Transmitted Data Hold Time at End of Transmit

(2) 4-bit and 8-bit receive modes

After setting the control registers to the receive mode, set SIOS to “1” to enable receiving. The data are then transferred to the shift register via the SI0, SI1 pin in synchronous with the serial clock. When one word of data has been received, it is transferred from the shift register to the data buffer register (DBR). When the number of words specified with the BUF has been received, an INTSIO0, INTSIO1 (buffer full) interrupt is generated to request that these data be read out. The data are then read from the data buffer registers by the interrupt service program.

When the internal clock is used, and the previous data are not read from the data buffer register before the next data are received, the serial clock will stop and an automatic-wait will be initiated until the data are read. A wait will not be initiated if even one data word has been read.

Note: Waits are also canceled by reading a DBR not being used as a received data buffer register is read; therefore, during SIO do not use such DBR for other applications.

When an external clock is used, the shift operation is synchronized with the external clock; therefore, the previous data are read before the next data are transferred to the data buffer register. If the previous data have not been read, the next data will not be transferred to the data buffer register and the receiving of any more data will be canceled. When an external clock is used, the maximum transfer speed is determined by the delay between the time when the interrupt request is generated and when the data received have been read.

The receiving is ended by clearing SIOS to “0” or setting SIOINH to “1” in buffer full interrupt service program. When SIOS is cleared, the current data are transferred to the buffer. After SIOS cleared, the receiving is ended at the time that the final bit of the data has been received. That the receiving has ended can be determined from the status of SIOF (bit 7 in SIO0SR, SIO1SR). SIOF is cleared to “0” when the receiving is ended. After confirmed the receiving termination, the final receiving data is read. When SIOINH is set, the receiving is immediately ended and SIOF is cleared to “0”. (The received data is ignored, and it is not required to be read out.)

If it is necessary to change the number of words in external clock operation, SIOS should be cleared to “0” then BUF must be rewritten after confirming that SIOF has been cleared to “0”.

If it is necessary to change the number of words in internal clock, during automatic-wait operation which occurs after completion of data receiving, BUF must be rewritten before the received data is read out.

Note: The buffer contents are lost when the transfer mode is switched. If it should become necessary to switch the transfer mode, end receiving by clearing SIOS to “0”, read the last data and then switch the transfer mode.

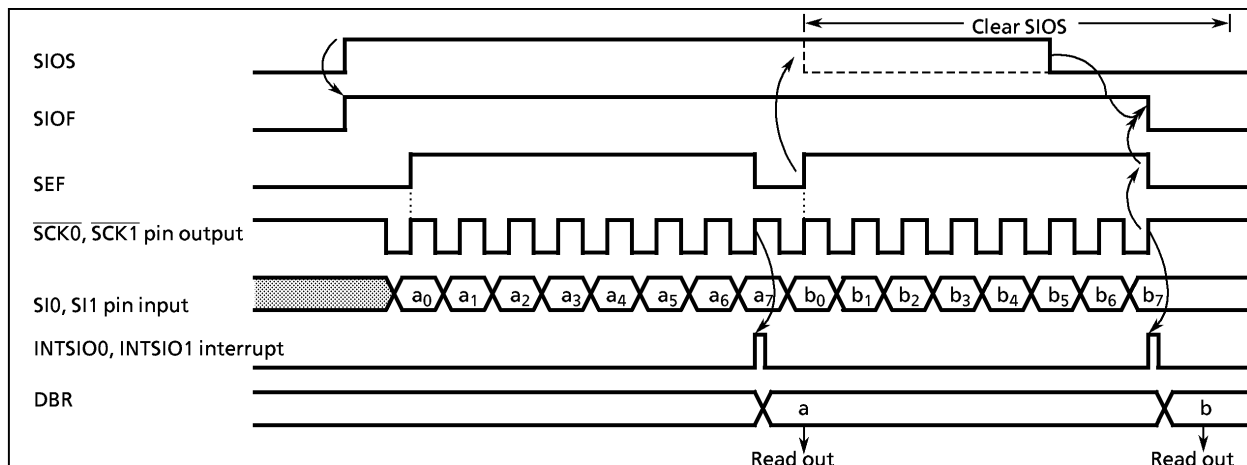


Figure 2-57. Receive Mode (Example: 8-bit, 1 word, internal clock)

(3) 8-bit transmit/receive mode

After setting the control registers to the 8-bit transmit/receive mode, write the data to be transmitted first to the data buffer registers (DBR). After that, enable transeiving by setting SIOS to "1". When transmitting, the data are output from the SO0, SO1 pin at leading edges of the serial clock. When receiving, the data are input to the SI pin at the trailing edges of the serial clock. 8-bit data are transferred from the shift register to the data buffer register. An INTSIO0, INTSIO1 interrupt is generated when the number of data words specified with the BUF has been transferred. The interrupt service program reads the received data from the data buffer register and then writes the data to be transmitted. The data buffer register is used for both transmitting and receiving; therefore, always write the data to be transmitted after reading the received data.

When the internal clock is used, a wait is initiated until the received data are read and the next data are written. A wait will not be initiated if even one data word has been written.

Note: The wait is also canceled by writing to a DBR not being used as a transmit data buffer registers; therefore, during SIO do not use such DBR for other applications.

When an external clock is used, the shift operation is synchronized with the external clock; therefore, it is necessary to read the received data and write the data to be transmitted next before starting the next shift operation. When an external clock is used, the transfer speed is determined by the maximum delay between generation of an interrupt request and the received data are read and the data to be transmitted next are written.

The transmit/receive operation is ended by clearing SIOS to "0" or setting SIOINH to "1" in interrupt service program.

When SIOINH is set, the transmit/receive operation is immediately ended and SIOF is cleared to "0".

If it is necessary to change the number of words in external clock operation, SIOS should be cleared to "0", then BUF must be rewritten after confirming that Siof has been cleared to "0".

If it is necessary to change the number of words in internal clock, during automatic-wait operation which occurs after completion of transmit/receive operation, BUF must be rewritten before reading and writing of the receive/transmit data.

Note: The buffer contents are lost when the transfer mode is switched. If it should become necessary to switch the transfer mode, end receiving by clearing SIOS to "0", read the last data and then switch the transfer mode.

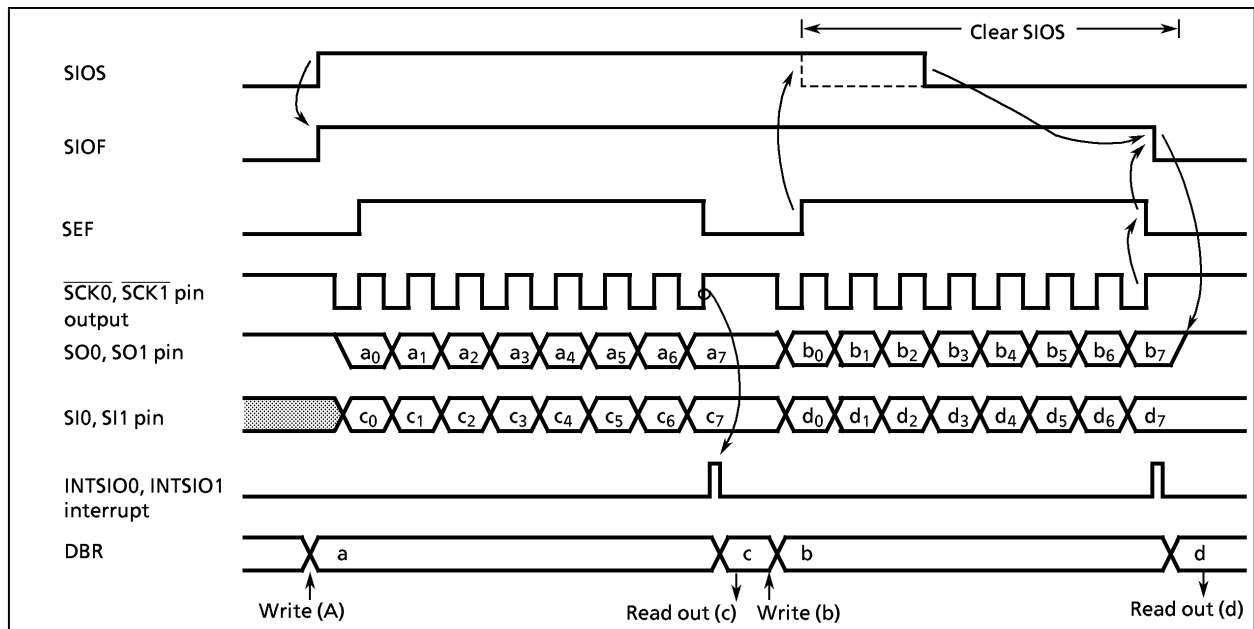


Figure 2-58. Transmit/Receive Mode (Example: 8-bit, 1word, internal clock)

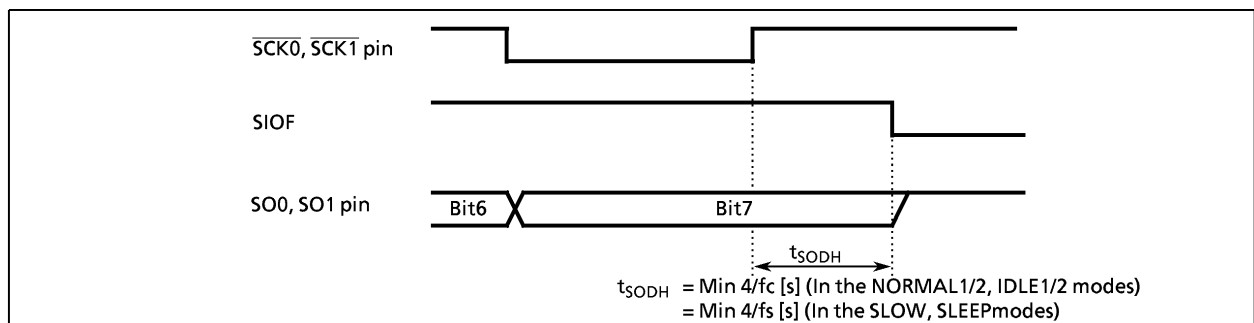


Figure 2-59. Transmitted Data Hold Time at End of Transmit/Receive

2.11 8-Bit AD Converter (ADC)

The TMP86CM25/S25 have a 8-bit successive approximation type AD converter.

2.11.1 Configuration

The circuit configuration of the 8-bit AD converter is shown in Figure 2-60.

It consists of control registers ADCCR1 and ADCCR2, converted value registers ADCDR1 and ADCDR2, a DA converter, a sample-and-hold circuit, a comparator, and a successive comparison circuit.

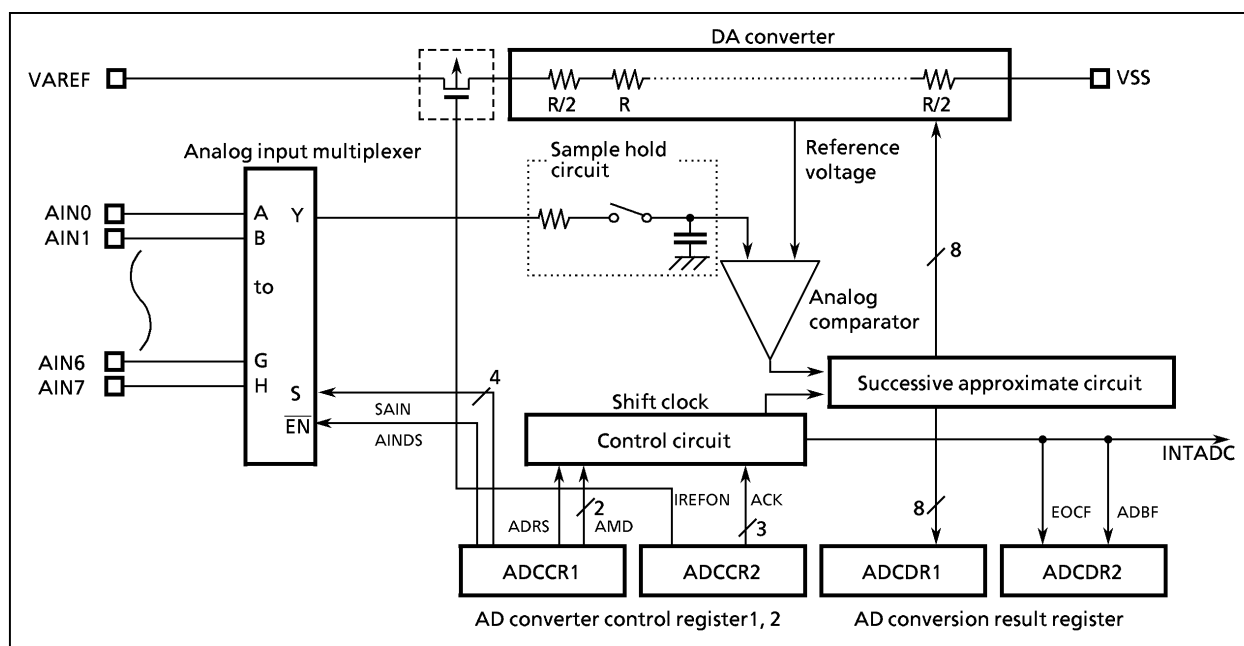


Figure 2-60. AD Converter (ADC)

2.11.2 Register Configuration

The AD converter consists of the following four registers:

- AD converter control register 1 (ADCCR1)
- AD converter control register 2 (ADCCR2)
- AD converted value register H/L (ADCDR1/ADCDR2)

(1) AD converter control register 1 (ADCCR1)

This register selects the analog channels and operation mode in which to perform AD conversion and controls the AD converter as it starts operating.

(2) AD converter control register 2 (ADCCR2)

This register selects the AD conversion time and controls the connection of the DA converter (ladder resistor network).

(3) AD converted value register (ADCDR1)

This register is used to store the digital value after being converted by the AD converter.

(4) AD converted value register (ADCDR2)

This register monitors the operating status of the AD converter.

The AD converter control register configurations are shown in Figures 2-61 and 2-62.

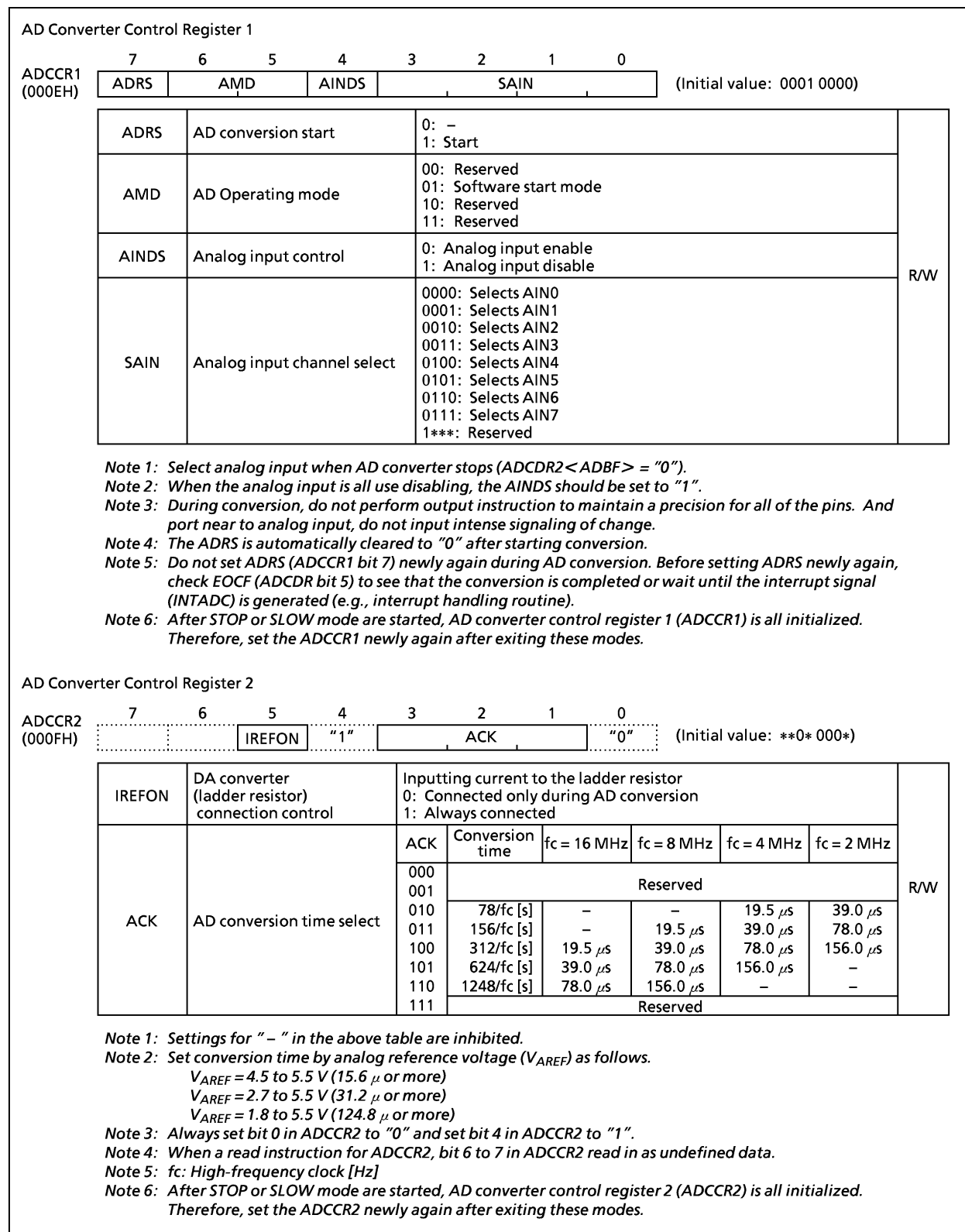


Figure 2-61. AD Converter Control Register

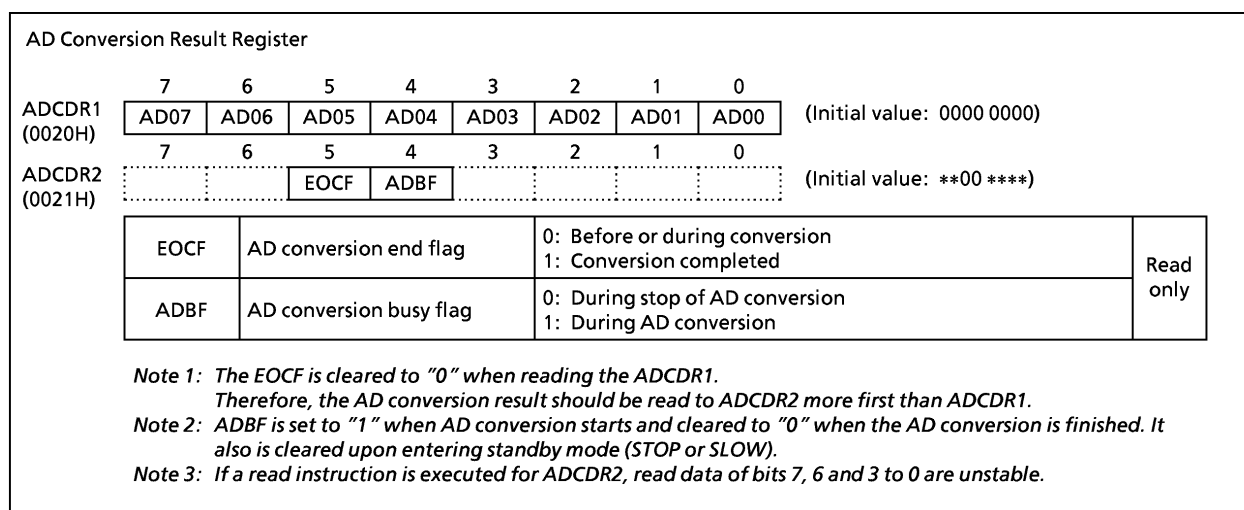


Figure 2-62. AD Converter Result Register

2.11.3 AD Converter Operation

- (1) Set up the AD converter control register 1 (ADCCR1) as follows:
 - Choose the channel to AD convert using AD input channel select (SAIN).
 - Specify analog input enable for analog input control (AINDS).
 - Specify AMD for the AD converter control operation mode (software or repeat mode).
- (2) Set up the AD converter control register 2 (ADCCR2) as follows:
 - Set the AD conversion time using AD conversion time (ACK). For details on how to set the conversion time, refer to Note 2 for AD converter control register 2.
 - Choose IREFON for DA converter control.
- (3) After setting up (1) and (2) above, set AD conversion start (ADRS) of AD converter control register 1 (ADCCR1) to "1".
- (4) After an elapse of the specified AD conversion time, the AD converted value is stored in AD converted value register 1 (ADCDR1) and the AD conversion finished flag (EOCF) of AD converted value register 2 (ADCDR2) is set to "1", upon which time AD conversion interrupt INTADC is generated.
- (5) EOCF is cleared to "0" by a read of the conversion result. However, if reconverted before a register read, although EOCF is cleared the previous conversion result is retained until the next conversion is completed.

2.11.4 AD Converter Operation Modes

(1) Software start mode

After setting AMD (ADCCR1 bits 5, 6) to “01” (software start mode), set ADRS (ADCCR1 bit 7) to “1”. AD conversion of the voltage at the analog input pin specified by SAIN (ADCCR1 bits 0-3) is thereby started.

After completion of the AD conversion, the conversion result is stored in AD converted value registers (ADCDR1) and at the same time EOCF (ADCDR2 bit 5) is set to 1, the AD conversion finished interrupt (INTADC) is generated.

ADRS is automatically cleared after AD conversion has started. Do not set ADRS (ADCCR1 bit 7) newly again (restart) during AD conversion. Before setting ADRS newly again, check EOCF (ADCDR bit 5) to see that the conversion is completed or wait until the interrupt signal (INTADC) is generated (e.g., interrupt handling routine).

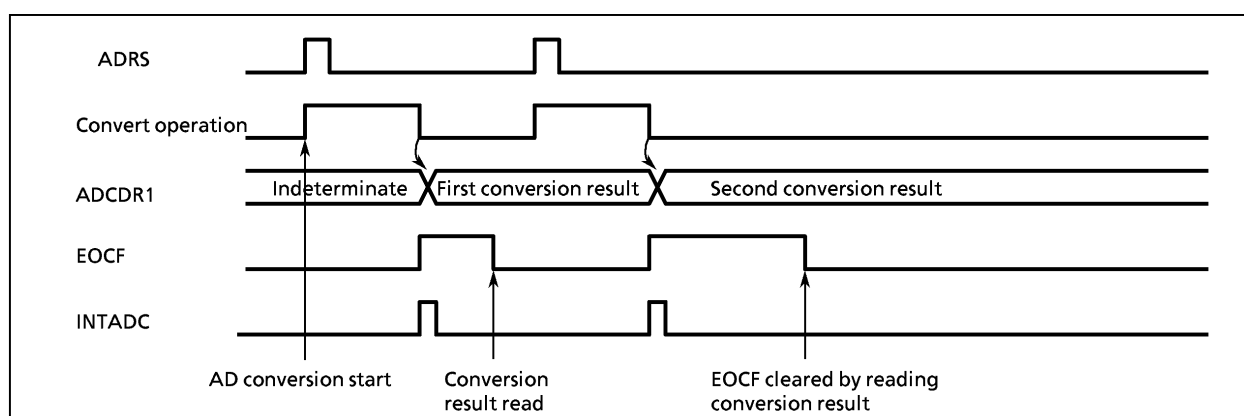


Figure 2-63. Operation in Software Start Mode

2.11.5 STOP and SLOW Modes during AD Conversion

When standby mode (STOP or SLOW mode) is entered forcibly during AD conversion, the AD convert operation is suspended and the AD converter is initialized (ADCCR1 and ADCCR2 are initialized to initial value.). Also, the conversion result is indeterminate. (Conversion results up to the previous operation are cleared, so be sure to read the conversion results before entering standby mode.) When restored from standby mode, AD conversion is not automatically restarted, so it is necessary to restart AD conversion. Note that since the analog reference voltage is automatically disconnected, there is no possibility of current flowing into the analog reference voltage.

2.11.6 Analog Input Voltage and AD Conversion Result

Example: After selecting the conversion time of $19.5 \mu\text{s}$ at 16 MHz and the analog input channel AIN3 pin, perform AD conversion once. After checking EOCF, read the converted value and store the 8-bit data in address $009F_H$ on RAM. The operation mode is software start mode.

```

; AIN SELECT
LD  (P6CR), 00000000B ; P6CR bit 3 = 0
LD  (P6DR), 00000000B ; P6DR bit 3 = 0
LD  (ADCCR1), 00100011B ; Select AIN3
LD  (ADCCR2), 11011000B ; Select conversion time (312/fc) and operation mode
; AD CONVERT START
SET (ADCCR1). 7 ; ADRS = 1
SLOOP: TEST (ADCDR2). 5 ; EOCF = 1?
      JRS  T, SLOOP
; RESULT DATA READ
LD  A, (ADCDR1)
LD  (9FH), A

```

The analog input voltage is corresponded to the 8-bit digital value converted by the AD as shown in Figure 2-64.

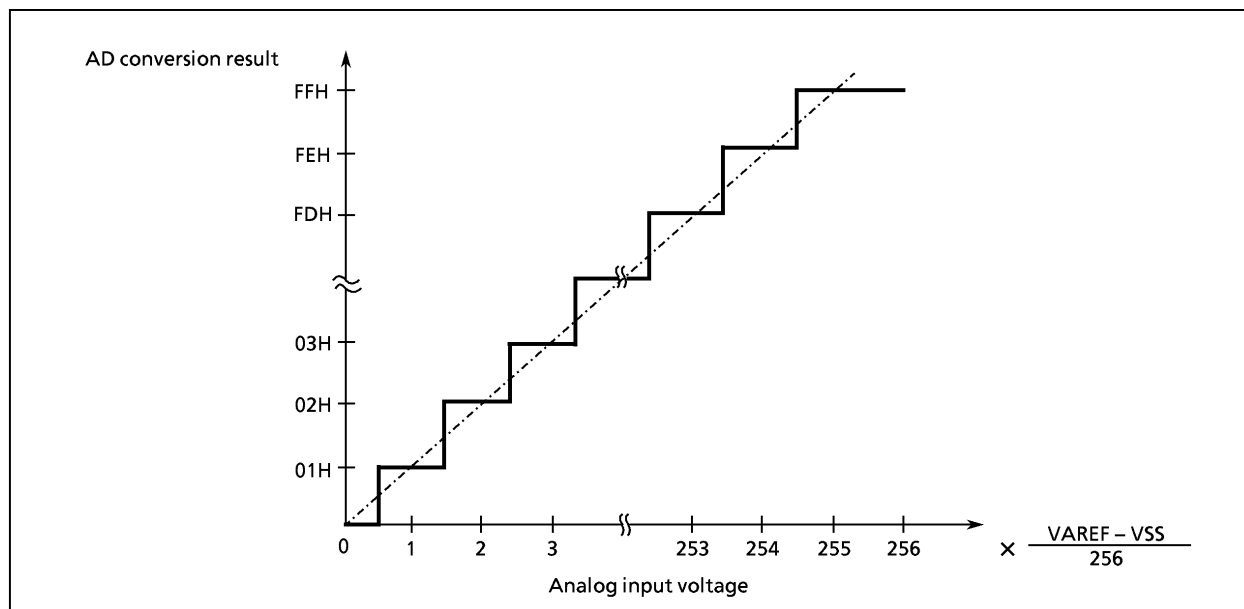


Figure 2-64. Analog Input Voltage and AD Conversion Result (typ.)

2.11.7 Precautions about AD Converter

(1) Analog input pin voltage range

Make sure the analog input pins (AIN0 to AIN7) are used at voltages within VSS below VAREF. If any voltage outside this range is applied to one of the analog input pins, the converted value on that pin becomes uncertain. The other analog input pins also are affected by that.

(2) Analog input shared pins

The analog input pins (AIN0 to AIN7) are shared with input/output ports. When using any of the analog inputs to execute AD conversion, do not execute input/output instructions for all other ports. This is necessary to prevent the accuracy of AD conversion from degrading. Not only these analog input shared pins, some other pins may also be affected by noise arising from input/output to and from adjacent pins.

(3) Noise countermeasure

The internal equivalent circuit of the analog input pins is shown in Figure 2-65. The higher the output impedance of the analog input source, more easily they are susceptible to noise. Therefore, make sure the output impedance of the signal source in your design is 5 k Ω or less. Toshiba also recommends attaching a capacitor external to the chip.

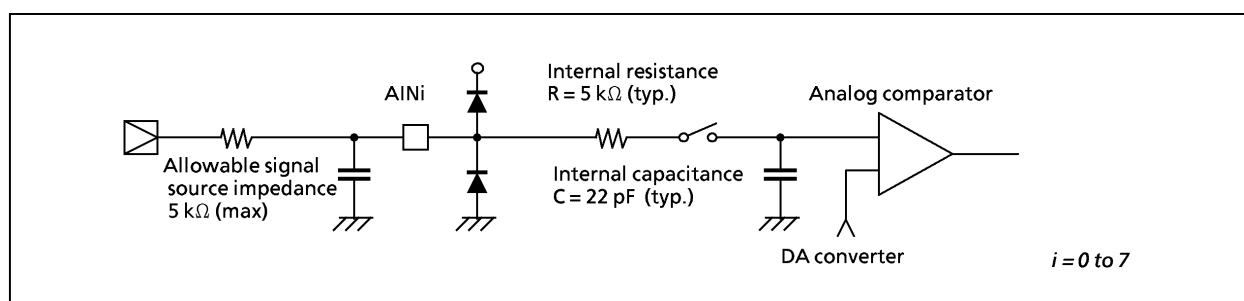


Figure 2-65. Analog Input Equivalent Circuit and Example of Input Pin Processing

2.12 Key-On Wake-Up (KWU)

In the TMP86CM25/S25, the STOP mode must be released by not only P20 ($\overline{\text{INT5/STOP}}$) pin but also P64 to P67 pins.

When the STOP mode is released by P64 to P67 pins, the P20 ($\overline{\text{INT5/STOP}}$) pin needs to be used.

2.12.1 Configuration

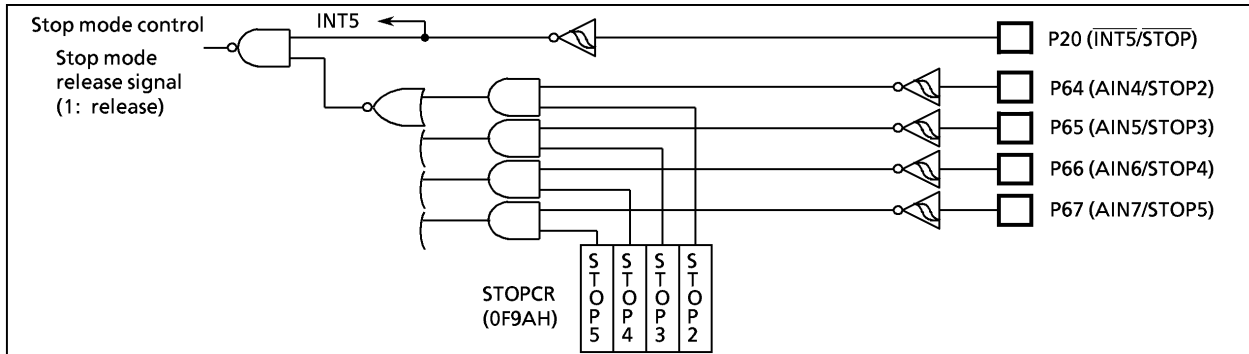
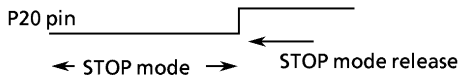


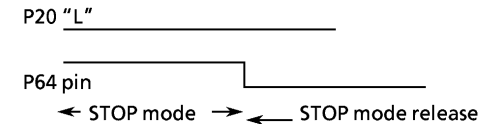
Figure 2-66. Stop Mode Control Circuit

Note: When STOP mode is released by P20 and P64 to P67, the most priority is P20.
When P64 to P67 pin used to release the STOP mode, then P20 must be input to "0" level.

a) at P20



b) at P64 to P67



2.12.2 Control

P64 to P67 (STOP2 to STOP5) pin can controlled by Key-on wake-up control register (STOPCR). It can be configured as enable/disable in one-bit unit. When those pins are used by STOP mode release, those pins must be set input mode (P6CR, P6DR, ADCCR1).

The STOP mode is set by SYSCR1 and release is the selected pin (STOPCR bit) to "0". When the STOP mode is released by inputting to P64 to P67 pins, the STOP mode (level mode) must be started by the system register 1 (SYSCR1). When the STOP mode, check the level that P64 to P67 pin is high. When the STOP mode release, P20 pin must be use.

Note: When the STOP mode release by edge mode, do not use STOP2 to STOP5 or must be set "1" level into STOP2 to STOP5 pins.

Key On Wake Up control register			
STOPCR (0F9AH)	7 6 5 4 3 2 1 0		
	STOP5 STOP4 STOP3 STOP2 - - - -	(Initial value: 0000 ****)	
STOP2	Stop mode released by P64 port	0: Disable 1: Enable	Write only
STOP3	Stop mode released by P65 port	0: Disable 1: Enable	
STOP4	Stop mode released by P66 port	0: Disable 1: Enable	
STOP5	Stop mode released by P67 port	0: Disable 1: Enable	

Figure 2-67. Key On Wake Up Control Register

2.13 LCD Driver

The TMP86CM25/S25 incorporates a driver to directly drive the liquid crystal display (LCD) and its control circuit. The connecting pins with the LCD are as shown below:

- (1) Segment output pin 40 pins (SEG39 to SEG0)
- (2) Segment output/ I/O port pin (shared) 20 pins (SEG59 to SEG40)
- (3) Common output pin 5 pins (COM4 to COM0)
- (4) Common output I/O port pin (shared) 11 pins (COM15 to COM5)

In addition, C0, C1, V1, V2, V3 and V4 are provided as the LCD drive booster circuit pins. The following three types of LCD can be driven directly:

- (1) 1/4 duty LCD: Maximum 240 pixels (60 segments x 4 digits)
- (2) 1/8 duty LCD: Maximum 480 pixels (60 segments x 8 digits)
- (3) 1/16 duty LCD: Maximum 960 pixels (60 segments x 16 digits)

2.13.1 Configuration of LCD Driver

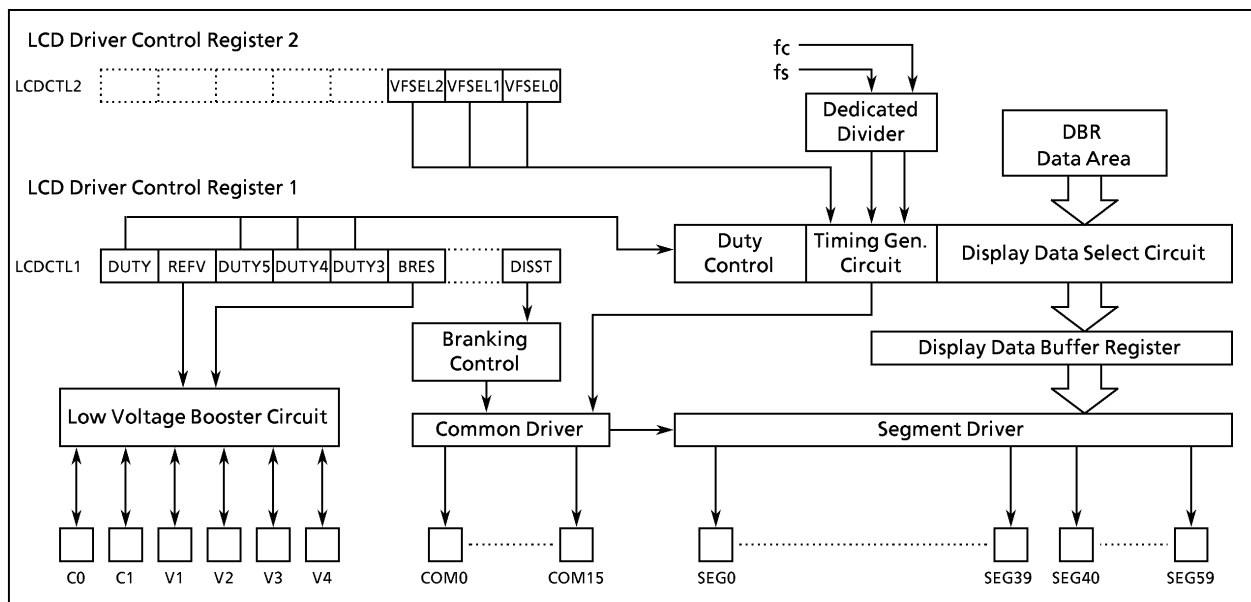


Figure 2-68. LCD Driver Block Diagram

Note: The LCD driver circuit has a built-in dedicated divider circuit. Thus, during use of the tool, LCD outputting is not stopped by debugger break processing.

2.13.2 Controlling LCD Driver

The LCD driver is controlled by the LCD control register 1 (LCDCTL1) and the LCD control register 2 (LCDCTL2). The display of the LCD driver is enabled by DISST.

LCD Control Register 1								
LCDCTL1 (0027H)								
7	6	5	4	3	2	1	0	(Initial value: 0000 00*0)
DUTY7	REFV	DUTY5	DUTY4	DUTY3	BRES		DISST	
DUTY7 DUTY5 DUTY4 DUTY3	Select duty.	0***: Reserved 1000: 1/4 duty 1001: Reserved 1010: 1/8 duty 1011: Reserved 1100: Reserved 1101: Reserved 1110: 1/16 duty 1111: Reserved					R/W	
REFV	Sets LCD reference voltage.	0: $V4 \leq VDD$ 1: $VDD < V4 \leq 5.5\text{ V}$						
BRES	Sets booster circuit.	0: Booster circuit disable (Note 4) 1: Booster circuit enable						
DISST	Controls LCD display.	0: LCD display blanking 1: LCD display enable						
<p>Note 1: After reset, <DUTYs> are set to "0000" (Initial value: reserved). Set the duty as appropriate for LCD panel.</p> <p>Note 2: Switch <REFV> according to VDD. If it is not set appropriately, an overcurrent may flow causing damage to the device. Caution is especially required when VDD is battery-driven.</p> <p>Note 3: If <DISST> is set to "0" (LCD display blanking), all SEG/COM pins become VSS level.</p> <p>Note 4: When <REFV> for the LCD reference voltage <REFV> is set to "0", always make sure the reference power supply is entered from the V4 pin.</p> <p>Note 5: Reserved: Not to be set.</p>								
LCD Control Register 2								
LCDCTL2 (0028H)								
7	6	5	4	3	2	1	0	(Initial value: **** *011)
					VFSEL2	VFSEL1	VFSEL0	
VFSEL	Selects base frequency for frame frequency.	000: $f_c/2^9$ (at 16 MHz) 001: $f_c/2^8$ (at 8 MHz) 010: $f_c/2^7$ (at 4 MHz) 011: $f_c/2^6$ (at 2 MHz) 1** : fs (at 32.768 kHz)					R/W	
<p>Note 1: Set the LCD control register 2 according to operating frequency. For details of the actual frame frequency, see Table 2-9.</p>								

Figure 2-69. LCD Driver Control Register

(1) Frame frequency

The frame frequency is set depending on the driving method and the base frequency as shown in Table 2-18.

The base frequency is selected with LCDCTL2 <VFSEL> depending on the basic clock frequencies f_c and f_s to be used.

Table 2-18. Frame Frequency Settings

VFSEL	Base frequency [Hz]	Frame frequency [Hz]		
		1/4 duty	1/8 duty	1/16 duty
000	$\frac{f_c}{2^9}$	$\frac{f_c}{2^9 \cdot 84 \cdot 4}$	$\frac{f_c}{2^9 \cdot 42 \cdot 8}$	$\frac{f_c}{2^9 \cdot 21 \cdot 16}$
	($f_c = 16$ MHz)	93	93	93
001	$\frac{f_c}{2^8}$	$\frac{f_c}{2^8 \cdot 84 \cdot 4}$	$\frac{f_c}{2^8 \cdot 42 \cdot 8}$	$\frac{f_c}{2^8 \cdot 21 \cdot 16}$
	($f_c = 8$ MHz)	93	93	93
010	$\frac{f_c}{2^7}$	$\frac{f_c}{2^7 \cdot 84 \cdot 4}$	$\frac{f_c}{2^7 \cdot 42 \cdot 8}$	$\frac{f_c}{2^7 \cdot 21 \cdot 16}$
	($f_c = 4$ MHz)	93	93	93
011	$\frac{f_c}{2^6}$	$\frac{f_c}{2^6 \cdot 84 \cdot 4}$	$\frac{f_c}{2^6 \cdot 42 \cdot 8}$	$\frac{f_c}{2^6 \cdot 21 \cdot 16}$
	($f_c = 2$ MHz)	93	93	93
1**	f_s	$\frac{f_s}{84 \cdot 4}$	$\frac{f_s}{42 \cdot 8}$	$\frac{f_s}{21 \cdot 16}$
	($f_s = 32.768$ kHz)	97.5	97.5	97.5

Note 1: f_c : High-frequency clock frequency [Hz], f_s : Low-frequency clock frequency [Hz]

Note 2: Although this product is guaranteed to operate at $f_c = 1.32$ [MHz] or less is not recommended for LCD display as the frame frequency becomes 61 [Hz] or less.

2.13.3 LCD Booster Circuit

(1) LCD booster circuit

The TMP86CM25/S25 can boost (divide) the externally-supplied reference voltage using the built-in booster circuit as a power supply for driving the LCD. When V1 pin is the reference voltage, the inputted reference voltage is boosted by two times (V2), 3 times (V3) and 4 times (V4) to generate a voltage for a segment/common signal. When V2 pin is the reference voltage, the inputted reference voltage is divided/boosted by 1/2 time (V1), 3/2 times (V3) and two times (V4). Likewise, when V3 pin or V4 pin is the reference, the inputted reference voltage is boosted/divided and the voltage ratio is $V1 \times 4 = V2 \times 2 = V3 \times (4/3) = V4$. As this circuit uses a 4-times boosting method, the bias ratio is 1/4 only.

Note: When the reference pin is other than V1 pin, a condenser is required between V1 pin and GND.

2.13.4 Methods of Connecting LCD Booster Circuit

(1) Method of connecting booster circuit by using a regulator

If VDD is not stable because it is battery-driven, etc., we recommend a connection method using a regulator as shown below in order to preserve the quality of display.

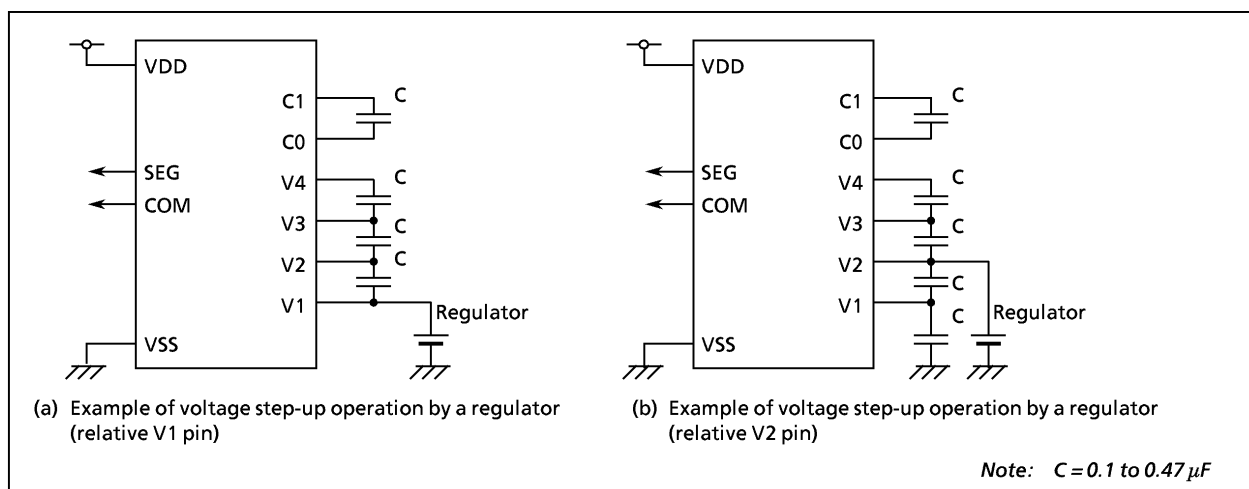


Figure 2-70. Method of Connecting Booster Circuit by Using a Regulator

Note: For use with $VDD \geq V4$ ($LCDCTL1 < REFV > = 0$), always make sure the reference power supply is entered from V4.

(2) Method of connecting booster circuit without using a regulator

If stable VDD supply is achieved ($VDD \geq V4$), the booster circuit can be connected without using a regulator as shown below. In this case, set LCDCTL1 <REFV> to "0" and make sure the reference power supply is entered from the V4 pin.

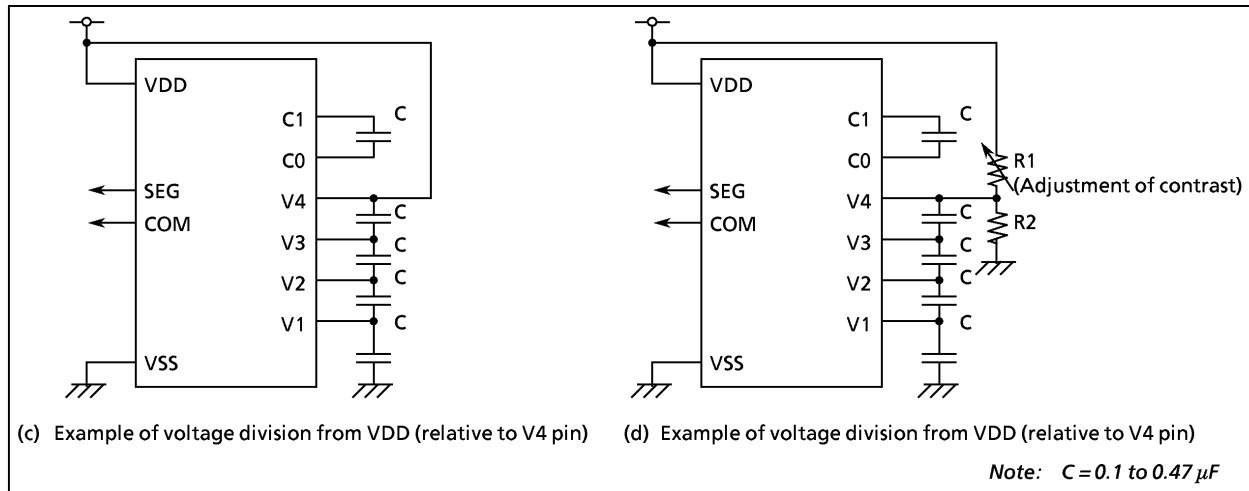


Figure 2-71. Method of Connecting Booster Circuit without Using a Regulator

2.13.5 LCD Display Operation

(1) Setting display data

Display data is stored in display data area (128 bytes in addresses 0F00H to 0F7FH) provided in DBR. Display data stored in the display data area is automatically read by hardware and sent to the LCD driver. The LCD driver generates segment and common signals according to display data and the driving method. Thus, display patterns can be changed simply by rewriting the contents of display data area in the program.

Figure 2-72 shows the correspondence between display data areas and SEG/COM pins. The light comes on when display data is "1" and it goes out when "0". Because the number of pixels that can be driven varies with the method of driving the LCD, the number of bytes in the display data area used to store display data also varies. Thus, bytes not used to store display data and data memory corresponding to addresses not connected to the LCD can be used for storing generally processed data. (See Table 2-19)

Note: Because the contents of display data area become unstable at powering on, execute the initialize routine for the initial setting.

0F00H	0F10H	0F20H	0F30H	0F40H	0F50H	0F60H	0F70H	COM0
0F01H	0F11H	0F21H	0F31H	0F41H	0F51H	0F61H	0F71H	COM1
0F02H	0F12H	0F22H	0F32H	0F42H	0F52H	0F62H	0F72H	COM2
0F03H	0F13H	0F23H	0F33H	0F43H	0F53H	0F63H	0F73H	COM3
0F04H	0F14H	0F24H	0F34H	0F44H	0F54H	0F64H	0F74H	COM4
0F05H	0F15H	0F25H	0F35H	0F45H	0F55H	0F65H	0F75H	COM5
0F06H	0F16H	0F26H	0F36H	0F46H	0F56H	0F66H	0F76H	COM6
0F07H	0F17H	0F27H	0F37H	0F47H	0F57H	0F67H	0F77H	COM7
0F08H	0F18H	0F28H	0F38H	0F48H	0F58H	0F68H	0F78H	COM8
0F09H	0F19H	0F29H	0F39H	0F49H	0F59H	0F69H	0F79H	COM9
0F0AH	0F1AH	0F2AH	0F3AH	0F4AH	0F5AH	0F6AH	0F7AH	COM10
0F0BH	0F1BH	0F2BH	0F3BH	0F4BH	0F5BH	0F6BH	0F7BH	COM11
0F0CH	0F1CH	0F2CH	0F3CH	0F4CH	0F5CH	0F6CH	0F7CH	COM12
0F0DH	0F1DH	0F2DH	0F3DH	0F4DH	0F5DH	0F6DH	0F7DH	COM13
0F0EH	0F1EH	0F2EH	0F3EH	0F4EH	0F5EH	0F6EH	0F7EH	COM14
0F0FH	0F1FH	0F2FH	0F3FH	0F4FH	0F5FH	0F6FH	0F7FH	COM15

SEG7	SEG15	SEG23	SEG31	SEG39	SEG47	SEG55	SEG59
to	to	to	to	to	to	to	to
SEG0	SEG8	SEG16	SEG24	SEG32	SEG40	SEG48	SEG56

Figure 2-72. LCD Display Data Area (DBR)

Table 2-19. Areas Used to Store Display Data

Driving Method	COM number to be used
1/16 Duty	COM15 to COM0
1/8 Duty	COM7 to COM0
1/4 Duty	COM3 to COM0

(2) Blanking

The LCD display can be blanked by clearing DISST to “0”. Blanking extinguishes the LCD by outputting GND level to COM/SEG pins.

If the STOP mode is entered while the LCD display is on, DISST is cleared to “0” and blanking is performed automatically. If the STOP mode is then reverted, DISST is set to “1” and display is resumed automatically.

Note: At reset, the segment dedicated pins (SEG39 to SEG0) and common output becomes GND level, whereas the I/O port/segment shared pins (P1, P3, P5 ports) output, the I/O port/common shared pins (P3, P7 ports) output become the high-impedance state. Thus, if an external reset input lasts for a significant length of time, it may affect the LCD display such as blurring.

2.13.6 Method of Controlling LCD Driver

(1) Initial setting

The procedure of initial setting is shown below.

Example: When 60 seg × 8 com, 1/8 duty, 5 V-system LCD operates with $f_c = 8$ MHz (at $V_{DD} = 5$ V)

```
LD (LCDCTL1),10010100B ; 1/8 duty, LCD reference voltage ( $V_{DD} = V_4$ ), booster
                        ; circuit enable set
LD (P1LCR),0FFH ; Set P1 port for segment output
LD (P3LCR),0FFH ; Set P3 port for segment/common output
LD (P5LCR),0FFH ; Set P5 port for segment output
;
LD (LCDCTL1),10010101B ; LCD display enable set
```

(2) Storing display data

Display data is normally prepared as fixed data in the program memory (ROM) and stored in the display data area by a load instruction.

Example 1: Corresponding to the connection and display using a 1/8 duty LCD shown in Figure 2-73, the Table 2-20 shows display data and Figure 2-74 shows display timing.

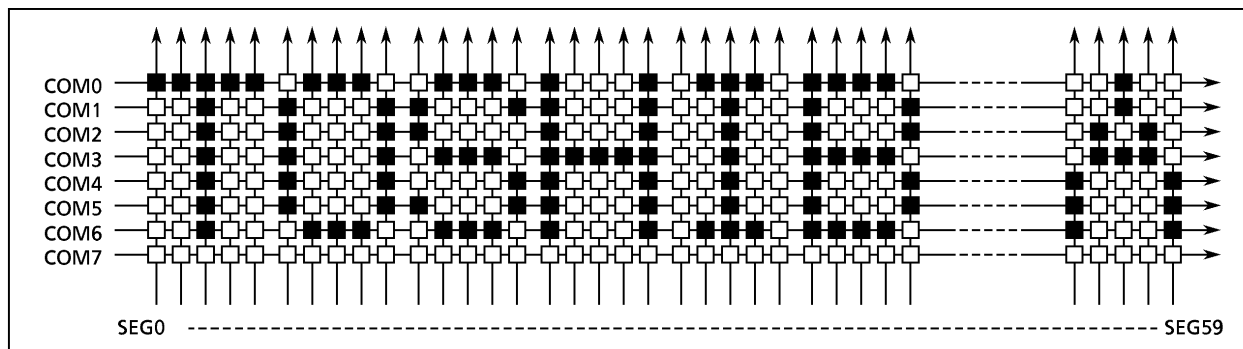


Figure 2-73. Example of Display Data (1/8 duty)

Table 2-20. Example of Display Data (1/8 duty)

	DBR	SEG	SEG	SEG	SEG	SEG	SEG	SEG	HEX	DBR	SEG	SEG	SEG	SEG	SEG	SEG	SEG	HEX	...		
		0	1	2	3	4	5	6			7	8	9	10	11	12	13			14	15
		Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7		Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7			
COM0	0F00H	1	1	1	1	1	0	1	1	DF	0F10H	1	0	0	1	1	1	0	1	B9	...
COM1	0F01H	0	0	1	0	0	1	0	0	24	0F11H	0	1	1	0	0	0	1	1	C6	...
COM2	0F02H	0	0	1	0	0	1	0	0	24	0F12H	0	1	1	0	0	0	0	1	86	...
COM3	0F03H	0	0	1	0	0	1	0	0	24	0F13H	0	1	0	1	1	1	0	1	BA	...
COM4	0F04H	0	0	1	0	0	1	0	0	24	0F14H	0	1	0	0	0	0	1	1	C2	...
COM5	0F05H	0	0	1	0	0	1	0	0	24	0F15H	0	1	1	0	0	0	1	1	C6	...
COM6	0F06H	0	0	1	0	0	0	1	1	C4	0F16H	1	0	0	1	1	1	0	1	B9	...
COM7	0F07H	0	0	0	0	0	0	0	0	00	0F17H	0	0	0	0	0	0	0	0	00	...

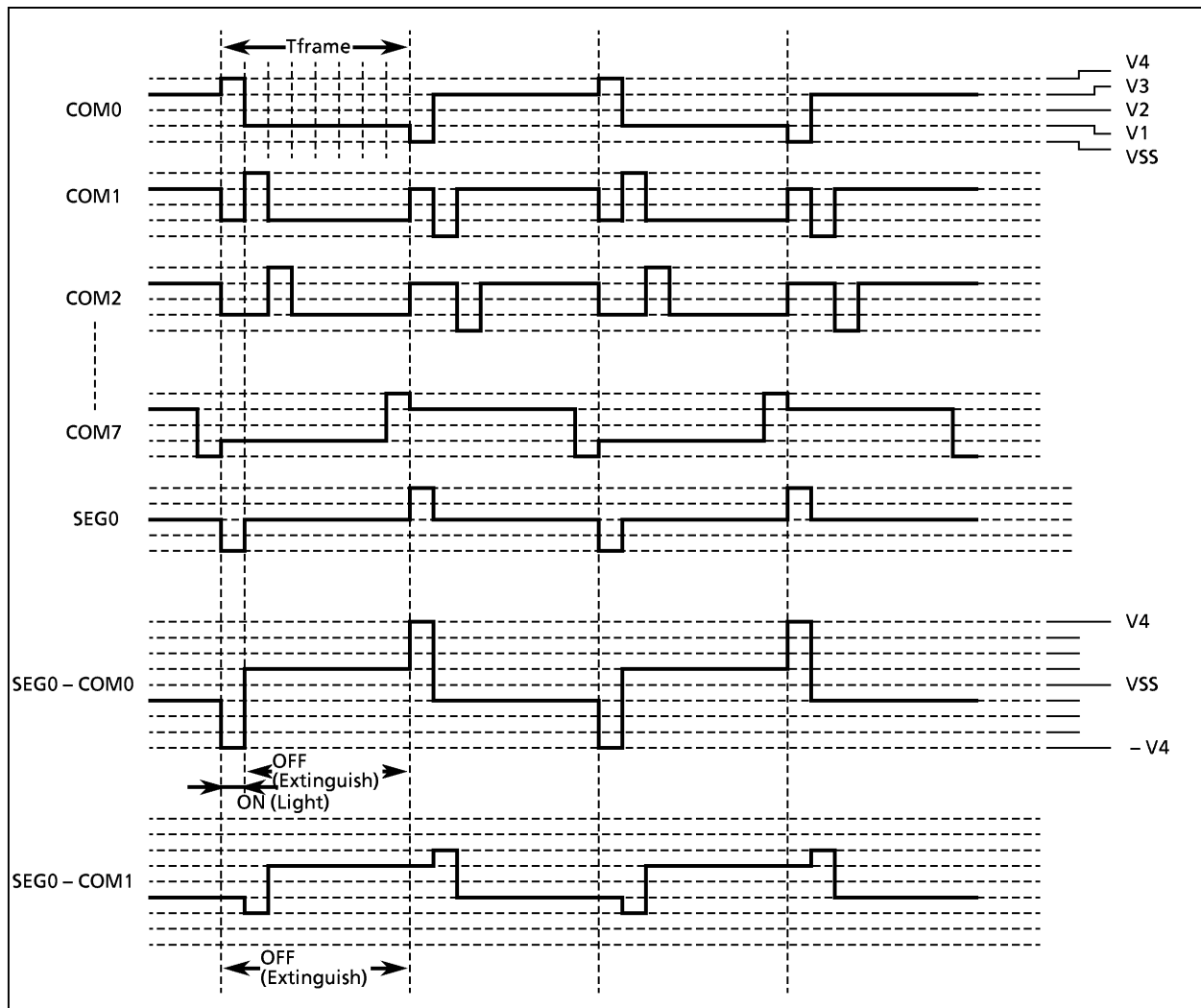


Figure 2-74. Example of Display Timing (1/8 duty)

Example 2: Corresponding to the connection and display using a 1/16 duty LCD shown in Figure 2-75, Table 2-21 shows display data and Figure 2-76 shows display timing.

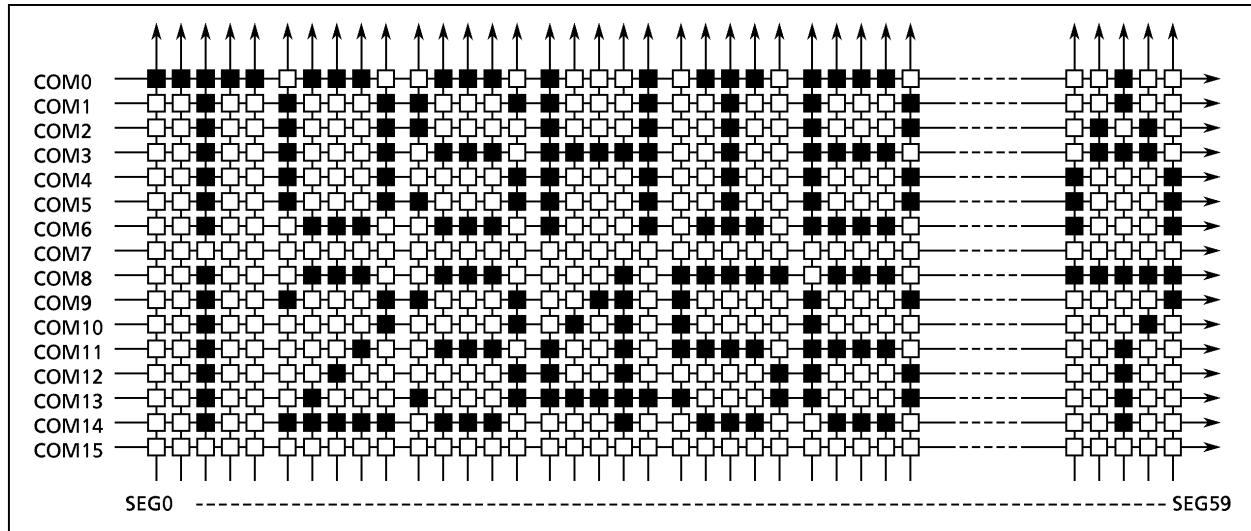


Figure 2-75. Example of Display Data (1/16 duty)

Table 2-21. Example of Display Data (1/16 duty)

	DBR	SEG	SEG	SEG	SEG	SEG	SEG	SEG	SEG	HEX	DBR	SEG	SEG	SEG	SEG	SEG	SEG	SEG	HEX		
		0	1	2	3	4	5	6	7			8	9	10	11	12	13	14		15	
		Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7			Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7		
COM0	0F00H	1	1	1	1	1	0	1	1	DF	0F10H	1	0	0	1	1	1	0	1	B9	...
COM1	0F01H	0	0	1	0	0	1	0	0	24	0F11H	0	1	1	0	0	0	1	1	C6	...
COM2	0F02H	0	0	1	0	0	1	0	0	24	0F12H	0	1	1	0	0	0	0	1	86	...
COM3	0F03H	0	0	1	0	0	1	0	0	24	0F13H	0	1	0	1	1	1	0	1	BA	...
COM4	0F04H	0	0	1	0	0	1	0	0	24	0F14H	0	1	0	0	0	0	1	1	C2	...
COM5	0F05H	0	0	1	0	0	1	0	0	24	0F15H	0	1	1	0	0	0	1	1	C6	...
COM6	0F06H	0	0	1	0	0	0	1	1	C4	0F16H	1	0	0	1	1	1	0	1	B9	...
COM7	0F07H	0	0	0	0	0	0	0	0	00	0F17H	0	0	0	0	0	0	0	0	00	...
COM8	0F08H	0	0	1	0	0	0	1	1	C4	0F18H	1	0	0	1	1	1	0	0	39	...
COM9	0F09H	0	0	1	0	0	1	0	0	24	0F19H	0	1	1	0	0	0	1	0	46	...
COM10	0F0AH	0	0	1	0	0	0	0	0	04	0F1AH	0	1	0	0	0	0	1	0	42	...
COM11	0F0BH	0	0	1	0	0	0	0	0	04	0F1BH	1	0	0	1	1	1	0	1	B9	...
COM12	0F0CH	0	0	1	0	0	0	0	1	84	0F1CH	0	0	0	0	0	0	1	1	C0	...
COM13	0F0DH	0	0	1	0	0	0	1	0	44	0F1DH	0	0	1	0	0	0	1	1	C4	...
COM14	0F0EH	0	0	1	0	0	1	1	1	E4	0F1EH	1	1	0	1	1	1	0	0	3B	...
COM15	0F0FH	0	0	0	0	0	0	0	0	00	0F1FH	0	0	0	0	0	0	0	0	00	...

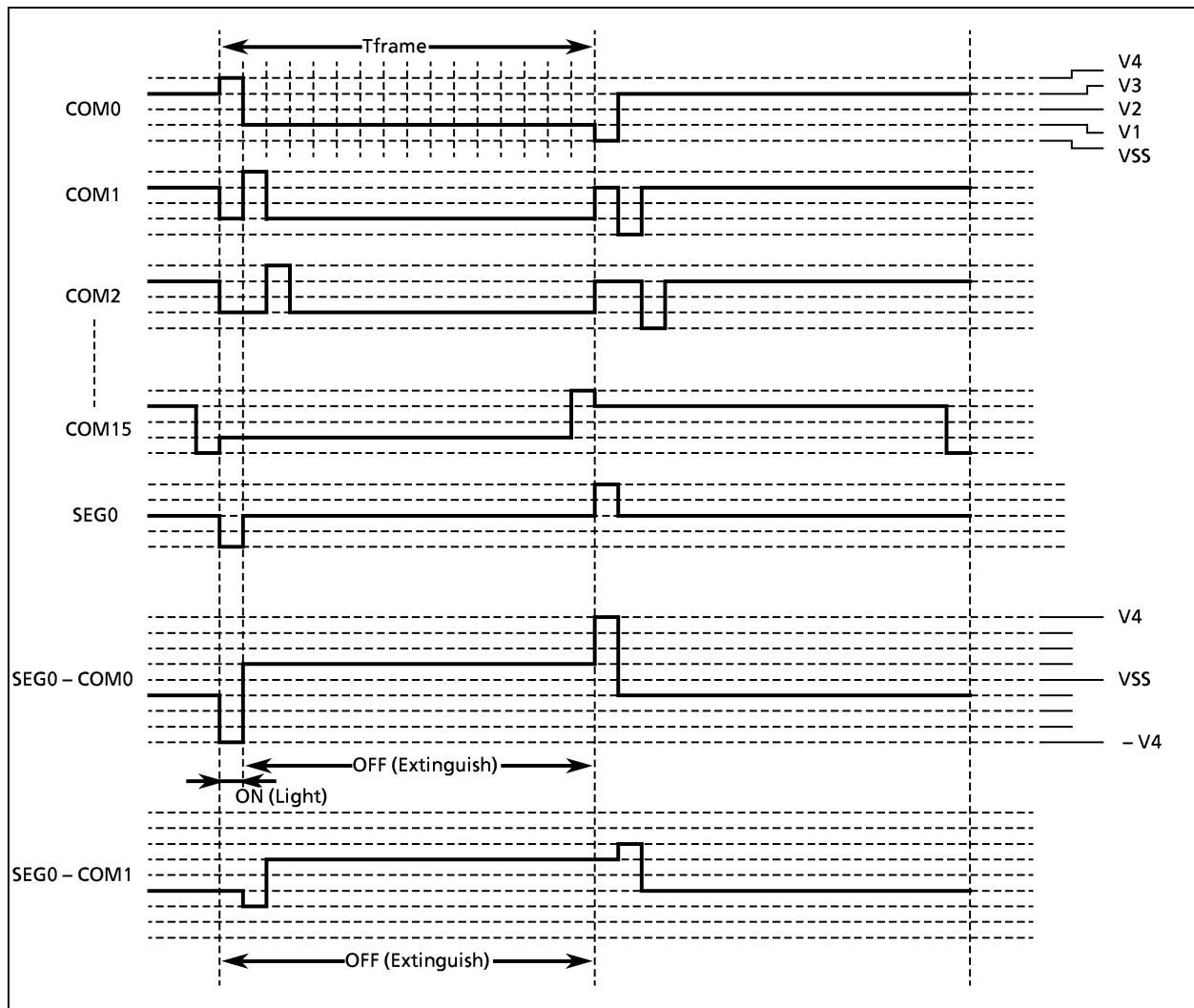


Figure 2-76. Example of Display Timing (1/16 duty)

Input/Output Circuitry

(1) Control Pins

The input/output circuitries of the TMP86CM25/S25 control pins are shown below.

Control Pin	I/O	Input/Output Circuitry	Remarks
XIN XOUT	Input Output		Resonator connecting pins (high-frequency) $R_f = 1.2\text{ M}\Omega$ (typ.) $R_o = 1.0\text{ k}\Omega$ (typ.)
XTIN XTOUT	Input Output	<p>NORMAL1 mode</p> <p>Refer to port P2</p> <p>NORMAL2 mode</p>	Resonator connecting pins (Low-frequency) $R_f = 6\text{ M}\Omega$ (typ.) $R_o = 220\text{ k}\Omega$ (typ.)
$\overline{\text{RESET}}$	I/O		Sink open drain output Hysteresis input Pull-up resistor $R_{IN} = 220\text{ k}\Omega$ (typ.)
$\overline{\text{STOP/INT5}}$	Input		Hysteresis input
TEST	Input		Pull-down resistor $R_{IN} = 70\text{ k}\Omega$ (typ.)

Note: The TEST pin of the TMP86PS25 does not have a pull-down resistor. Fix the test pin at low-level.

(2) Input/Output Ports

Port	I/O	Input/Output Circuitry	Remarks
P1 P7	I/O	<p>Initial "High-Z" P1LCR/P7LCR</p> <p>SEG output</p> <p>Data output</p> <p>Input from output latch</p> <p>Pin input</p>	Sink open drain output Hysteresis input
P5	I/O	<p>Initial "High-Z" P5 LCR</p> <p>SEG output</p> <p>Data output</p> <p>Input from output latch</p> <p>Pin input</p>	Sink open drain output
P2	I/O	<p>Initial "High-Z"</p> <p>Data output</p> <p>Input from output latch</p> <p>Pin input</p>	Sink open drain output Hysteresis input
P30 P31 P32 P33	I/O	<p>Initial "High-Z" P3LCR</p> <p>SEG output</p> <p>Data output</p> <p>Input from output latch</p> <p>Pin input</p>	Sink open drain output Hysteresis input High current output (Nch)
P34 P35 P36	I/O	<p>Initial "High-Z" P3LCR</p> <p>COM output</p> <p>Data output</p> <p>Input from output latch</p> <p>Pin input</p>	Sink open drain output Hysteresis input
P6	I/O	<p>Initial "High-Z"</p> <p>Data output</p> <p>Disable</p> <p>Pin input</p>	Tri-state I/O Hysteresis input

Note: Port P1, P3, P5 and P7 are sink open drain output. But they are also used as a segment output of LCD. Therefore, absolute maximum ratings of port input voltage should be used in -0.3 to $V_{DD} + 0.3$ volts

Electrical Characteristics

Absolute Maximum Ratings	($V_{SS} = 0\text{ V}$)
--------------------------	---------------------------

Parameter	Symbol	Pins	Rating	Unit
Supply Voltage	V_{DD}		- 0.3 to 6.5	V
Input Voltage	V_{IN}		- 0.3 to $V_{DD} + 0.3$	
Output Voltage	V_{OUT1}		- 0.3 to $V_{DD} + 0.3$	
Output Current (Per 1 pin)	I_{OUT1}	P6 Port	- 1.8	mA
	I_{OUT2}	P1, P2, P34 to P36, P5, P6, P7 Port	3.2	
	I_{OUT3}	P30 to P33 Port	30	
Output Current (Total)	ΣI_{OUT2}	P1, P2, P34 to P36, P5, P6, P7 Port	60	
	ΣI_{OUT3}	P30 to P33 Port	80	
Power Dissipation [$T_{opr} = 85^\circ\text{C}$]	PD		350	mW
Soldering Temperature (time)	Tsld		260 (10 s)	$^\circ\text{C}$
Storage Temperature	Tstg		- 55 to 125	
Operating Temperature	Topr		- 40 to 85	

Note: The absolute maximum ratings are rated values which must not be exceeded during operation, even for an instant. Any one of the ratings must not be exceeded. If any absolute maximum rating is exceeded, a device may break down or its performance may be degraded, causing it to catch fire or explode resulting in injury to the user. Thus, when designing products which include this device, ensure that no absolute maximum rating value will ever be exceeded.

Recommended Operating Condition	(V _{SS} = 0 V, T _{opr} = -40 to 85°C)
---------------------------------	---

Parameter	Symbol	Pins	Condition	Min	Max	Unit	
Supply Voltage	V _{DD}		f _c = 16 MHz	NORMAL1, 2 mode	4.5	5.5	V
				IDLE0, 1, 2 mode			
			f _c = 8 MHz	NORMAL1, 2 mode	2.7		
				IDLE0, 1, 2 mode			
			f _c = 4.2 MHz	NORMAL1, 2 mode	1.8		
IDLE0, 1, 2 mode							
f _s = 32.768 kHz	SLOW1, 2 mode	1.8					
	SLEEP0, 1, 2 mode						
			STOP mode				
Input high Level	V _{IH1}	Except Hysteresis input	V _{DD} ≥ 4.5 V	V _{DD} × 0.70	V _{DD}	V	
	V _{IH2}	Hysteresis input		V _{DD} × 0.75			
	V _{IH3}			V _{DD} < 4.5 V			V _{DD} × 0.90
Input low Level	V _{IL1}	Except Hysteresis input	V _{DD} ≥ 4.5 V	0	V _{DD} × 0.30	V	
	V _{IL2}	Hysteresis input			V _{DD} × 0.25		
	V _{IL3}				V _{DD} < 4.5 V		V _{DD} × 0.10
LCD reference voltage range	V _{1IN}	V1	LCDCTL1 < REFV > = "1"	1.0	1.375	V	
	V _{2IN}	V2		2.0	2.750		
	V _{3IN}	V3		V _{DD} < V4 (Note 2)	3.0		4.125
	V _{4IN}	V4		4.0	5.500		
	V _{4IN}	V4 (Note 3)		LCDCTL1 < REFV > = "0"	–		V _{DD}
Clock Frequency	f _c	XIN, XOUT	V _{DD} = 1.8 to 5.5 V	1.0	4.2	MHz	
			V _{DD} = 2.7 to 5.5 V		8.0		
			V _{DD} = 4.5 to 5.5 V		16.0		
	f _s	XTIN, XTOUT		30.0	34.0	kHz	

Note 1: The recommended operating conditions for a device are operating conditions under which it can be guaranteed that the device will operate as specified. If the device is used under operating conditions other than the recommended operating conditions (supply voltage, operating temperature range, specified AC/DC values etc.), malfunction may occur. Thus, when designing products which include this device, ensure that the recommended operating conditions for the device are always adhered to.

Note 2: When LCDCTL1 < REFV > is set to "1", always keep the condition of V_{DD} < V4.

Note 3: When LCDCTL1 < REFV > is set to "0", always supply the reference voltage from V4 pin.

DC Characteristics

 $(V_{SS} = 0 \text{ V}, T_{opr} = -40 \text{ to } 85^\circ\text{C})$

Parameter	Symbol	Pins	Condition	Min	Typ.	Max	Unit		
Hysteresis Voltage	V_{HS}	Hysteresis input		–	0.9	–	V		
Input Current	I_{IN1}	TEST	$V_{DD} = 5.5 \text{ V}, V_{IN} = 5.5 \text{ V}/0 \text{ V}$	–	–	± 2	μA		
	I_{IN2}	Sink Open Drain, Tri-state							
	I_{IN3}	RESET, STOP							
Input Resistance	R_{IN1}	TEST Pull-Down		–	70	–	$\text{k}\Omega$		
	R_{IN2}	RESET Pull-Up		100	220	450			
Output Leakage Current	I_{LO}	Sink Open Drain, Tri-state	$V_{DD} = 5.5 \text{ V}, V_{OUT} = 5.5 \text{ V}/0 \text{ V}$	–	–	± 2	μA		
Output High Voltage	V_{OH2}	Tri-st Port	$V_{DD} = 4.5 \text{ V}, I_{OH} = -0.7 \text{ mA}$	4.1	–	–	V		
Output Low Voltage	V_{OL}	Except XOUT and P30 to P33 Port	$V_{DD} = 4.5 \text{ V}, I_{OL} = 1.6 \text{ mA}$	–	–	0.4			
Output Low Current	I_{OL}	High Current Port (P30 to P33 Port)	$V_{DD} = 4.5 \text{ V}, V_{OL} = 1.0 \text{ V}$	–	20	–	mA		
Supply Current in NORMAL 1, 2 mode	I_{DD}		$V_{DD} = 5.5 \text{ V}$ $V_{IN} = 5.3/0.2 \text{ V}$ $f_c = 16 \text{ MHz}$ $f_s = 32.768 \text{ kHz}$	TMP86CM25	–	4.5		7.0	
Supply Current in IDLE 0, 1, 2 mode					–	2.7		3.5	
Supply Current in SLOW 1 mode			$V_{DD} = 3.0 \text{ V}$ $V_{IN} = 2.8 \text{ V}/0.2 \text{ V}$ $f_s = 32.768 \text{ kHz}$ LCD driver is not enable.		–	6.0		25	
Supply Current in SLEEP 1 mode					–	4.0		15	
Supply Current in SLEEP 0 mode			$V_{DD} = 5.5 \text{ V}$ $V_{IN} = 5.3 \text{ V}/0.2 \text{ V}$		–	2.5		13	
Supply Current in STOP mode					–	0.5		10	
Supply Current in NORMAL 1, 2 mode			$V_{DD} = 5.5 \text{ V}$ $V_{IN} = 5.3/0.2 \text{ V}$ $f_c = 16 \text{ MHz}$ $f_s = 32.768 \text{ kHz}$		TMP86CS25	–		6.0	7.0
Supply Current in IDLE 0, 1, 2 mode						–		4.2	5.0
Supply Current in SLOW 1 mode			$V_{DD} = 3.0 \text{ V}$ $V_{IN} = 2.8 \text{ V}/0.2 \text{ V}$ $f_s = 32.768 \text{ kHz}$ LCD driver is not enable.		–	8.5		25	
Supply Current in SLEEP 1 mode					–	5.0		15	
Supply Current in SLEEP 0 mode			$V_{DD} = 5.5 \text{ V}$ $V_{IN} = 5.3 \text{ V}/0.2 \text{ V}$		–	3.0		13	
Supply Current in STOP mode					–	0.5		10	

Note 1: Typical values show those at $T_{opr} = 25^\circ\text{C}$, $V_{DD} = 5 \text{ V}$

Note 2: Input current (I_{IN1} , I_{IN2}): The current through pull-up or pull-down resistor is not included.

Note 3: I_{DD} does not include I_{REF} current.

Note 4: The supply currents of SLOW 2 and SLEEP 2 modes are equivalent to IDLE 0, 1, 2.

AD Conversion Characteristics

($V_{SS} = 0.0 \text{ V}$, $4.5 \text{ V} \leq V_{DD} \leq 5.5 \text{ V}$, $T_{opr} = -40 \text{ to } 85^\circ\text{C}$)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Analog Reference Voltage	V_{AREF}		$V_{DD} - 1.5$	–	V_{DD}	V
Analog Reference Voltage Range (Note 4)	ΔV_{AREF}		3.0	–	–	
Analog Input Voltage	V_{AIN}		V_{SS}	–	V_{AREF}	
Power Supply Current of Analog Reference Voltage	I_{REF}	$V_{DD} = V_{AREF} = 5.5 \text{ V}$ $V_{SS} = 0.0 \text{ V}$	–	0.6	1.0	mA
Non linearity Error		$V_{DD} = 5.0 \text{ V}$, $V_{SS} = 0.0 \text{ V}$ $V_{AREF} = 5.0 \text{ V}$	–	–	± 1	LSB
Zero Point Error			–	–	± 1	
Full Scale Error			–	–	± 1	
Total Error			–	–	± 2	

($V_{SS} = 0.0 \text{ V}$, $2.7 \text{ V} \leq V_{DD} < 4.5 \text{ V}$, $T_{opr} = -40 \text{ to } 85^\circ\text{C}$)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Analog Reference Voltage	V_{AREF}		$V_{DD} - 1.5$	–	V_{DD}	V
Analog Reference Voltage Range (Note 4)	ΔV_{AREF}		2.5	–	–	
Analog Input Voltage	V_{AIN}		V_{SS}	–	V_{AREF}	
Power Supply Current of Analog Reference Voltage	I_{REF}	$V_{DD} = V_{AREF} = 4.5 \text{ V}$ $V_{SS} = 0.0 \text{ V}$	–	0.5	0.8	mA
Non linearity Error		$V_{DD} = 2.7 \text{ V}$, $V_{SS} = 0.0 \text{ V}$ $V_{AREF} = 2.7 \text{ V}$	–	–	± 1	LSB
Zero Point Error			–	–	± 1	
Full Scale Error			–	–	± 1	
Total Error			–	–	± 2	

($V_{SS} = 0.0 \text{ V}$, $2.0 \text{ V} \leq V_{DD} < 2.7 \text{ V}$, $T_{opr} = -40 \text{ to } 85^\circ\text{C}$) Note 5
($V_{SS} = 0.0 \text{ V}$, $1.8 \text{ V} \leq V_{DD} < 2.0 \text{ V}$, $T_{opr} = -10 \text{ to } 85^\circ\text{C}$) Note 5

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Analog Reference Voltage	V_{AREF}		$V_{DD} - 0.9$	–	V_{DD}	V
Analog Reference Voltage Range (Note 4)	ΔV_{AREF}	$1.8 \text{ V} \leq V_{DD} < 2.0 \text{ V}$	1.8	–	–	
		$2.0 \text{ V} \leq V_{DD} < 2.7 \text{ V}$	2.0	–	–	
Analog Input Voltage	V_{AIN}		V_{SS}	–	V_{AREF}	
Power Supply Current of Analog Reference Voltage	I_{REF}	$V_{DD} = V_{AREF} = 2.7 \text{ V}$ $V_{SS} = 0.0 \text{ V}$	–	0.3	0.5	mA
Non linearity Error		$V_{DD} = 1.8 \text{ V}$, $V_{SS} = 0.0 \text{ V}$ $V_{AREF} = 1.8 \text{ V}$	–	–	± 2	LSB
Zero Point Error			–	–	± 2	
Full Scale Error			–	–	± 2	
Total Error			–	–	± 4	

Note 1: The total error includes all errors except a quantization error, and is defined as maximum deviation from the ideal conversion line.

Note 2: Conversion time is different in recommended value by power supply voltage. About conversion time, please refer to "2.11.2 Register Configuration".

Note 3: Please use input voltage to AIN input Pin in limit of $V_{AREF} - V_{SS}$. When voltage of range outside is input, conversion value becomes unsettled and gives affect to other channel conversion value.

Note 4: Analog Reference Voltage Range: $\Delta V_{AREF} = V_{AREF} - V_{SS}$

Note 5: When AD is used with $V_{DD} < 2.7 \text{ V}$, the guaranteed temperature range varies with the operating voltage.

AC Characteristics

($V_{SS} = 0\text{ V}$, $V_{DD} = 4.5\text{ to }5.5\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Machine Cycle Time	tcy	NORMAL 1, 2 mode	0.25	-	4	μs
		IDLE 0, 1, 2 mode				
		SLOW 1, 2 mode	117.6	-	133.3	
		SLEEP 0, 1, 2 mode				
High Level Clock Pulse Width	twcH	For external clock operation (XIN input)	-	31.25	-	ns
Low Level Clock Pulse Width	twcL	fc = 16 MHz				
High Level Clock Pulse Width	twcH	For external clock operation (XTIN input)	-	15.26	-	μs
Low Level Clock Pulse Width	twcL	fc = 32.768 kHz				

($V_{SS} = 0\text{ V}$, $V_{DD} = 2.7\text{ to }4.5\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Machine Cycle Time	tcy	NORMAL 1, 2 mode	0.5	-	4	μs
		IDLE 0, 1, 2 mode				
		SLOW 1, 2 mode	117.6	-	133.3	
		SLEEP 0, 1, 2 mode				
High Level Clock Pulse Width	twcH	For external clock operation (XIN input)	-	62.5	-	ns
Low Level Clock Pulse Width	twcL	fc = 8 MHz				
High Level Clock Pulse Width	twcH	For external clock operation (XTIN input)	-	15.26	-	μs
Low Level Clock Pulse Width	twcL	fc = 32.768 kHz				

($V_{SS} = 0\text{ V}$, $V_{DD} = 1.8\text{ to }2.7\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

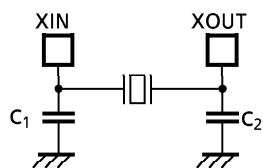
Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Machine Cycle Time	tcy	NORMAL 1, 2 mode	0.95	-	4	μs
		IDLE 0, 1, 2 mode				
		SLOW 1, 2 mode	117.6	-	133.3	
		SLEEP 0, 1, 2 mode				
High Level Clock Pulse Width	twcH	For external clock operation (XIN input)	-	119.05	-	ns
Low Level Clock Pulse Width	twcL	fc = 4.2 MHz				
High Level Clock Pulse Width	twcH	For external clock operation (XTIN input)	-	15.26	-	μs
Low Level Clock Pulse Width	twcL	fc = 32.768 kHz				

Timer Counter 1 input (ECIN) Characteristics
--

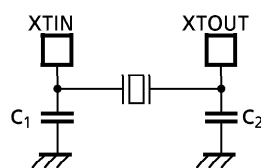
($V_{SS} = 0\text{ V}$, $T_{opr} = -40\text{ to }85^{\circ}\text{C}$)

Parameter	Symbol	Condition		Min	Typ.	Max	Unit
TC1 input (ECIN input)	t_{TC1}	Frequency measurement mode	Single edge count	-	-	1.0	MHz
		$V_{DD} = 4.5\text{ to }5.5\text{ V}$					
		Frequency measurement mode	Single edge count			0.5	
		$V_{DD} = 2.7\text{ to }4.5\text{ V}$	Single edge count	-	-	0.262	
		$V_{DD} = 1.8\text{ to }2.7\text{ V}$	Single edge count	-	-	0.262	

Recommended Oscillating Conditions



(1) High-frequency Oscillation



(2) Low-frequency Oscillation

Note 1: An electrical shield by metal shield plate on the surface of IC package is recommended in order to protect the device from the high electric field stress applied from CRT (Cathodic Ray Tube) for continuous reliable operation.

Note 2: The product numbers and specifications of the resonators by Murata Manufacturing Co., Ltd. are subject to change.

For up-to-date information, please refer to the following URL:

<http://www.murata.co.jp/search/index.html>

