



## TWO-CHANNEL DIGITAL-TO-RESOLVER CONVERTER

### DESCRIPTION

The DRC-11522 is a dual 16-bit digital-to-resolver (D/R) converter. Each channel is independent from the other with the exception of the 16 digital lines. The DRC-11522 allows the user to program the gain of the resolver output.

Packaged in a 36-pin double DIP, the DRC-11522 is two digital-to-resolver converters in one hybrid module. Using an AC reference input, the DRC-11522 is a digital-to-resolver converter. When using a DC reference input, the unit can be used as a hybrid digital-to-sin/cos DC converter. With the reference input proportional to the radius vector, the DRC-11522 converts polar to rectangular coordinates.

The circuit design in the DRC-11522 allows for higher accuracy and reduces the output scale factor variation so that the output can drive dis-

plays directly. The output line-to-line voltage can be scaled by pin programming. Other features include buffered reference input, and a wide operating temperature range.

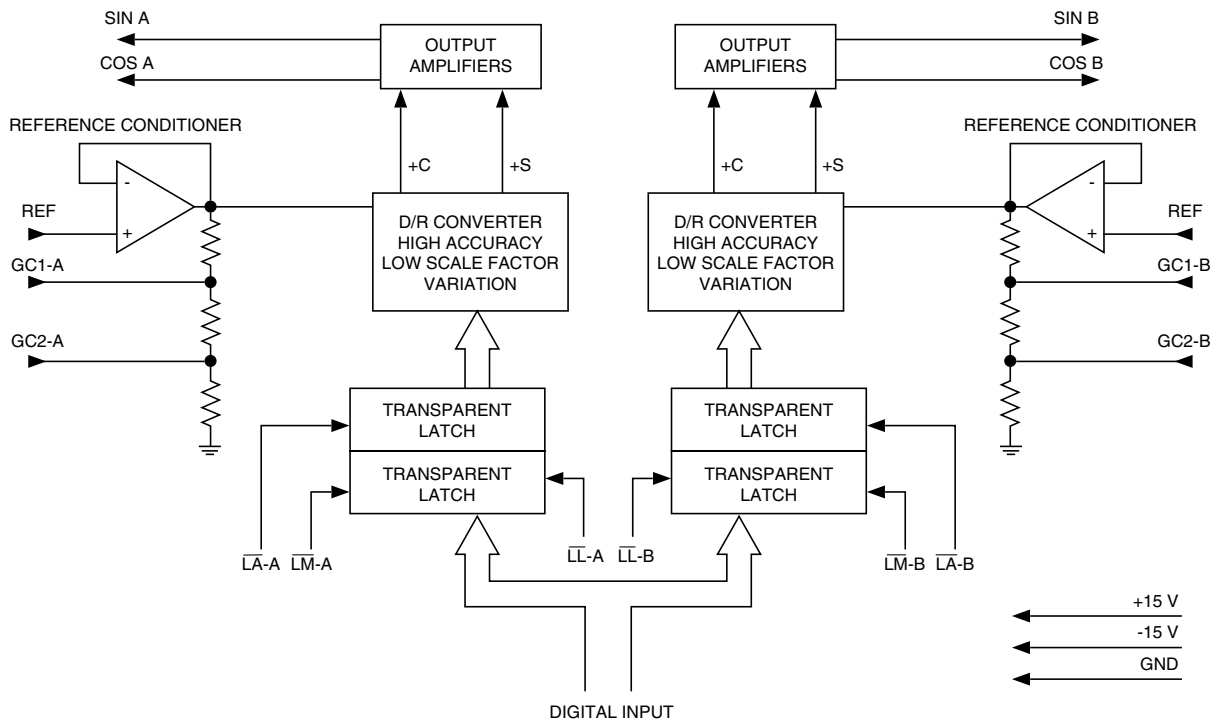
### APPLICATIONS

Because of its high reliability, small size and low power consumption, the DRC-11522 is ideal for the most stringent and severe industrial and military ground or avionics applications. All units are available with MIL-PRF-38534 processing.

Among the many possible applications are computer-based systems in which digital information is processed, such as simulators, flight trainers, flight instrumentation, fire control systems, radar and navigation systems.

### FEATURES

- **16-Bit Resolution**
- **Pin Programmable Gain Control**
- **Two Channels in One 36-Pin DDIP**
- **Accuracy: to  $\pm 2$  Min.**
- **0.1% Scale Factor Variation with Angle**
- **DC-Coupled Reference**
- **High Reliability CMOS D/R Chip**
- **8-Bit/2-Byte Double-Buffered Transparent Latches**



**FIGURE 1. DRC-11522 BLOCK DIAGRAM**

<b>TABLE 1. SPECIFICATIONS</b> (for each channel)		
Apply over temperature range, power supply ranges, reference voltage, and frequency range, and 10% harmonic distortion in the reference.		
<b>PARAMETER</b>	<b>VALUE</b>	<b>DESCRIPTION/REMARKS</b>
<b>RESOLUTION</b>	16 bits (0.33 arc minutes)	MSB = 180° LSB = 0.0055°
<b>ACCURACY</b> Output Accuracy Differential Linearity Radius accuracy	±8 minutes to ±1 minute (see ordering info) ±1 LSB max ±0.03%	Accuracy applies over operating temp. range  Simultaneous amplitude variation in both outputs as a function of digital angle
<b>DYNAMICS</b> Output Settling Time	Less than 20 µsec for any digital step change.	For any analog or digital step change
<b>DIGITAL INPUT</b> Logic Type  Logic "0" Logic "1" Load Current	  -0.3 V-dc to 1.25 V-dc +2.0 V-dc to +5.5 V-dc 20 µA max to GND 20 µA to V <sub>L</sub>	Natural binary angle parallel positive logic CMOS and TTL compatible. Inputs are CMOS transient protected. Each input has a 20 µA max pull down to GND.  External logic voltage not needed. TTL compatible.  Bits 1-16 LL, LM, LA (See timing Diagram, FIGURE 2)
<b>REFERENCE INPUT</b> Type Frequency Range Voltage Input Impedance	 DC to 1000 Hz 3.5 V ±10% 10 M Ohm min	Programmable (See TABLE 2.) DC to 10 kHz with reduced accuracy. 0 to ±10 peak AC or DC Operational Amplifier Buffer
<b>ANALOG OUTPUT</b> Type Output Current Max Output Voltage (Tracks Reference Input Voltage) Converter Gain (K) Transformation Ratio Tol. Scale Factor Variation DC Offset	 2 mA rms max $K * V_{in} * \sin \theta$ also $K * V_{in} * \cos \theta$  0.5, 1.0, or 2.0 ±1% ±0.2% max ±0.1% max ±10 mV typical, ±25 mV max	Resolver  ±10 V peak AC or DC  See TABLE 2.  Each Line to GND
<b>POWER SUPPLIES</b> Voltage Max Voltage Without Damage Current or Impedance	±15 VDC ±10% ±18 VDC ±40 mA max	For ±10 V peak output
<b>TEMPERATURE RANGES (CASE)</b> Operating -1 Option -3 Option Storage	 -55°C to +125°C 0°C to +70°C -60°C to +135°C	
<b>PHYSICAL CHARACTERISTICS</b> Type Size Weight	36-pin double DIP 0.78 x 1.9 x 0.21 inch (2.0 x 4.8 x 0.53 cm) 0.6 oz (17g) max	

## TECHNICAL INFORMATION

### DIGITAL INPUTS

For each channel, the 16-bit digital angle is double buffered with transparent latches (See FIGURE 1). The latch controls have internal pull-up current sources to +5 V, this puts the latches in the transparent mode when they are not connected.

Angle is determined by adding the logic bits. The enable inputs are  $\overline{LL}$  (1st Latch LSBs),  $\overline{LM}$  (1st Latch MSBs), and  $\overline{LA}$  (2nd Latch All); see FIGURE 2 for timing.

### OUTPUT SCALING AND REFERENCE LEVEL ADJUSTMENT

The DRC-11522 operates like a multiplying D/A converter in that the voltage of each output line is directly proportional to the reference voltage. The maximum line-to-line levels are determined by the output amplifiers and are programmable for a gain of 0.5, 1.0, or 2.0 (See TABLE 2.).

GC1-A (PIN 7)	GC2-A (PIN 8)	GAIN (K)
GND	OPEN	0.5
OPEN	GND	1.0
OPEN	OPEN	2.0
GC1-B (PIN 4)	GC2-B (PIN 5)	GAIN (K)

### OUTPUT PHASING AND OUTPUT SCALE FACTOR

The analog output signals have the following phasing:

$$\sin = (\text{REF} * K) A_0 [1 + A(\theta)] \sin \theta$$

$$\cos = (\text{REF} * K) A_0 [1 + A(\theta)] \cos \theta$$

The output amplifiers simultaneously track reference voltage fluctuations because they are proportional to (REF \* K). The transformation ratio  $A_0$  is determined by the programmable gain inputs (0.5, 1.0, or 2.0). The maximum variation in  $A_0$  from all causes is 0.1%. The term  $A(\theta)$  represents the variation of the amplitude with the digital signal input angle.  $A(\theta)$ , which is called the scale factor variation, is a smooth function of  $(\theta)$  without discontinuities and is less than  $\pm 0.1\%$  for all values of  $(\theta)$ . The total maximum variation in  $A_0[1 + A(\theta)]$  is therefore  $\pm 0.2\%$ .

Because the amplitude factor  $(\text{REF} * K) A_0 [1 + A(\theta)]$  varies simultaneously on all output lines, it is not a source of error when the DRC-11522 is driving a ratiometric system. However, if the outputs are used independently, as in x-y plotters, the amplitude variations must be taken into account.

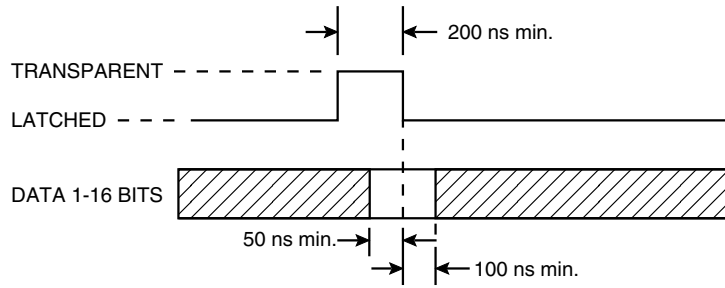
TABLE 3. PINOUTS

PIN	FUNCTION	PIN	FUNCTION
1	$\overline{LL}$ -B	19	Bit 16 (LSB)
2	COS A	20	Bit 15
3	SIN A	21	Bit 14
4	GC1-B	22	Bit 13
5	GC2-B	23	Bit 12
6	Ref B	24	Bit 11
7	GC1-A	25	Bit 10
8	GC2-A	26	Bit 9
9	Ref A	27	Bit 8
10	COS B	28	Bit 7
11	SIN B	29	Bit 6
12	NC	30	Bit 5
13	+15 V	31	Bit 4
14	-15 V	32	Bit 3
15	$\overline{LA}$ -B	33	Bit 2
16	$\overline{LA}$ -A	34	Bit 1 (MSB)
17	$\overline{LL}$ -A	35	$\overline{LM}$ -A
18	GND	36	$\overline{LM}$ -B

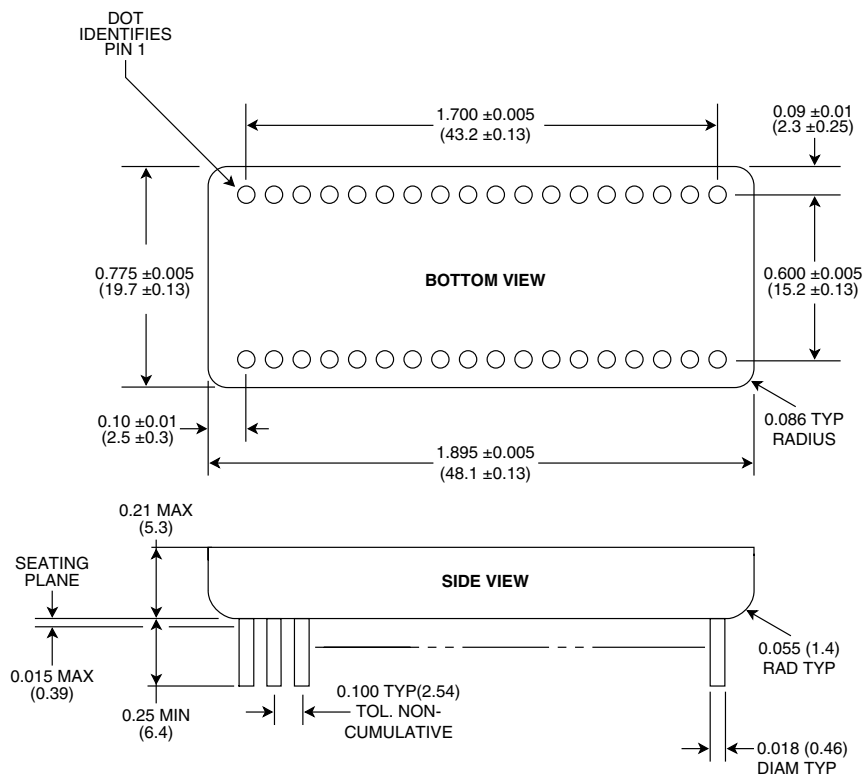
NOTE: Functions  $\overline{LL}$ ,  $\overline{LM}$ ,  $\overline{LA}$  both A and B may be left unconnected when not used.

TABLE 4. PIN DEFINITIONS

PIN	DEFINITION
GND	Power Supply Ground Digital Ground Analog Signal Ground
$\overline{B1-B16}$	Digital Input bits B1, = MSB = 180 degrees
$\overline{LM-A}$	High Byte Enable (B1-B8) for MSB's 8-bit Input register of channel A. Logic high enables, low holds.
$\overline{LM-B}$	High Byte Enable (B1-B8) for MSB's 8-bit Input register channel B Logic high enables, low holds.
$\overline{LL-A}$	Low Byte Enable (B9-B16) for LSB's 8-bit Input register of channel A. Logic high enables, low holds
$\overline{LL-B}$	Low Byte Enable (B9-B16) for LSB's 8-bit Input register of channel B. Logic high enables, low holds.
$\overline{LA-A}$	Channel A Load Converter. Logic high transfers Channel A input registers data into 16-bit holding register. When low, Channel A is in hold mode.
$\overline{LA-B}$	Channel B Load Converter. Logic high transfers Channel B input registers data into 16-bit holding register. When low, Channel B is in hold mode.
+15 V	Power Supply Voltage.
-15 V	Power Supply Voltage.
	<b>CAUTION:</b> <b>REVERSAL OF POWER SUPPLIES</b> <b>WILL DAMAGE THE CONVERTER.</b>
Ref-A	Channel A reference voltage Input
Ref-B	Channel B reference voltage input
GC1-A	Channel A gain programming pin
GC2-A	Channel A gain programming pin
GC1-B	Channel B gain programming pin
GC2-B	Channel B gain programming pin
Sin A	Analog output of Channel A
Cos A	Analog output of Channel A
Sin B	Analog output of Channel B
Cos B	Analog output of Channel B



**FIGURE 2.  $\overline{LL}$ ,  $\overline{LM}$ , AND  $\overline{LA}$  TIMING DIAGRAM**



**Notes:**

- Dimensions shown are in inches (millimeters)
- Lead identification numbers are for reference only.
- Lead cluster shall be centered within  $\pm 0.010$  ( $\pm 2.54$ ) of outline dimensions.  
Lead spacing dimensions apply only at seating plane.
- Pin material meets solderability requirements of MIL-STD-202E, Method 208C.
- Package is Kovar with electroless nickel plating.
- Case is electrically floating.

**FIGURE 3. DRC-11522 MECHANICAL OUTLINE (36-PIN DOUBLE DIP)**

**ORDERING INFORMATION**

DRC-11522-XXXX

**Supplemental Process Requirements:**

- S = Pre-Cap Source Inspection
- L = Pull Test
- Q = Pull Test and Pre-Cap Inspection
- K = One Lot Date Code
- W = One Lot Date Code and PreCap Source
- Y = One Lot Date Code and 100% Pull Test
- Z = One Lot Date Code, PreCap Source and 100% Pull Test
- Blank = None of the Above

**Accuracy:**

- 3 = ±4 minutes
- 4 = ±2 minutes

**Process Requirements:**

- 0 = Standard DDC Processing, no Burn-In (See table below.)
- 1 = MIL-PRF-38534 Compliant
- 2 = B\*
- 3 = MIL-PRF-38534 Compliant with PIND Testing
- 4 = MIL-PRF-38534 Compliant with Solder Dip
- 5 = MIL-PRF-38534 Compliant with PIND Testing and Solder Dip
- 6 = B\* with PIND Testing
- 7 = B\* with Solder Dip
- 8 = B\* with PIND Testing and Solder Dip
- 9 = Standard DDC Processing with Solder Dip, no Burn-In (See table below.)

**Temperature Grade/Data Requirements:**

- 1 = -55°C to +125°C
- 2 = -40°C to +85°C
- 3 = 0°C to +70°C
- 4 = -55°C to +125°C with Variables Test Data
- 5 = -40°C to +85°C with Variables Test Data
- 8 = 0°C to +70°C with Variables Test Data

\*Standard DDC Processing with burn-in and full temperature test — see table below.

STANDARD DDC PROCESSING		
TEST	MIL-STD-883	
	METHOD(S)	CONDITION(S)
INSPECTION	2009, 2010, 2017, and 2032	—
SEAL	1014	A and C
TEMPERATURE CYCLE	1010	C
CONSTANT ACCELERATION	2001	A
BURN-IN	1015, Table 1	—

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105 Wilbur Place, Bohemia, New York 11716-2482

**For Technical Support - 1-800-DDC-5757 ext. 7389 or 7413**

**Headquarters** - Tel: (631) 567-5600 ext. 7389 or 7413, Fax: (631) 567-7358

**West Coast** - Tel: (714) 895-9777, Fax: (714) 895-4988

**Southeast** - Tel: (703) 450-7900, Fax: (703) 450-6610

**United Kingdom** - Tel: +44-(0)1635-811140, Fax: +44-(0)1635-32264

**Ireland** - Tel: +353-21-341065, Fax: +353-21-341568

**France** - Tel: +33-(0)1-41-16-3424, Fax: +33-(0)1-41-16-3425

**Germany** - Tel: +49-(0)8141-349-087, Fax: +49-(0)8141-349-089

**Sweden** - Tel: +46-(0)8-54490044, Fax +46-(0)8-7550570

**Japan** - Tel: +81-(0)3-3814-7688, Fax: +81-(0)3-3814-7689

**World Wide Web** - <http://www.ddc-web.com>



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