

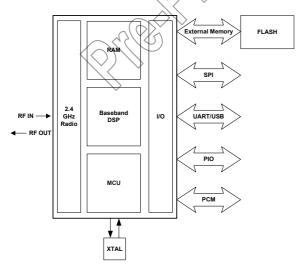
# **Device Features**

- Fully Qualified Bluetooth v1.2 system
- Enhanced Data Rate (EDR) compliant with v0.9 of specification for both 2Mbps and 3Mbps modulation modes
- Full Speed Bluetooth Operation with Full Piconet Support
- Scatternet Support
  - 1.8V core, 1.8 to 3.6V I/O
  - Low Power 1.8V Operation
  - Small footprint 8 x 8mm 96-ball VFBGA Package
  - Also available in 6 x 6mm package option
  - Minimum External Components
  - Integrated 1.8V regulator
  - USB and Dual UART Ports
  - Support for 802.11 Coexistence
  - Support for 8Mbit External Flash
  - RoHS Compliant

# **General Description**

**BlueCore4-External** is a single chip radio and baseband IC for Bluetooth 2.4GHz systems including enhanced data rates (EDR) to 3Mbps.

BC417143B interfaces to 8Mbit of external Flash memory. When used with the CSR Bluetooth software stack, it provides a fully compliant Bluetooth system to v1.2 of the specification for data and voice communications.



#### BlueCore4-External System Architecture

# BlueCore™4-External

# Single Chip Bluetooth® v1.2 System with EDR

**Pre-Production Information Datasheet For** 

BC417143B-ES-IQN BC417143B-ES-IRN

September 2004

# Applications

#### > PCs

- Personal Digital Assistants (PDAs)
- Computer Accessories (compact Flash Cards, PCMCIA Cards, SD Cards and USB Dongles)
- Access Points
- Digital Cameras

BlueCore4-External has been designed to reduce the number of external components required which ensures production costs are minimised.

The device incorporates auto-calibration and built-in self-test (BIST) routines to simplify development, type approval and production test. All hardware and device firmware is fully compliant with the Bluetooth v1.2 Specification (all mandatory and optional features).

To improve the performance of both Bluetooth and 802.11b/g co-located systems a wide range of co-existence features are available including two types of hardware signalling: basic activity signalling and Intel WCS activity & channel signalling



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# **Status Information**

The status of this Data Sheet is **Pre-Production Information**.

CSR Product Data Sheets progress according to the following format:

#### Advance Information:

Information for designers concerning CSR product in development. All values specified are the target values of the design. Minimum and maximum values specified are only given as guidance to the final specification limits and must not be considered as the final values.

All detailed specifications including pinouts and electrical specifications may be changed by CSR without notice.

#### **Pre-Production Information:**

Pinout and mechanical dimension specifications finalised. All values specified are the target values of the design. Minimum and maximum values specified are only given as guidance to the final specification limits and must not be considered as the final values.

All electrical specifications may be changed by CSR without notice.

#### **Production Information:**

Final Data Sheet including the guaranteed minimum and maximum limits for the electrical specifications.

Production Data Sheets supersede all previous document versions.

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#### **RoHS Compliance**

BlueCore4-External devices meet the requirements of Directive 2002/95/EC of the European Parliament and of the Council on the Restriction of Hazardous Substance (RoHS).

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# 1 Key Features

## Radio

- Common TX/RX terminal simplifies external matching; eliminates external antenna switch
- BIST minimises production test time. No external trimming is required in production
- Full RF reference designs available
- Bluetooth v1.2 Specification compliant

EDR v0.9 compliant

## Transmitter

- +6dBm RF transmit power with level control from on-chip 6-bit DAC over a dynamic range >30dB
- Class 2 and Class 3 support without the need for an external power amplifier or TX/RX switch
- Supports π/4 DQPSK (2Mbps) and 8DPSK (3Mbps) modulation

# Receiver

- Integrated channel filters
- Digital demodulator for improved sensitivity and co-channel rejection
- Real time digitised RSSI available on HCI interface
- Fast AGC for enhanced dynamic range
- Supports π/4 DQPSK and 8DPSK modulation
- Channel classification

# Synthesiser

- Fully integrated synthesiser requires no external VCO, varactor diode, resonator or loop filter
- Compatible with crystals between 8 and 32MHz (in multiples of 250kHz) or an external clock
- Accepts 7.68, 14.44, 15.36, 16.2, 16.8, 19.2, 19.44, 19.68, 19.8 and 38, 4MHz TCXO frequencies for GSM and CDMA devices with sinusoidal or logic level signals

# **Auxiliary Features**

- Crystal oscillator with built-in digital trimming
- Power management includes digital shut down, wake up commands with an integrated low power oscillator for ultra-low power Park/Sniff/Hold mode
- 'Clock request' output to control an external clock

# **Auxiliary Features (continued)**

- On-chip linear regulator; 1.8V output from a 2.2 4.2V input
- Can run in low power mode from external 32kHz clock signal
- Auto baud rate setting for different TCXO frequencies
- Power-on-reset cell detects low supply voltage
- Arbitrary power supply sequencing permitted
- 8-bit ADC available to applications

### **Baseband and Software**

- External 8Mbit Flash for complete system solution
- Internal 48Kbyte RAM, allows full speed data transfer, mixed voice and data, and full piconet operation, including all medium rate preset types
- Logic for forward error correction, header error control, access code correlation, CRC, demodulation, encryption bit stream generation, whitening and transmit pulse shaping. Supports all Bluetooth v1.2 features including eSCO and AFH
- Transcoders for A-law, μ-law and linear voice from host and A-law, μ-law and CVSD voice over air

# **Physical Interfaces**

Synchronous serial interface up to 4Mbaud for system debugging

- UART interface with programmable baud rate up to 1.5Mbaud with an optional bypass mode
- Full speed USB v2.0 interface supports OHCI and UHCI host interfaces
- Synchronous bi-directional serial programmable audio interface
- Optional I<sup>2</sup>C<sup>TM</sup> compatible interface
- Optional co-existence interfaces

## **Bluetooth Stack**

CSR's Bluetooth Protocol Stack runs on the on-chip MCU in a variety of configurations:

- Standard HCI (UART or USB)
- Fully embedded RFCOMM
- Customised builds with embedded application code

## Package Options

- 96-ball VFBGA, 8 x 8 x 1mm, 0.65mm pitch
- 96-ball VFBGA, 6 x 6 x 1mm, 0.5mm pitch



# 2 8 x 8mm VFBGA Package Information

# 2.1 BlueCore4-External Pinout Diagram

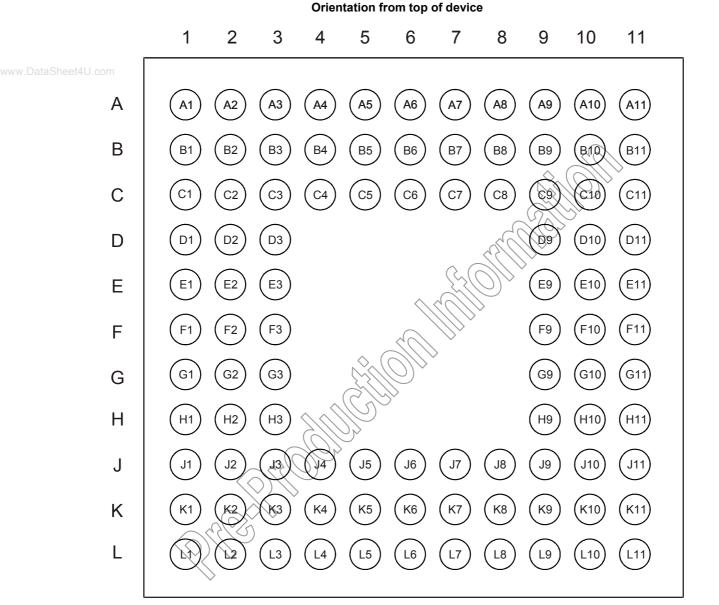


Figure 2.1: BlueCore4-External 8 x 8mm Device Pinout (BC417143B-ES-IQN)

# 2.2 Device Terminal Functions (BC417143B-ES-IQN)

Radio	Ball	Pad Type	Description
PIO[0]/RXEN	B1	Bi-directional with programmable strength internal pull-up/down	Control output for external LNA (if fitted)
PIO[1]/TXEN	B2	Bi-directional with programmable strength internal pull-up/down	Control output for external PA (If fitted)
RF_IN	D1	Analogue	Single ended receiver input
TX_A	F1	Analogue	Transmitter output/switched receiver input
TX_B	E1	Analogue	Complement of TX_A

Synthesiser and Oscillator	Ball	Pad Type	Description
XTAL_IN	L1	Analogue	For crystal or external clock input
XTAL_OUT	L2	Analogue	Drive for crystal

USB and UART	Ball	Pad Type	Description
UART_TX	G9	CMOS output, tri-state, with weak internal pull-up	UART data oùtput
UART_RX	H10	CMOS input with weak internal pull-down	UART data input
UART_RTS	H9	CMOS output, tri-state, with weak internal pull-up	WART request to send active low
UART_CTS	J11	CMOS input with weak internal pull-down	UART clear to send active low
USB_DP	K10	Bi-directional	USB data plus with selectable internal $1.5k\Omega$ pull-up resistor
USB_DN	K11	Bi-directional	USB data minus
		GIUN	

PCM Interface	Ball	Pad Type	Description
PCM_OUT	F9	CMOS output, tri-state, with weak internal pull-down	Synchronous data output
PCM_IN	(H1) )	CMOS input, with weak internal pull-down	Synchronous data input
PCM_SYNC	G11	Bi-directional with weak internal pull-down	Synchronous data sync
PCM_CLK	G10	Bi-directional with weak internal pull-down	Synchronous data clock

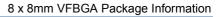


PIO Port	Ball	Pad Type	Description
PIO[11]	G3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[10]	F3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[9]	E3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[8]	D3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[7]	F10	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[6]/ WLAN_Active/ Ch_Data	F11	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line or Optionally WLAN_Active/Ch_Data input for co-existence signalling
PIO[5]/BT_Active	E9	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line or Optionally BT_Active output for co-existence signalling
PIO[4]/ BT_Priority/Ch_C	Cik E10	Bi-directional with programmable strength internal pull-up/down	Rrogrammable input/output line or Optionally BT_Priority/Ch_Clk output for co-existence signalling
PIO[3]	J3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[2]	H3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
AIO[0]	K1	Bi-directional	Programmable input/output line
AIO[1]	J2	Bi-directional	Programmable input/output line
AIO[2]	K2 ((	Bi-directional	Programmable input/output line
	$-\langle 0 \rangle / \langle 0 \rangle - \langle 0 \rangle - \langle 0 \rangle / \langle 0 \rangle - \langle 0 \rangle - \langle 0 \rangle / \langle 0 \rangle - \langle 0 \rangle - \langle 0 \rangle / \langle 0 \rangle - \langle 0 \rangle - \langle 0 \rangle / \langle 0 \rangle - \langle 0 \rangle - \langle 0 \rangle / \langle 0 \rangle - \langle 0$	$\sim$	
Test and Debug	Ball	<sup>&gt;</sup> Pad Type	Description

Test and Debug	Ball	Pad Type	Description
RESETB	В10	CMOS input with weak internal pull-up	Reset if low. Input debounced so must be low for >5ms to cause a reset
SPI_CSB	C11	CMOS input with weak internal pull-up	Chip select for Synchronous Serial Interface active low
SPI_CLK	C10	CMOS input with weak internal pull-down	Serial Peripheral Interface clock
SPI_MOSI	D10	CMOS input with weak internal pull-down	Serial Peripheral Interface data input
SPI_MISO	C9	CMOS output, tri-state, with weak internal pull-down	Serial Peripheral Interface data output
TEST_EN	C8	CMOS input with strong internal pull-down	For test purposes only (leave unconnected)



A[18]L7CMOS output, tri-stateAddress lineA[17]K7CMOS output, tri-stateAddress lineA[16]A10CMOS output, tri-stateAddress lineA[15]L10CMOS output, tri-stateAddress lineA[14]K9CMOS output, tri-stateAddress lineA[13]J9CMOS output, tri-stateAddress lineA[12]L9CMOS output, tri-stateAddress lineA[11]J8CMOS output, tri-stateAddress lineA[11]J8CMOS output, tri-stateAddress lineA[10]K8CMOS output, tri-stateAddress lineA[11]J5CMOS output, tri-stateAddress lineA[11]J5CMOS output, tri-stateAddress lineA[11]J5CMOS output, tri-stateAddress lineA[2]L6CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress lineA[1]A3CMOS output, tri-stateAddress line
A[16]A10CMOS output, tri-stateAddress lineA[15]L10CMOS output, tri-stateAddress lineA[14]K9CMOS output, tri-stateAddress lineA[13]J9CMOS output, tri-stateAddress lineA[12]L9CMOS output, tri-stateAddress lineA[11]J8CMOS output, tri-stateAddress lineA[10]K8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[8]J7CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[15]L10CMOS output, tri-stateAddress lineA[14]K9CMOS output, tri-stateAddress lineA[13]J9CMOS output, tri-stateAddress lineA[12]L9CMOS output, tri-stateAddress lineA[11]J8CMOS output, tri-stateAddress lineA[11]J8CMOS output, tri-stateAddress lineA[10]K8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[8]J7CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[2]J4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[14]K9CMOS output, tri-stateAddress lineA[13]J9CMOS output, tri-stateAddress lineA[12]L9CMOS output, tri-stateAddress lineA[11]J8CMOS output, tri-stateAddress lineA[10]K8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[8]J7CMOS output, tri-stateAddress lineA[7]J5CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[13]J9CMOS output, tri-stateAddress lineA[12]L9CMOS output, tri-stateAddress lineA[11]J8CMOS output, tri-stateAddress lineA[10]K8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[6]J7CMOS output, tri-stateAddress lineA[7]J5CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[2]J4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[12]L9CMOS output, tri-stateAddress lineA[11]J8CMOS output, tri-stateAddress lineA[10]K8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[8]J7CMOS output, tri-stateAddress lineA[7]J5CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[2]J4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[11]J8CMOS output, tri-stateAddress lineA[10]K8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[8]J7CMOS output, tri-stateAddress lineA[7]J5CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[2]J4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[10]K8CMOS output, tri-stateAddress lineA[9]L8CMOS output, tri-stateAddress lineA[8]J7CMOS output, tri-stateAddress lineA[7]J5CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[2]J4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[9]L8CMOS output, tri-stateAddress lineA[8]J7CMOS output, tri-stateAddress lineA[7]J5CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[2]J4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[8]J7CMOS output, tri-stateAddress lineA[7]J5CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[2]J4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[7]J5CMOS output, tri-stateAddress lineA[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[2]J4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[6]   L6   CMOS output, tri-state   Address line     A[5]   K6   CMOS output, tri-state   Address line     A[4]   K5   CMOS output, tri-state   Address line     A[3]   L5   CMOS output, tri-state   Address line     A[2]   J4   CMOS output, tri-state   Address line     A[1]   K4   CMOS output, tri-state   Address line
A[6]L6CMOS output, tri-stateAddress lineA[5]K6CMOS output, tri-stateAddress lineA[4]K5CMOS output, tri-stateAddress lineA[3]L5CMOS output, tri-stateAddress lineA[2]J4CMOS output, tri-stateAddress lineA[1]K4CMOS output, tri-stateAddress line
A[4]   K5   CMOS output, tri-state   Address line     A[3]   L5   CMOS output, tri-state   Address line     A[2]   J4   CMOS output, tri-state   Address line     A[1]   K4   CMOS output, tri-state   Address line
A[3]   L5   CMOS output, tri-state   Address time     A[2]   J4   CMOS output, tri-state   Address time     A[1]   K4   CMOS output, tri-state   Address time
A[2] J4 CMOS output, tri-state Address line   A[1] K4 CMOS output, tri-state Address line
A[1] K4 CMOS output, tri-state Address line





External Memory Data Interface	Ball	Pad Type	Description
D[15]	B9	Bi-directional with weak internal pull-down	Data line
D[14]	B8	Bi-directional with weak internal pull-down	Data line
D[13]	C7	Bi-directional with weak internal pull-down	Data line
D[12]	A7	Bi-directional with weak internal pull-down	Data line
.D[11]	B6	Bi-directional with weak internal pull-down	Data line
D[10]	C5	Bi-directional with weak internal pull-down	Data line
D[9]	A5	Bi-directional with weak internal pull-down	Data line
D[8]	B4	Bi-directional with weak internal pull-down	Data line
D[7]	A9	Bi-directional with weak internal pull-down	Data line
D[6]	A8	Bi-directional with weak internal pull-down	Data line
D[5]	B7	Bi-directional with weak internal pull-down	Data line
D[4]	C6	Bi-directional with weak internal pull-down	Data line
D[3]	A6	Bi-directional with weak internal pull-down	Data line
D[2]	B5	Bi-directional with weak internal pull-down	Data line
D[1]	C4	Bi-directional with weak	Data line
D[0]	A4	Bi-directional with weak	Data line
External Memory	Ball	Pad Type	Description

External Memory Interface	Ball	Pad Type	Description
REB	Сз	CMOS output, tri-state with internal weak pull-up	Read enable for external memory (active low)
WEB	J6	CMOS output, tri-state with internal weak pull-up	Write enable for external memory (active low)
CSB	B3	CMOS output, tri-state with internal weak pull-up	Chip select for external memory (active low)



	ower Supplies and ontrol	Ball	Pad Type	Description
V	REG_IN	L4	VDD/Regulator input	Linear regulator input
V	REG_EN	H2	CMOS input	High or not connected to enable regulator. VSS to disable regulator
V	DD_USB	L11	VDD	Positive supply for UART/USB ports
V	DD_PIO	A2	VDD	Positive supply for PIO <sup>(1)</sup>
V	DD_PADS	D11	VDD	Positive supply for all other digital Input/Output ports <sup>(2)</sup>
VI 4U.co	DD_MEM	A11	VDD	Positive supply for external memory and AIO ports
V	DD_CORE	E11	VDD	Positive supply for internal digital circuitry
V	DD_RADIO	C1	VDD	Positive supply for RF circuitry
V	DD_LO	J1	VDD	Positive supply for VCO and synthesiser circuitry
V	DD_ANA	G2, L3	VDD/Regulator output	Positive supply for analogue circuitry and 1.8V regulated output. For performance, regulator decoupling and loads should be connected to ball adjacent to VREG_IN
V	SS_DIG	A1, D9, J10	VSS	Ground connection for digital ports
V	SS_RADIO	D2, E2, F2	VSS	Ground connections for RF circuitry
V	SS_LO	H1	vss	Ground connections for VCO and synthesiser
V	SS_ANA	G1, K3	vss	Ground connections for analogue circuitry

#### Notes:

- <sup>(1)</sup> Positive supply for PIO[3:0] and PIO[11:8]
- (2) Positive supply for SPI/PCM ports and PIO[7:4]

Unconnected Terminals	r Oler	Description
B11, C2	1, C2	Leave unconnected



# 3 6 x 6mm VFBGA Package Information

# 3.1 BlueCore4-External 6 x 6mm Pinout Diagram

#### 1 2 3 4 5 6 7 8 9 10 11 Α A1 A2 A3 A4 A5 A6 A8 A9 A10 A11 A7 В B1 B2 В3 Β4 Β5 B6 B7 B8 B10 B11 В9 С C1 C2 C3 C4 C5 C6 C8 C10 C11 C7 <u>69</u> D D1 D2 D3 D9 D10 D11 Ε E1 E2 E3 E9 E10 E11 F1 F2 F3 F9 F10 F11 F G3 G9 G10 G1 G2 G11 G Η H3 H9 H10 H1 H2 H11 J2 **(**3 J5 J6 J7 J8 J9 J10 J J1 .)14 J11

kз

L3

K4

L4

K5

L5

Orientation from top of device

Figure 3.1: BlueCore4-External 6 x 6mm Device Pinout (BC417143B-ES-IRN)

K6

L6

K7

L7

K8

L8

K9

L9

K10

L10

K11

L11

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K1

K2

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# 3.2 Device Terminal Functions (BC417143B-ES-IRN)

Radio	Ball	Pad Type	Description
PIO[0]/RXEN	C1	Bi-directional with programmable strength internal pull-up/down	Control output for external LNA (if fitted)
PIO[1]/TXEN	C2	Bi-directional with programmable strength internal pull-up/down	Control output for external PA (If fitted)
RF_IN	D1	Analogue	Single ended receiver input
TX_A	F1	Analogue	Transmitter output/switched receiver input
TX_B	E1	Analogue	Complement of TX_A

Synthesiser and Oscillator	Ball	Pad Type	Description
XTAL_IN	L1	Analogue	For crystal or external clock input
XTAL_OUT	L2	Analogue	Drive for crystal

USB and UART	Ball	Pad Type	Description
UART_TX	G9	CMOS output, tri-state, with weak internal pull-up	UART data oùtput
UART_RX	H10	CMOS input with weak internal pull-down	UART data input
UART_RTS	H9	CMOS output, tri-state, with weak internal pull-up	WART request to send active low
UART_CTS	J11	CMOS input with weak internal pull-down	UART clear to send active low
USB_DP	K10	Bi-directional	USB data plus with selectable internal $1.5k\Omega$ pull-up resistor
USB_DN	K11	Bi-directional	USB data minus
		2///VP	

PCM Interface	Ball	Pad Type	Description
PCM_OUT	F9	CMOS output, tri-state, with weak internal pull-down	Synchronous data output
PCM_IN	(H1) )	CMOS input, with weak internal pull-down	Synchronous data input
PCM_SYNC	G11	Bi-directional with weak internal pull-down	Synchronous data sync
PCM_CLK	G10	Bi-directional with weak internal pull-down	Synchronous data clock

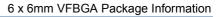


PIO Port	Ball	Pad Type	Description
PIO[11]	D2	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[10]	F3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[9]	G3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[8]	H3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[7]	F10	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[6]/ WLAN_Active/ Ch_Data	F11	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line or Optionally WLAN_Active/Ch_Data input for co-existence signalling
PIO[5]/BT_Active	E9	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line or Optionally B1_Active output for co-existence signalling
PIO[4]/ BT_Priority/Ch_Clk	E10	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line or Optionally BT_Priority/Ch_Clk output for co-existence signalling
PIO[3]	B2	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
PIO[2]	J3	Bi-directional with programmable strength internal pull-up/down	Programmable input/output line
AIO[0]	L4	Bi-directional	Programmable input/output line
AIO[1]	K3	Bi-directional	Programmable input/output line
AIO[2]	K2 ((	Bi-directional	Programmable input/output line
	$\bigcirc)$		

Test and Debug	Ball	Pad Type	Description
RESETB	D10	CMOS input with weak internal pull-up	Reset if low. Input debounced so must be low for >5ms to cause a reset
SPI_CSB	C11	CMOS input with weak internal pull-up	Chip select for Synchronous Serial Interface active low
SPI_CLK	B9	CMOS input with weak internal pull-down	Serial Peripheral Interface clock
SPI_MOSI	C10	CMOS input with weak internal pull-down	Serial Peripheral Interface data input
SPI_MISO	C9	CMOS output, tri-state, with weak internal pull-down	Serial Peripheral Interface data output
TEST_EN	C8	CMOS input with strong internal pull-down	For test purposes only (leave unconnected)



External Memory Address Interface	Ball	Pad Type	Description
A[18]	L7	CMOS output, tri-state	Address line
A[17]	K7	CMOS output, tri-state	Address line
A[16]	A9	CMOS output, tri-state	Address line
A[15]	L10	CMOS output, tri-state	Address line
A[14]	K9	CMOS output, tri-state	Address line
A[13]	J9	CMOS output, tri-state	Address line
A[12]	L9	CMOS output, tri-state	Address line
A[11]	J8	CMOS output, tri-state	Address line
A[10]	K8	CMOS output, tri-state	Address line
A[9]	L8	CMOS output, tri-state	Address line
A[8]	J7	CMOS output, tri-state	Address line
A[7]	K6	CMOS output, tri-state	Address line
A[6]	L6	CMOS output, tri-state	Address line
A[5]	K5	CMOS output, tri-state	Address line
A[4]	J5	CMOS output, tri-state	Address line
A[3]	L5	CMOS output, tri-state	Address (tine
A[2]	J4	CMOS output, tri-state	Address line
A[1]	K4	CMOS output, tri-state	Address line
A[0]	A2	CMOS output, tri-state	Address line
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External Memory Data Interface	Ball	Pad Type	Description
D[15]	B8	Bi-directional with weak internal pull-down	Data line
D[14]	B7	Bi-directional with weak internal pull-down	Data line
D[13]	C7	Bi-directional with weak internal pull-down	Data line
D[12]	A6	Bi-directional with weak internal pull-down	Data line
D[11]	B5	Bi-directional with weak internal pull-down	Data line
D[10]	C5	Bi-directional with weak internal pull-down	Data line
D[9]	A4	Bi-directional with weak internal pull-down	Data line
D[8]	B3	Bi-directional with weak internal pull-down	Data line
D[7]	A8	Bi-directional with weak internal pull-down	Data line
D[6]	A7	Bi-directional with weak internal pull-down	Data line
D[5]	B6	Bi-directional with weak internal pull-down	<b>Data line</b>
D[4]	C6	Bi-directional with weak internal pull-down	Data line
D[3]	A5	Bi-directional with weak internal pull-down	Data line
D[2]	B4	Bi-directional with weak internal pull-down	Data line
D[1]	C4	Bi-directional with weak	Data line
D[0]	A3	Bi-directional with weak internal pull-down	Data line
	$\sim$		

	A3	Bi-directional with weak	Data line
	$\sim 420$	2)	
	Ball	Pad Type	Description
	Сз	CMOS output, tri-state with internal weak pull-up	Read enable for external memory (active low)
~	J6	CMOS output, tri-state with internal weak pull-up	Write enable for external memory (active low)
	D3	CMOS output, tri-state with internal weak pull-up	Chip select for external memory (active low)

REB

WEB

CSB



	Power Supplies and Control	Ball	Pad Type	Description
	VREG_IN	K1	VDD/Regulator input	Linear regulator input
Ī	VREG_EN	H2	CMOS input	High or not connected to enable regulator. VSS to disable regulator
	VDD_USB	L11	VDD	Positive supply for UART/USB ports
	VDD_PIO	A1	VDD	Positive supply for PIO <sup>(1)</sup>
Ī	VDD_PADS	D11	VDD	Positive supply for all other digital Input/Output ports <sup>(2)</sup>
:4U	VDD_MEM .com	B10	VDD	Positive supply for external memory and AIO ports
	VDD_CORE	E11	VDD	Positive supply for internal digital circuitry
	VDD_RADIO	G1	VDD	Positive supply for RF circuitry
	VDD_LO	J1	VDD	Positive supply for VCO and synthesiser circuitry
	VDD_ANA	L3	VDD/Regulator output	Positive supply for analogue circuitry and 1.8V regulated output. For performance, regulator decoupling and loads should be connected to ball adjacent to VREG_IN
	VSS_DIG	B1, D9, J10	VSS	Ground connection for digital ports
Ī	VSS_RADIO	E2, F2, G2	VSS	Ground connections for RF circuitry
	VSS_LO	H1	vss	Ground connections for VCO and synthesiser
	VSS_ANA	J2	VSS	Ground connections for analogue circuitry

Notes:

<sup>(1)</sup> Positive supply for PIO[3:0] and PIO[11:8]

(2) Positive supply for SPI/PCM ports and PIO[7:4]

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Unconnected Terminals	Description
A10, A11, B11, E3	Leave unconnected



# **4** Electrical Characteristics

Absolute Maximum Ratings					
Rating	Min	Мах			
Storage Temperature	-40°C	+150°C			
Supply Voltage: VDD_RADIO, VDD_LO, VDD_ANA and VDD_CORE	-0.4V	2.2V			
Supply Voltage: VDD_MEM, VDD_PADS, VDD_PIO and VDD_USB	-0.4V	3.7V			
Supply Voltage: VREG_IN	-0.4V	5.6V			
Other Terminal Voltages	VSS-0.4V	VDD+0.4V			

Recommended Operating Conditions		~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~
Operating Condition	Min 🔷	Max
Operating Temperature Range	-40°C	+105°C
Guaranteed RF performance range (1)	-40°C	+105°C
Supply Voltage: VDD_RADIO, VDD_LO, VDD_ANA and VDD_CORE	1.70	1.9V
Supply Voltage: VDD_MEM, VDD_PADS, VDD_PIO and VDD_USB	(1.7V)	3.6V
Supply Voltage: VREG_IN	2.2V	4.2V <sup>(1)</sup>

Note:

<sup>(1)</sup> The device will operate without damage with VREG IN as high as 5.6V, however the RF performance is not guaranteed above 4.2V

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Min	Тур	Max	Unit
1.70	1.78	1.85	V
-250	-	+250	ppm/C
-	-	1	mV rms
-	-	50	mV/A
-	-	50	μS
140	-	-	mA
5	-	-	μA
-	-	4.2 <sup>(6)</sup>	V
-	-	350	mV
25	35	50	μA
	$\sim$	(0)	
4	7	10	μA
1.5	2.5	3.5	μA
	1.70 -250 - - - 140 5 - - 25 4	1.70 1.78   -250 -   - -   - -   140 -   5 -   - -   25 35   4 7	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$

Notes:

For optimum performance the VDD\_ANA ball adjacent to VREG\_IN should be used for regulator output

- <sup>(1)</sup> Regulator output connected to 47nF pure and  $4.7\mu F$  2.20 ESR capacitors.
- <sup>(2)</sup> Frequency range 100Hz to 100kHz
- <sup>(3)</sup> 1mA to 70mA pulsed load
- <sup>(4)</sup> Low power mode is entered and exited automatically when the chip enters/leaves Deep Sleep mode.
- <sup>(5)</sup> Regulator is disabled when VREG IN is either open circuit or driven to the same voltage as VDD\_ANA
- <sup>(6)</sup> Operation up to 5.6V is permissible without damage and without the output voltage rising sufficiently to damage the rest of BlueCore4, but output regulation and other specifications are no longer guaranteed at input voltages in excess of 4.2V.

Max

+0.8

VDD+0.4

0.2

0.4

-10

+100

-0.2

+5.0

+1

5.0

Unit

V

V

V

V

٧

V

٧

μA

μA

μA

μA

μA pF

Input/Output Terminal Cha	racteristics (Continued)		
Digital Terminals		Min	Тур
Input Voltage Levels			
VIL input logic level low	2.7V ≤ VDD ≤ 3.0V	-0.4	-
	1.7V ≤ VDD ≤ 1.9V	-0.4	-
V <sub>IH</sub> input logic level high		0.7VDD	-
Output Voltage Levels			
V <sub>OL</sub> output logic level low,			
$(I_0 = 4.0 \text{mA}), 2.7 \text{V} \le \text{VDD} \le 3$	.0V	-	-
V <sub>OL</sub> output logic level low,		_	_
$(I_o = 4.0 \text{mA}), 1.7 \text{V} \le \text{VDD} \le 1$	.9V		
V <sub>OH</sub> output logic level high,		VDD-0.2	-
$(I_o = -4.0 \text{mA}), 2.7 \text{V} \le \text{VDD} \le 3$	3.0V	100 0.2	
V <sub>OH</sub> output logic level high,		VDD-0.4	-
$(I_o = -4.0 \text{mA}), 1.7 \text{V} \leq \text{VDD} \leq$			0
Input and Tri-state Current	with:		6
Strong pull-up		-100	-4Ŏ(
Strong pull-down		+10	(+40
Weak pull-up		-5.0	-1.0
Weak pull-down		+0.2	) +1.0
I/O pad leakage current		$\langle \langle \langle \mathcal{L} \rangle \rangle \rangle$	0
C <sub>I</sub> Input Capacitance		1.0	-

	9		1	
USB Terminals	Min	Тур	Мах	Unit
VDD_USB for correct USB operation	3.1		3.6	V
Input threshold				
V <sub>IL</sub> input logic level low	-	-	0.3VDD_USB	V
$V_{H}$ input logic level high	0.7VDD_USB	-	-	V
Input leakage current				
VSS_PADS < $V_{IN}$ < $VDD_USB^{(1)}$	-1	1	5	μA
C <sub>I</sub> Input capacitance	2.5	-	10.0	pF
Output Voltage levels				
To correctly terminated USB Cable				
V <sub>OL</sub> output logic level low	0.0	-	0.2	V
V <sub>OH</sub> output logic level high	2.8	-	VDD_USB	V

Input/Output Terminal Characteristics (Continued)					
Power-on reset	Min	Тур	Мах	Unit	
VDD_CORE falling threshold	1.40	1.50	1.60	V	
VDD_CORE rising threshold	1.50	1.60	1.70	V	
Hysteresis	0.05	0.10	0.15	V	



Input/Output Terminal Ch	aracteristics (Co	ntinued)			
Auxiliary ADC		Min	Тур	Max	Unit
Resolution		-	-	8	Bits
Input voltage range (LSB size = VDD_ANA/255	)	0	-	VDD_ANA	V
Accuracy	INL	-1	-	1	LSB
(Guaranteed monotonic)	DNL	0	-	1	LSB
Offset		-1	-	1	LSB
Gain Error		-0.8	-	0.8	%
Input Bandwidth		-	100	-	kHz
Conversion time		-	2.5	-	μS
Sample rate <sup>(2)</sup>		-	-	700	Samples/s

Input/Output Terminal Character	istics (Continued)	)	$\diamond$		
Crystal Oscillator		Min	Тур	Max	Unit
Crystal frequency <sup>(4)</sup>		8.0		32.0	MHz
Digital trim range <sup>(5)</sup>		5.0	62	8.0	pF
Trim step size <sup>(5)</sup>			0.1	-	pF
Transconductance		2.0	)) <u> </u>	-	mS
Negative resistance <sup>(6)</sup>		<b>870</b>	1500	2400	Ω
External Clock					
Input frequency <sup>(7)</sup>	$\diamond$	7.5	-	40.0	MHz
Clock input level <sup>(8)</sup>		0.2	-	VDD_ANA	V pk-pk
Allowable Jitter	C	- <	-	15	ps rms
XTAL_IN input impedance	9/100	-	-	-	kΩ
XTAL_IN input capacitance	61102	-	7	-	pF

#### Notes:

VDD\_CORE, VDD\_RADIO, VDD\_LO and VDD\_ANA are at 1.8V unless shown otherwise

VDD\_PADS, VDD\_PIO and VDD\_USB are at 3.0V unless shown otherwise

The same setting of the digital trim is applied to both XTAL\_IN and XTAL\_OUT.

Current drawn into a pin is defined as positive; current supplied out of a pin is defined as negative.

- (1) Internal USB pull-up disabled
- <sup>(2)</sup> Access of ADC is through VM function and therefore sample rate given is achieved as part of this function
- <sup>(3)</sup> Specified for an output voltage between 0.2V and VDD\_PIO -0.2V
- <sup>(4)</sup> Integer multiple of 250kHz
- <sup>(5)</sup> The difference between the internal capacitance at minimum and maximum settings of the internal digital trim
- <sup>(6)</sup> XTAL frequency = 16MHz; XTAL C0 = 0.75pF; XTAL load capacitance = 8.5pF
- <sup>(7)</sup> Clock input can be any frequency between 8 and 40MHz in steps of 250kHz and also covers the CDMA/3G TCXO frequencies of 7.68, 14.44, 15.36, 16.2, 16.8, 19.2, 19.44, 19.68, 19.8 and 38.4MHz
- <sup>(8)</sup> Clock input can either be sinusoidal or square wave. If the peaks of the signal are below VSS\_ANA or above VDD\_ANA a DC blocking capacitor is required between the signal and XTAL\_IN



# 5 Radio Characteristics – Basic Data Rate

# 5.1 Temperature +20°C

# 5.1.1 Transmitter

Radio Characteristics	VDD = 1.8V Tem	perature	= +20°C			
		Min	Тур	Мах	Bluetooth Specification	Unit
Maximum RF transmit power	r <sup>(1)(2)(3)</sup>	3	6.5	-	-6 to +4 <sup>(4)</sup>	dBm
RF power control range <sup>(1)(2)</sup>		25	35	-	≥16	dB
RF power range control resolution		-	0.5	-	-	dB
20dB bandwidth for modulate	ed carrier	-	820	1000	≤1000	kHz
Adjacent channel transmit po	ower $F=F_0 \pm 2MHz^{(5)}$	-	-35	-20	≤-20	dBm
Adjacent channel transmit po	ower $F=F_0 \pm 3MHz^{(5)}$	-	-45	-40	_ ≨-40	dBm
∆f1avg "Maximum Modulatio	n"	140	165	175	140 <f1avg<175< td=""><td>kHz</td></f1avg<175<>	kHz
∆f2max "Minimum Modulatio	n"	115	140	-	145	kHz
∆f1avg/∆f2avg		0.8	0.9	-	0≥0.80	-
Initial carrier frequency tolera	ance	-	10	35	±75	kHz
Drift Rate		-	8	20	≤20	kHz/ 50μs
Drift (single slot packet)		-	9	20	≤25	kHz
Drift (five slot packet)		-	190/	25	≤40	kHz
Emissions	Frequency (GHz)	Min	Тур	Max	Specification	Unit
Emitted power in cellular bands measured at chip	0.925-0.960		-143	-138	Integrated in 200kHz bandwidth	dBMHz
terminals Output power ≤4dBm	1.570-1.580		-138	-135	Integrated in 1MHz bandwidth	dBMHz
	1.805-1.880	-	-131	-115	Integrated in 200kHz bandwidth	dBMHz
<	1.930-1.990	-	-135	-125	Integrated in 30kHz bandwidth	dBMHz
	1.930-1.990	-	-135	-126	Integrated in 200kHz bandwidth	dBMHz
	1.930-1.990	-	-137	-130	Integrated in 1.2MHz bandwidth	dBMHz
	2.110-2.170	-	-132	-122	Integrated in 1.2MHz bandwidth	dBMHz
	2.110-2.170	-	-135	-127	Integrated in 5MHz bandwidth	dBMHz

#### Notes:

Results shown are referenced to input of the RF balun

- <sup>(1)</sup> Power at the chip pads
- <sup>(2)</sup> Measured according to the Bluetooth specification v1.2
- <sup>(3)</sup> The firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits
- <sup>(4)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- $^{(5)}$  Measured at F<sub>0</sub> = 2441MHz



# 5.1.2 Receiver

Radio Characteristics VD	D = 1.8V Tem	perature	= +20°C			
	Frequency (GHz)	Min	Тур	Мах	Bluetooth Specification	Unit
	2.402	-	-84	-80		dBm
Sensitivity at 0.1% BER for all packet types	2.441	-	-85	-81	≤-70	dBm
	2.480	-	-85	-81		dBm
Maximum received signal at 0.1	% BER	0	3	-	≥-20	dBm
C/l co-channel		-	9	11	≤11	dB
Adjacent channel selectivity C/I	$F=F_0+1MHz^{(1)(2)}$	-	-4	0	≤0	dB
Adjacent channel selectivity C/I	F=F <sub>0</sub> -1MHz <sup>(1)(2)</sup>	-	-4	0	≤0	dB
Adjacent channel selectivity C/I	F=F <sub>0</sub> +2MHz <sup>(1)(2)</sup>	-	-35	-30	≤-30	dB
Adjacent channel selectivity C/I	F=F <sub>0</sub> -2MHz <sup>(1)(2)</sup>	-	-21	-20	≤~20	dB
Adjacent channel selectivity C/I $F \ge F_0 + 3MHz^{(1)(2)}$		-	-45	-	♦ (≤-40)	dB
	Adjacent channel selectivity C/I $F \le F_0 - 5MHz^{(1)(2)}$		-45	-	_40	dB
Adjacent channel selectivity C/I	Adjacent channel selectivity C/I F=F <sub>Image</sub> <sup>(1)(2)</sup>		-18	-9	≤-9	dB
Maximum level of intermodulation		-	-30	-39	≥-39	dBm
Spurious output level <sup>(4)</sup>		-	-140	<u> (6-//))</u>	-	dBm/Hz
Blocking	Frequency (GHz)	Min	Тур	Max	Modulation	Unit
Continuous power in cellular	0.824-0.849 <sup>(5)</sup>	-2	2	-	GSM	dBm
bands required to block Bluetooth reception (for	0.880-0.915	5	$\langle \gamma \rangle$	-	GSM	dBm
Bluetooth sensitivity of	1.710-1.785		<u>کہ</u> 6	-	GSM	dBm
-67dBm with 0.1% BER)	1.850-1.910		5	-	GSM	dBm
Measured at chip terminals	1.920-1.980	>>8	-6	-	W_CDMA	dBm
Continuous power in cellular	0.824-0.849	-10	-5	-	GSM	dBm
bands required to block	0.880-0.915	-8	-4	-	GSM	dBm
Bluetooth reception (for sensitivity of -80dBm with	1,710-1.785	-7	-3	-	GSM	dBm
0.1% BER) measured at chip	1.850-1.910	-11	-4	-	GSM	dBm
terminals	1.920-1.980	-18	-14	-	W_CDMA	dBm

#### Notes:

Results shown are referenced to input of the RF balun

- <sup>(1)</sup> Up to five exceptions are allowed in v1.2 of the Bluetooth specification. BlueCore4-External is guaranteed to meet the C/I performance as specified by the Bluetooth specification v1.2.
- (2) Measured at  $F_0$  = 2405MHz, 2441MHz, 2477MHz
- <sup>(3)</sup> Measured at f1-f2 = 5MHz. Measurement is performed in accordance with Bluetooth RF test RCV/CA/05/c. i.e. wanted signal at -64dBm
- <sup>(4)</sup> Integrated in 100kHz bandwidth. Actual figure is typically below -140dBm/Hz except for peaks of -125dBm/Hz at 1.2GHz and -100dBm/Hz in-band at 2.4GHz
- <sup>(5)</sup>  $| 3f_{Blocking} f_{Bluetooth} | > 4MHz$



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# 5.2 Temperature -40°C

# 5.2.1 Transmitter

	Min	Тур	Max	Bluetooth Specification	Unit
Maximum RF transmit power <sup>(1)</sup>	4	8	-	-6 to +4 <sup>(2)</sup>	dBm
RF power control range	25	35	-	≥16	dB
RF power range control resolution	-	0.5	-	-	dB
20dB bandwidth for modulated carrier	-	820	1000	≤1000	kHz
Adjacent channel transmit power $F=F_0 \pm 2MHz^{(3)}$ <sup>(4)</sup>	-	-35	-20	≤-20	dBm
Adjacent channel transmit power $F=F_0 \pm 3MHz^{(3)}$ <sup>(4)</sup>	-	-45	-40	≤-40	dBm
$\Delta$ f1avg "Maximum Modulation"	140	165	175	140<∆f1 <sub>avg</sub> <175	kHz
$\Delta$ f2max "Minimum Modulation"	115	135	-	$\left( 115 \right)$	kHz
∆f2avg / ∆f1avg	0.8	0.9	- (	≷0.80	-
Initial carrier frequency tolerance	-	10	35	±75	kHz
Drift Rate	-	8	25	≤25	kHz/50μ
Drift (single slot packet)	-	9	25	≤25	kHz
Drift (five slot packet)	-	05	40	≤40	kHz

Notes:

<sup>(1)</sup> BlueCore4-External firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits

- <sup>(2)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- $^{(3)}$  Measured at F<sub>0</sub> = 2441MHz
- $^{(4)}$  Up to three exceptions are allowed in v1.2 of the Bluetooth specification

# 5.2.2 Receiver

Radio Characteristics VDD = 1.8V Temperature = -40°C							
A COLIN	Frequency (GHz)	Min	Тур	Мах	Bluetooth Specification	Unit	
	2.402	-	-86.0	-81	≤-70		
Sensitivity at 0.1% BER for all packet types	2.441	-	-88.0	-82		dBm	
()poo	2.480	-	-86.5	-82			
Maximum received signal at 0.1% BER		-2	1	-	≥-20	dBm	

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# 5.3 Temperature -25°C

# 5.3.1 Transmitter

	Min	Тур	Max	Bluetooth Specification	Unit
Maximum RF transmit power <sup>(1)</sup>	3.5	7	-	-6 to +4 <sup>(2)</sup>	dBm
RF power control range	25	35	-	≥16	dB
RF power range control resolution	-	0.5	-	-	dB
20dB bandwidth for modulated carrier	-	820	1000	≤1000	kHz
Adjacent channel transmit power $F=F_0 \pm 2MHz^{(3)}$ <sup>(4)</sup>	-	-35	-20	≤-20	dBm
Adjacent channel transmit power $F=F_0 \pm 3MHz^{(3)}$ <sup>(4)</sup>	-	-45	-40	≤-40	dBm
∆f1avg "Maximum Modulation"	140	165	175	140<∆f1 <sub>avg</sub> <175	kHz
∆f2max "Minimum Modulation"	115	140	-	$\left( 115 \right)$	kHz
∆f2avg / ∆f1avg	0.8	0.9	-	≷0.80	-
Initial carrier frequency tolerance	-	10	35	±75	kHz
Drift Rate	-	8	~ 20	≤25	kHz/50μ
Drift (single slot packet)	-	9	20	≤25	kHz
Drift (five slot packet)	-	(~10C)	25	≤40	kHz

Notes:

<sup>(1)</sup> BlueCore4-External firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits

- <sup>(2)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- $^{(3)}$  Measured at F<sub>0</sub> = 2441MHz
- $^{(4)}$  Up to three exceptions are allowed in v1.2 of the Bluetooth specification

# 5.3.2 Receiver

Radio Characteristics VDD = 1.8V Temperature = -25°C							
A CONT	Frequency (GHz)	Min	Тур	Max	Bluetooth Specification	Unit	
	2.402	-	-85.5	-81	≤-70	dBm	
Sensitivity at 0.1% BER for all packet types	2.441	-	-86.5	-82			
,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,	2.480	-	-86.5	-82			
Maximum received signal at 0.1% BER		-2	1	-	≥-20	dBm	



# 5.4 Temperature +85°C

# 5.4.1 Transmitter

	Min	Тур	Max	Bluetooth Specification	Unit
Maximum RF transmit power <sup>(1)</sup>	0	3	-	-6 to +4 <sup>(2)</sup>	dBm
RF power control range	25	35	-	≥16	dB
<sup>C</sup> RF power range control resolution	-	0.5	-	-	dB
20dB bandwidth for modulated carrier	-	820	1000	≤1000	kHz
Adjacent channel transmit power $F=F_0 \pm 2MHz^{(3)}$ <sup>(4)</sup>	-	-35	-20	≤-20	dBm
Adjacent channel transmit power $F=F_0 \pm 3MHz^{(3)}$ <sup>(4)</sup>	-	-45	-40	≤-40	dBm
∆f1avg "Maximum Modulation"	140	165	175	140<∆f1 <sub>avg</sub> <175	kHz
∆f2max "Minimum Modulation"	115	140	-	$\left( 115 \right)$	kHz
∆f2avg / ∆f1avg	0.8	0.9	-	≷0.80	-
Initial carrier frequency tolerance	-	10	35	±75	kHz
Drift Rate	-	9	20	≤25	kHz/50µs
Drift (single slot packet)	-	9	20	≤25	kHz
Drift (five slot packet)	-	(7) (C	28	≤40	kHz

Notes:

<sup>(1)</sup> BlueCore4-External firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits

- <sup>(2)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- $^{(3)}$  Measured at F<sub>0</sub> = 2441MHz
- $^{(4)}$  Up to three exceptions are allowed in v1.2 of the Bluetooth specification

# 5.4.2 Receiver

Radio Characteristics VDD = 1.8V Temperature = +85°C							
	Frequency (GHz)	Min	Тур	Мах	Bluetooth Specification	Unit	
	2.402	-	-81	-77			
Sensitivity at 0.1% BER for all packet types	2.441	-	-83	-79	≤-70	dBm	
	2.480	-	-83	-79			
Maximum received signal at 0.1% BER		0	5	-	≥-20	dBm	



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# 5.5 Temperature +105°C

# 5.5.1 Transmitter

	Min	Тур	Max	Bluetooth Specification	Unit
Maximum RF transmit power <sup>(1)</sup>	-2.5	1	-	-6 to +4 <sup>(2)</sup>	dBm
RF power control range	25	35	-	≥16	dB
RF power range control resolution	-	0.5	-	-	dB
20dB bandwidth for modulated carrier	-	820	1000	≤1000	kHz
Adjacent channel transmit power $F=F_0 \pm 2MHz^{(3)}$ <sup>(4)</sup>	-	-35	-20	≤-20	dBm
Adjacent channel transmit power $F=F_0 \pm 3MHz^{(3)}$ <sup>(4)</sup>	-	-45	-40	≤-40	dBm
∆f1avg "Maximum Modulation"	140	165	175	140<∆f1 <sub>ev9</sub> <175	kHz
∆f2max "Minimum Modulation"	-	135	-	$\left( 115\right)$	kHz
∆f2avg / ∆f1avg	-	0.9	-	≷0.80	-
Initial carrier frequency tolerance	-	10	35	±75	kHz
Drift Rate	-	9	25	≤25	kHz/50μs
Drift (single slot packet)	-	9	25	≤25	kHz
Drift (five slot packet)	-	570	40	≤40	kHz

Notes:

<sup>(1)</sup> BlueCore4-External firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits

- <sup>(2)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- $^{(3)}$  Measured at F<sub>0</sub> = 2441MHz
- $^{(4)}$  Up to three exceptions are allowed in v1.2 of the Bluetooth specification

# 5.5.2 Receiver

Radio Characteristics VDD = 1.8V Temperature = +105°C							
	Frequency (GHz)	Min	Тур	Мах	Bluetooth Specification	Unit	
Sensitivity at 0.1% BER for all packet types	2.402	-	-81	-77			
	2.441	-	-82	-78	≤-70	dBm	
	2.480	-	-82	-78			
Maximum received signal at 0.1% BER		0	5	-	≥-20	dBm	



# 5.6 Power Consumption

#### Typical Average Current Consumption

Mode	Average	Unit
SCO connection HV3 (30ms interval Sniff Mode) (Slave)	21	mA
SCO connection HV3 (30ms interval Sniff Mode) (Master)	21	mA
SCO connection HV3 (No Sniff Mode) (Slave)	28	mA
SCO connection HV1 (Slave)	42	mA
SCO connection HV1 (Master)	42	mA
ACL data transfer 115.2kbps UART no traffic (Master)	5	mA
ACL data transfer 115.2kbps UART no traffic (Slave)	22	mA
ACL data transfer 720kbps UART (Master or Slave)	45	mA
ACL data transfer 720kbps USB (Master or Slave)	45 🛇 (	mA
ACL connection, Sniff Mode 40ms interval, 38.4kbps UART	3.2	mA
ACL connection, Sniff Mode 1.28s interval, 38.4kbps UART	0.45	mA
Parked Slave, 1.28s beacon interval, 38.4kbps UART	0.55	mA
Standby Mode (Connected to host, no RF activity)	(47.0)	μΑ
Reset (RESETB low)	( ( <b>15</b> 0	μA
	$\mathbf{\mathbf{\hat{\mathbf{A}}}}$	



# 6 Radio Characteristics – Enhanced Data Rate

# 6.1 Temperature +20°C

## 6.1.1 Transmitter

	Radio Characteristics	VDD = 1.8V	Temperature = +20°C						
			Min	Тур	Мах	Bluetooth Specification	Unit		
40	Maximum RF transmit pov	ver <sup>(1)</sup>		6		-6 to +4 <sup>(2)</sup>	dBm		
	Relative transmit power <sup>(3)</sup>			-1		-4 to +1	dB		
	Carrier frequency stability	3)		3		≤10	kHz		
		RMS DEVM		10		≤13 <sup>(5)</sup>	%		
	Modulation Accuracy <sup>(3)(4)</sup>	99% DEVM		15		≤20 <sup>(5)</sup>	%		
		Peak DEVM		20		25(5)	%		

#### Notes:

Results shown are referenced to input of the RF balun

- <sup>(1)</sup> BlueCore4-External firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits
- <sup>(2)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- <sup>(3)</sup> Measurements methods are in accordance with the EDR RF Test Specification v0.9
- <sup>(4)</sup> Modulation accuracy utilises differential error vector magnitude (DEVM) with tracking of the carrier frequency drift.
- <sup>(5)</sup> The Bluetooth specification values are for 8DPSK modulation (values for the  $\pi/4$  DQPSK modulation are less stringent)



# 6.1.2 Receiver

Radio Characteristics VDD = 1.8V Temperature = +20°C									
	Modulation	Min	Тур	Мах	Bluetooth Specification	Unit			
Sensitivity at 0.01% BER <sup>(1)</sup>	π/4 DQPSK		-86		≤-70	dBm			
Sensitivity at 0.01% BER	8DPSK		-79		≤-70	dBm			
Maximum received signal at	π/4 DQPSK		-6		≥-20	dBm			
0.1% BER <sup>(1)</sup>	8DPSK		-7		≥-20	dBm			
C/I co-channel at 0.1% BER <sup>(1)</sup>	π/4 DQPSK		+11		≤+13	dB			
C/I CO-Channel at 0.1% DER	8DPSK		+19		≤+21	dB			
Adjacent channel selectivity C/I F=F <sub>0</sub> +1MHz <sup>(1)(2)(3)</sup>	π/4 DQPSK		-8		≤0	dB			
C/I F=F <sub>0</sub> +1MHz <sup>(1)(2)(3)</sup>	8DPSK		-2		≤+5	dB			
Adjacent channel selectivity C/I F=F <sub>0</sub> -1MHz <sup>(1)(2)(3)</sup>	π/4 DQPSK		-8			dB			
C/I F=F <sub>0</sub> -1MHz <sup>(1)(2)(3)</sup>	8DPSK		-2		<u> </u>	dB			
Adjacent channel selectivity C/I F=F <sub>0</sub> +2MHz <sup>(1)(2)(3)</sup>	π/4 DQPSK		-35	G	30	dB			
$C/I F = F_0 + 2MHz^{(1)(2)(3)}$	8DPSK		-35			dB			
Adjacent channel selectivity	π/4 DQPSK		-23		≤-20	dB			
C/I F=F <sub>0</sub> -2MHz <sup>(1)(2)(3)</sup>	8DPSK		-19	$\mathcal{M}$	≤-13	dB			
Adjacent channel selectivity C/I F $\ge$ F <sub>0</sub> +3MHz <sup>(1)(2)(3)</sup>	π/4 DQPSK		-43		≤-40	dB			
C/I F≥F₀ +3MHz <sup>(1)(2)(3)</sup>	8DPSK	<	-40	$\geq$	≤-33	dB			
Adjacent channel selectivity C/I F≤F₀ –5MHz <sup>(1)(2)(3)</sup>	π/4 DQPSK		-43		≤-40	dB			
C/I F≤F₀ –5MHz <sup>(1)(2)(3)</sup>	8DPSK		-38		≤-33	dB			
Adjacent channel selectivity	π/4 DQPSK	$\langle (O) \rangle$	-17		≤-7	dB			
Adjacent channel selectivity C/I F= $F_{Image}^{(1)(2)(3)}$	8DPSK		-10		≤0	dB			

#### Notes:

Results shown are referenced to input of the RF balun

- <sup>(1)</sup> Measurements methods are in accordance with the EDR RF Test Specification v0.9
- <sup>(2)</sup> Up to five exceptions are allowed in EDR RF Test Specification v0.9. BlueCore4-External is guaranteed to meet the C/I performance as specified by the EDR RF Test Specification v0.9.
- (3) Measured at (F) = 2405MHz, 2441MHz, 2477MHz



# 6.2 Temperature -40°C

## 6.2.1 Transmitter

Radio Characteristics	VDD = 1.8V	Temperature = -40°C						
		Min	Тур	Max	Bluetooth Specification	Unit		
Maximum RF transmit power <sup>(1)</sup>			8		-6 to +4 <sup>(2)</sup>	dBm		
Relative transmit power <sup>(3)</sup>			-1		-4 to +1	dB		
Carrier frequency stability	Carrier frequency stability <sup>(3)</sup>		3		≤10	kHz		
	RMS DEVM		10		≤13 <sup>(5)</sup>	%		
Modulation Accuracy <sup>(3)(4)</sup>	99% DEVM		15		≤20 <sup>(5)</sup>	%		
	Peak DEVM		20		≤25 <sup>(5)</sup>	%		

#### Notes:

Results shown are referenced to input of the RF balun

- <sup>(1)</sup> BlueCore4-External firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits
- <sup>(2)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- <sup>(3)</sup> Measurements methods are in accordance with the EDR RF Test Specification v0.9
- <sup>(4)</sup> Modulation accuracy utilises differential error vector magnitude (DEVM) with tracking of the carrier frequency drift.
- <sup>(5)</sup> The Bluetooth specification values are for 8DPSK-modulation (values for the  $\pi/4$  DQPSK modulation are less stringent)

## 6.2.2 Receiver

Radio Characteristics VDD = 1.8V Temperature = -40°C									
Modulation	Min	Тур	Мах	Bluetooth Specification	Unit				
π/4 DQPSK		-89		≤-70	dBm				
8DPSK		-82		≤-70	dBm				
$\rightarrow$ $\pi$ /4 DQPSK		-10		≥-20	dBm				
8DPSK		-10		≥-20	dBm				
	Modulation π/4 DQPSK 8DPSK π/4 DQPSK	Modulation Min   π/4 DQPSK   8DPSK   π/4 DQPSK	Modulation     Min     Typ       π/4 DQPSK     -89       8DPSK     -82       π/4 DQPSK     -10	Modulation Min Typ Max   π/4 DQPSK -89   8DPSK -82   π/4 DQPSK -10	ModulationMinTypMaxBluetooth Specification $\pi/4$ DQPSK-89 $\leq$ -708DPSK-82 $\leq$ -70 $\pi/4$ DQPSK-10 $\geq$ -20				

#### Notes:

Results shown are referenced to input of the RF balun

<sup>(1)</sup> Measurements methods are in accordance with the EDR RF Test Specification v0.9



# 6.3 Temperature -25°C

## 6.3.1 Transmitter

Radio Characteristics	VDD = 1.8V	Temperature = -25°C						
		Min	Тур	Max	Bluetooth Specification	Unit		
Maximum RF transmit power <sup>(1)</sup>			7		-6 to +4 <sup>(2)</sup>	dBm		
Relative transmit power <sup>(3)</sup>			-1		-4 to +1	dB		
Carrier frequency stability <sup>(3)</sup>			3		≤10	kHz		
	RMS DEVM		10		≤13 <sup>(5)</sup>	%		
Modulation Accuracy <sup>(3)(4)</sup>	99% DEVM		15		≤20 <sup>(5)</sup>	%		
	Peak DEVM		20		≤25 <sup>(5)</sup>	%		

#### Notes:

Results shown are referenced to input of the RF balun

- <sup>(1)</sup> BlueCore4-External firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits
- <sup>(2)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- <sup>(3)</sup> Measurements methods are in accordance with the EDR RF Test Specification v0.9
- <sup>(4)</sup> Modulation accuracy utilises differential error vector magnitude (DEVM) with tracking of the carrier frequency drift.
- <sup>(5)</sup> The Bluetooth specification values are for 8DPSK-modulation (values for the  $\pi/4$  DQPSK modulation are less stringent)

## 6.3.2 Receiver

Radio Characteristics VDD = 1.8V Temperature = -25°C									
	Modulation	Min	Тур	Мах	Bluetooth Specification	Unit			
Sensitivity at 0.01% BER <sup>(1)</sup>	π/4 DQPSK		-87		≤-70	dBm			
Sensitivity at 0.01% BER	8DPSK		-80		≤-70	dBm			
Maximum received signal at	☆ π/4 DQPSK		-10		≥-20	dBm			
0.1% BER <sup>(1)</sup>	8DPSK		-10		≥-20	dBm			

#### Notes:

Results shown are referenced to input of the RF balun

<sup>(1)</sup> Measurements methods are in accordance with the EDR RF Test Specification v0.9



# 6.4 Temperature +85°C

## 6.4.1 Transmitter

Radio Characteristics	VDD = 1.8V	Temperature = +85°C						
		Min	Тур	Max	Bluetooth Specification	Unit		
Maximum RF transmit power <sup>(1)</sup>			1		-6 to +4 <sup>(2)</sup>	dBm		
Relative transmit power <sup>(3)</sup>			-1		-4 to +1	dB		
Carrier frequency stability <sup>(3)</sup>			3		≤10	kHz		
	RMS DEVM		10		≤13 <sup>(5)</sup>	%		
Modulation Accuracy <sup>(3)(4)</sup>	99% DEVM		15		≤20 <sup>(5)</sup>	%		
	Peak DEVM		20		≤25 <sup>(5)</sup>	%		

#### Notes:

Results shown are referenced to input of the RF balun

- <sup>(1)</sup> BlueCore4-External firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits
- <sup>(2)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- <sup>(3)</sup> Measurements methods are in accordance with the EDR RF Test Specification v0.9
- <sup>(4)</sup> Modulation accuracy utilises differential error vector magnitude (DEVM) with tracking of the carrier frequency drift.
- <sup>(5)</sup> The Bluetooth specification values are for 8DPSK-modulation (values for the  $\pi/4$  DQPSK modulation are less stringent)

## 6.4.2 Receiver

Radio Characteristics VDD = 1.8V Temperature = +85°C									
Modulation	Min	Тур	Мах	Bluetooth Specification	Unit				
π/4 DQPSK		-84		≤-70	dBm				
8DPSK		-77		≤-70	dBm				
$\rightarrow$ $\pi$ /4 DQPSK		0		≥-20	dBm				
8DPSK		-3		≥-20	dBm				
	Modulation π/4 DQPSK 8DPSK π/4 DQPSK	Modulation Min   π/4 DQPSK   8DPSK   π/4 DQPSK	Modulation     Min     Typ       π/4 DQPSK     -84       8DPSK     -77       π/4 DQPSK     0	Modulation Min Typ Max   π/4 DQPSK -84   8DPSK -77   π/4 DQPSK 0	ModulationMinTypMaxBluetooth Specification $\pi/4$ DQPSK-84 $\leq$ -708DPSK-77 $\leq$ -70 $\pi/4$ DQPSK0 $\geq$ -20				

#### Notes:

Results shown are referenced to input of the RF balun

<sup>(1)</sup> Measurements methods are in accordance with the EDR RF Test Specification v0.9



#### 6.5 Temperature +105°C

#### 6.5.1 Transmitter

Radio Characteristics	VDD = 1.8V	Temperature = +105°C				
		Min	Тур	Max	Bluetooth Specification	Unit
Maximum RF transmit power <sup>(1)</sup>			-1		-6 to +4 <sup>(2)</sup>	dBm
Relative transmit power <sup>(1)</sup>			-1		-4 to +1	dB
Carrier frequency stability <sup>(1)</sup>			3		≤10	kHz
	RMS DEVM		10		≤13 <sup>(5)</sup>	%
Modulation Accuracy <sup>(1)(2)</sup>	99% DEVM		15		≤20 <sup>(5)</sup>	%
	Peak DEVM		20		≤25 <sup>(5)</sup>	%

#### Notes:

Results shown are referenced to input of the RF balun

- <sup>(1)</sup> BlueCore4-External firmware maintains the transmit power to be within the Bluetooth specification v1.2 limits
- <sup>(2)</sup> Class 2 RF transmit power range, Bluetooth specification v1.2
- <sup>(3)</sup> Measurements methods are in accordance with the EDR RF Test Specification v0.9
- <sup>(4)</sup> Modulation accuracy utilises differential error vector magnitude (DEVM) with tracking of the carrier frequency drift.
- <sup>(5)</sup> The Bluetooth specification values are for 8DPSK-modulation (values for the  $\pi/4$  DQPSK modulation are less stringent)

#### 6.5.2 Receiver

Radio Characteristics VDD = 1.8V Temperature = +105°C						
	Modulation	Min	Тур	Мах	Bluetooth Specification	Unit
Sensitivity at 0.01% BER <sup>(1)</sup>	π/4 DQPSK		-83		≤-70	dBm
	8DPSK		-76		≤-70	dBm
Maximum received signal at 0.1% BER <sup>(1)</sup>	π/4 DQPSK		0		≥-20	dBm
	8DPSK		-1		≥-20	dBm

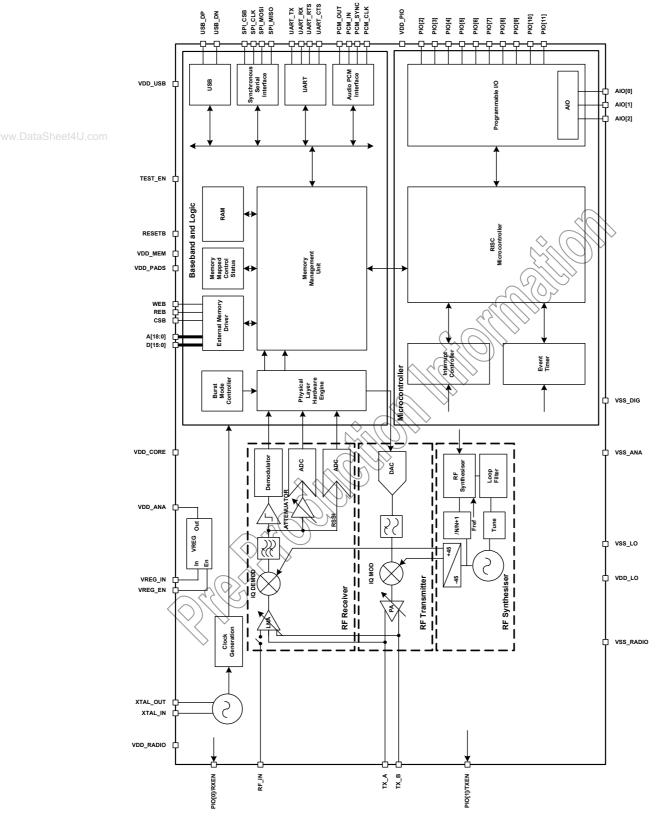
#### Notes:

Results shown are referenced to input of the RF balun

 $^{(1)}$  Measurements methods are in accordance with the EDR RF Test Specification v0.9



# 7 Device Diagram





# 8 Description of Functional Blocks

## 8.1 RF Receiver

The receiver features a near-zero Intermediate Frequency (IF) architecture that allows the channel filters to be integrated on to the die. Sufficient out-of-band blocking specification at the Low Noise Amplifier (LNA) input allows the radio to be used in close proximity to Global System for Mobile Communications (GSM) and Wideband Code Division Multiple Access (W-CDMA) cellular phone transmitters without being desensitised. The use of a digital Frequency Shift Keying (FSK) discriminator means that no discriminator tank is needed and its excellent performance in the presence of noise allows BlueCore4-External to exceed the Bluetooth requirements for co-channel and adjacent channel rejection.

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For EDR, an ADC is used to digitise the IF received signal.

#### 8.1.1 Low Noise Amplifier

The LNA can be configured to operate in single-ended or differential mode. Single-ended mode is used for Class 1 Bluetooth operation; differential mode is used for Class 2 operation.

## 8.1.2 Analogue to Digital Converter

The Analogue to Digital Converter (ADC) is used to implement fast Automatic Gain Control (AGC). The ADC samples the Received Signal Strength Indicator (RSSI) voltage on a slot-by slot basis. The front-end LNA gain is changed according to the measured RSSI value, keeping the first mixer input signal within a limited range. This improves the dynamic range of the receiver, improving performance in interference limited environments.

### 8.2 RF Transmitter

#### 8.2.1 IQ Modulator

The transmitter features a direct IQ modulator to minimise the frequency drift during a transmit timeslot, which results in a controlled modulation index. Digital baseband transmit circuitry provides the required spectral shaping.

### 8.2.2 Power Amplifier

The internal Power Amplifier (PA) has a maximum output power of +6dBm allowing BlueCore4-External to be used in Class 2 and Class 3 radios without an external RF PA. Support for transmit power control allows a simple implementation for Class 1 with an external RF PA.

# 8.3 RF Synthesiser

The radio synthesiser is fully integrated onto the die with no requirement for an external Voltage Controlled Oscillator (VCO) screening can, varactor tuning diodes, LC resonators or loop filter. The synthesiser is guaranteed to lock in sufficient time across the guaranteed temperature range to meet the Bluetooth specification v1.2.

### 8.4 Clock Input and Generation

The reference clock for the system is generated from a TCXO or crystal input between 8 and 40MHz. All internal reference clocks are generated using a phase locked loop, which is locked to the external reference frequency.



### 8.5 Baseband and Logic

#### 8.5.1 Memory Management Unit

The Memory Management Unit (MMU) provides a number of dynamically allocated ring buffers that hold the data which is in transit between the host and the air. The dynamic allocation of memory ensures efficient use of the available Random Access Memory (RAM) and is performed by a hardware MMU to minimise the overheads on the processor during data/voice transfers.

### 8.5.2 Burst Mode Controller

www.DataSheet4U.During radio transmission the Burst Mode Controller (BMC) constructs a packet from header information previously loaded into memory-mapped registers by the software and payload data/voice taken from the appropriate ring buffer in the RAM. During radio reception, the BMC stores the packet header in memory-mapped registers and the payload data in the appropriate ring buffer in RAM. This architecture minimises the intervention required by the processor during transmission and reception.

### 8.5.3 Physical Layer Hardware Engine DSP

Dedicated logic is used to perform the following:

- Forward error correction
- Header error control
- Cyclic redundancy check
- Encryption
- Data whitening
- Access code correlation
- Audio transcoding

The following voice data translations and operations are performed by firmware:

- A-law/µ-law/linear voice data (from host)
- A-law/μ-law/Continuously Variable Slope Delta (CVSD) (over the air)
- Voice interpolation for lost packets
- Rate mismatches

The hardware supports all optional and mandatory features of Bluetest v1.2 including AFH and eSCO.

# 8.5.4 RAM

48Kbytes of on-chip RAM is provided to support the RISC MCU and is shared between the ring buffers used to hold voice/data for each active connection and the general purpose memory required by the Bluetooth stack.

#### 8.5.5 External Memory Driver

The External Memory Driver interface can be used to connect to the external Flash memory and also to the optional external RAM for memory intensive applications.

### 8.5.6 USB

This is a full speed Universal Serial Bus (USB) interface for communicating with other compatible digital devices. BlueCore4-External acts as a USB peripheral, responding to requests from a Master host controller such as a PC.



# 8.5.7 Synchronous Serial Interface

This is a synchronous serial port interface (SPI) for interfacing with other digital devices. The SPI port can be used for system debugging. It can also be used for programming the Flash memory.

### 8.5.8 UART

This is a standard Universal Asynchronous Receiver Transmitter (UART) interface for communicating with other serial devices.

# 8.5.9 Audio PCM Interface

www.DataSheet4U.com The audio pulse code modulation (PCM) Interface supports continuous transmission and reception of PCM encoded audio data over Bluetooth.

## 8.6 Microcontroller

The microcontroller (MCU), interrupt controller and event timer run the Bluetooth software stack and control the radio and host interfaces. A 16-bit reduced instruction set computer (RISC) microcontroller is used for low power consumption and efficient use of memory.

### 8.6.1 Programmable I/O

BlueCore4-External has a total of 15 (12 digital and 3 analogue) programmable I/O terminals. These are controlled by firmware running on the device.

### 8.6.2 802.11 Coexistence Interface

Dedicated hardware is provided to implement a variety of coexistence schemes. Channel skipping AFH, priority signalling, channel signalling and host passing of channel instructions are all supported. The features are configured in firmware. Since the details of some methods are proprietary (e.g. Intel WCS) please contact CSR for details.



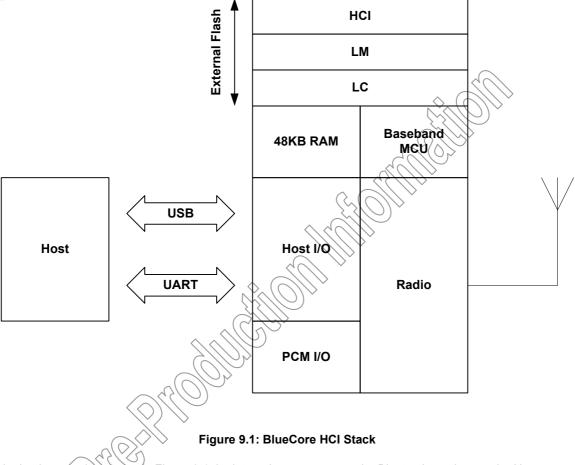
# 9 CSR Bluetooth Software Stacks

BlueCore4-External is supplied with Bluetooth v1.2 compliant stack firmware, which runs on the internal RISC microcontroller.

The BlueCore4-External software architecture allows Bluetooth processing and the application program to be shared in different ways between the internal RISC microcontroller and an external host processor (if any). The upper layers of the Bluetooth stack (above HCI) can be run either on-chip or on the host processor.

#### 9.1 BlueCore HCI Stack

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In the implementation shown in Figure 9.1 the internal processor runs the Bluetooth stack up to the Host Controller Interface (HCI). The Host processor must provide all upper layers including the application.



#### 9.1.1 Key Features of the HCI Stack - Standard Bluetooth Functionality

#### New Bluetooth v1.2 Mandatory Functionality:

- Adaptive frequency hopping (AFH), including classifier
- Faster connection enhanced inquiry scan (immediate FHS response)
- LMP improvements
- Parameter ranges

#### Optional v1.2 functionality supported:

- DataSheet4U.com Adaptive Frequency Hopping (AFH) as Master and Automatic Channel Classification
  - Fast Connect Interlaced Inquiry and Page Scan plus RSSI during Inquiry
  - Extended SCO (eSCO), eV3 +CRC, eV4, eV5
  - SCO handle
  - Synchronisation

#### The firmware has been written against the Bluetooth Core Specification v1.2.

- Bluetooth components:
  - Baseband (including LC)
  - LM
  - HCI
- Standard USB v2.0 (full speed) and UART HCI Transport Layers
- All standard radio packet types
- Full Bluetooth data rate, enhanced data rates of 2 and 3Mbps<sup>(1)</sup>
- Operation with up to seven active slaves<sup>(1)</sup>
- Scatternet v2.5 operation
- Maximum number of simultaneous active ACL connections: 7<sup>(2)</sup>
- Maximum number of simultaneous active SCO connections: 3<sup>(2)</sup>
- Operation with up to three SCO links, routed to one or more slaves
- All standard SCO voice coding, plus "transparent SCO"
- Standard operating modes: page, inquiry, page-scan and inquiry-scan
- All standard pairing, authentication, link key and encryption operations
- Standard Bluetooth power saving mechanisms: Hold, Sniff and Park modes, including "Forced Hold"
- Dynamic control of peers' transmit power via LMP
- Master/Slave switch
- Broadcast
- Channel quality driven data rate
- All standard Bluetooth Test Modes

The firmware's supported Bluetooth features are detailed in the standard Protocol Implementation Conformance Statement (PICS) documents, available from <a href="http://www.csr.com">http://www.csr.com</a>.



Note:

- <sup>(1)</sup> Supports basic data rate up to 723.2kbps asymmetric, maximum allowed by Bluetooth specification v1.2
- <sup>(2)</sup> BlueCore4-External supports all combinations of active ACL and SCO channels for both Master and Slave operation, as specified by the Bluetooth specification v1.2

### 9.1.2 Key Features of the HCI Stack - Extra Functionality

The firmware extends the standard Bluetooth functionality with the following features:

 Supports BlueCore Serial Protocol (BCSP) – a proprietary, reliable alternative to the standard Bluetooth UART Host Transport

.DataSheet4U. Im Provides a set of approximately 50 manufacturer-specific HCI extension commands. This command set (called BCCMD – "BlueCore Command"), provides:

- Access to the chip's general-purpose PIO port
- The negotiated effective encryption key length on established Bluetooth links
- Access to the firmware's random number generator
- Controls to set the default and maximum transmit powers these can help minimise interference between overlapping, fixed-location piconets
- Dynamic UART configuration

 Radio transmitter enable/disable – a simple command connects to a dedicated hardware switch that determines whether the radio can transmit

• The firmware can read the voltage on a pair of the chip's external pins. This is normally used to build a battery monitor, using either VM or host code

• A block of BCCMD commands provides access to the chip's "persistent store" configuration database (PS). The database sets the device's Bluetooth address, Class of Device, radio (transmit class) configuration, SCO routing, LM, USB and DFU constants, etc.

- A UART "break" condition can be used in three ways:
  - 1. Presenting a UART break condition to the chip can force the chip to perform a hardware reboot
  - 2. Presenting a break condition at boot time can hold the chip in a low power state, preventing normal initialisation while the condition exists
  - 3. With BCSP, the firmware can be configured to send a break to the host before sending data normally used to wake the host from a deep sleep state

• The DFU standard has been extended with public/private key authentication, allowing manufacturers to control the firmware that can be loaded onto their Bluetooth modules

A modified version of the DFU protocol allows firmware upgrade via the chip's UART

• A block of "radio test" or BIST commands allows direct control of the chip's radio. This aids the development of modules' radio designs, and can be used to support Bluetooth qualification.

• Virtual Machine (VM). The firmware provides the VM environment in which to run application-specific code. Although the VM is mainly used with BlueLab and "RFCOMM builds" (alternative firmware builds providing L2CAP, SDP and RFCOMM), the VM can be used with this build to perform simple tasks such as flashing LED's via the chip's PIO port.

 Hardware low power modes: shallow sleep and deep sleep. The chip drops into modes that significantly reduce power consumption when the software goes idle.

 SCO channels are normally routed via HCI (over BCSP). However, up to three SCO channels can be routed over the chip's single PCM port (at the same time as routing any remaining SCO channels over HCI).

• Co-operative existence with 802.11b/g chipsets. The device can be optionally configured to support a number of different co-existence schemes including:

- TDMA Bluetooth and WLAN avoid transmitting at the same time.
- FDMA Bluetooth avoids transmitting within the WLAN channel



- Combination TDMA & FDMA Bluetooth avoids transmitting in the WLAN channel only when WLAN is active.
- Please refer to separate documentation for full details of the co-existence schemes that CSR supports.

#### Note:

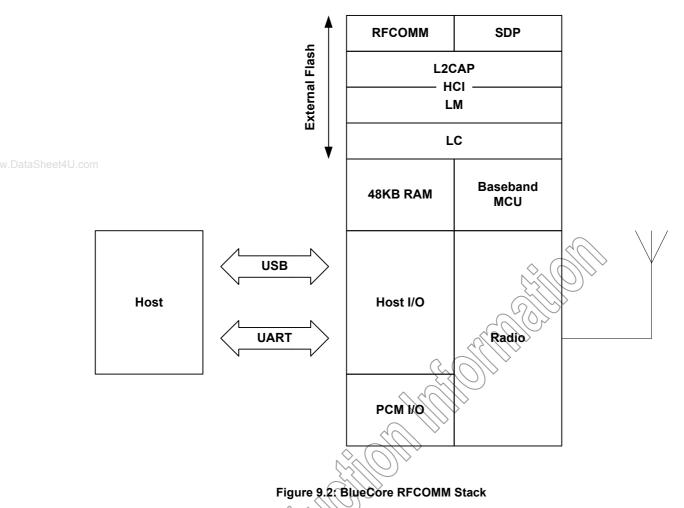
Always refer to the Firmware Release Note for the specific functionality of a particular build.

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## 9.2 BlueCore RFCOMM Stack



In the version of the firmware, shown in Figure 9.2 the upper layers of the Bluetooth stack up to RFCOMM are run on-chip. This reduces host-side software and hardware requirements at the expense of some of the power and flexibility of the HCI only stack.



### 9.2.1 Key Features of the BlueCore4-External RFCOMM Stack

#### Interfaces to Host:

- RFCOMM, an RS-232 serial cable emulation protocol
- SDP, a service database look-up protocol

#### Connectivity:

- Maximum number of active slaves: 3
- Maximum number of simultaneous active ACL connections: 3
- www.DataSheet4U.com Maximum number of simultaneous active SCO connections: 3
  - Data Rate: up to 350kbps<sup>(1)</sup>

#### Security:

Full support for all Bluetooth security features up to and including strong (128-bit) encryption.

#### **Power Saving:**

Full support for all Bluetooth power saving modes (Park, Sniff and Hold).

#### Data Integrity:

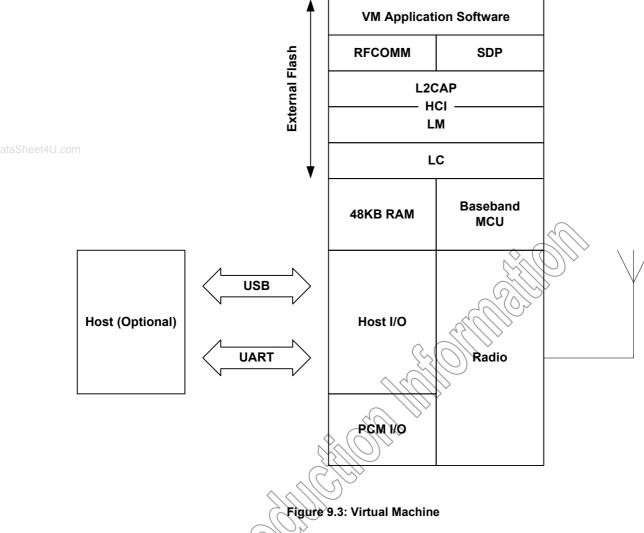
- CQDDR increases the effective data rate in noisy environments.
- RSSI used to minimise interference to other radio devices using the ISM band.

#### Note:

<sup>(1)</sup> The data rate is with respect to BlueCore4-External with basic data rate packets.



### 9.3 BlueCore Virtual Machine Stack



In Figure 9.3, this version of the stack firmware shown requires no host processor (but can use a host processor for debugging etc.). All software layers, including application software, run on the internal RISC processor in a protected user software execution environment known as a Virtual Machine (VM).

The user may write custom application code to run on the BlueCore VM using BlueLab™ software development kit (SDK) supplied with the BlueLab Multimedia and Casira development kits, available separately from CSR. This code will then execute alongside the main BlueCore firmware. The user is able to make calls to the BlueCore firmware for various operations.

The execution environment is structured so the user application does not adversely affect the main software routines, thus ensuring that the Bluetooth stack software component does not need re-qualification when the application is changed.

Using the VM and the BlueLab SDK the user is able to develop applications such as a cordless headset or other profiles without the requirement of a host controller. BlueLab is supplied with example code including a full implementation of the headset profile.

#### Note:

Sample applications to control PIO lines can also be written with BlueLab SDK and the VM for the HCI stack.



### 9.4 BlueCore HID Stack

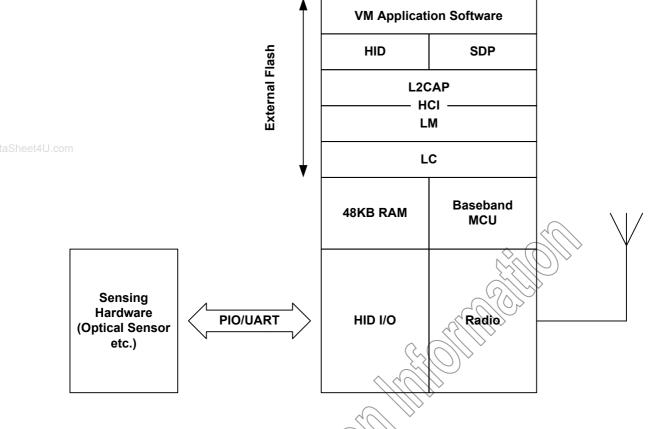


Figure 9.4: HID Stack

This version of the stack firmware requires no host processor. All software layers, including application software, run on the internal RISC microcontroller in a protected user software execution environment known as a virtual machine (VM).

The user may write custom application code to run on the BlueCore VM using BlueLab Professional software development kit (SDK) supplied with the BlueLab Professional and Casira development kits, available separately from CSR. This code will then execute alongside the main BlueCore firmware. The user is able to make calls to the BlueCore firmware for various operations.

The execution environment is structured so the user application does not adversely affect the main software routines, thus ensuring that the Bluetooth stack software component does not need re-qualification when the application is changed.

Using the VM and the BlueLab Professional SDK the user is able to develop Bluetooth HID devices such as an optical mouse or keyboard. The user is able to customise features such as power management and connect/reconnect behaviour.

The HID I/O component in the HID stack controls low latency data acquisition from external sensor hardware. With this component running in native code, it does not incur the overhead of the VM code interpreter. Supported external sensors include 5 mouse buttons, the Agilent ADNS-2030 optical sensor, quadrature scroll wheel, direct coupling to a keyboard matrix and a UART interface to custom hardware.

A reference schematic for implementing a three button, optical mouse with scroll wheel is available from CSR.



#### 9.5 BCHS Software

BlueCore Embedded Host Software is designed to enable CSR customers to implement Bluetooth functionality into embedded products quickly, cheaply and with low risk.

BCHS is developed to work with CSR's family of BlueCore IC's. BCHS is intended for embedded products that have a host processor for running BCHS and the Bluetooth application e.g. a mobile phone or a PDA. BCHS together with the BlueCore IC with embedded Bluetooth core stack (L2CAP, RFCOMM and SDP) is a complete Bluetooth system solution from RF to profiles.

BCHS includes most of the Bluetooth intelligence and gives the user a simple API. This makes it possible to develop a Bluetooth product without in-depth Bluetooth knowledge.

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The BlueCore Embedded Host Software contains 3 elements:

- Example Drivers (BCSP and proxies)
- Bluetooth Profile Managers
- Example Applications

The profiles are qualified which makes the qualification of the final product very easy. BCHS is delivered with source code (ANSI C). With BCHS also come example applications in ANSI C, which makes the process of writing the application easier.

#### 9.6 Additional Software for Other Embedded Applications

When the upper layers of the Bluetooth protocol stack are run as firmware on BlueCore4-External, a UART software driver is supplied that presents the L2CAP, RFCOMM and Service Discovery (SDP) APIs to higher Bluetooth stack layers running on the host. The code is provided as C source or object code.

#### 9.7 CSR Development Systems

CSR's BlueLab and Casira development kits are available to allow the evaluation of the BlueCore4-External hardware and software, and as toolkits for developing on-chip and host software.



# 10 Enhanced Data Rate

EDR has been introduced to provide 2x and optionally 3x data rates with minimal disruption to higher layers of the Bluetooth stack. CSR supports both of the new data rates, with BlueCore4-External. BlueCore4-External is compliant with revision v0.9 of the specification.

# 10.1 Enhanced Data Rate Baseband

At the baseband level EDR uses the same 1.6kHz slot rate as basic data rate and therefore the packets can be 1, 3, or 5 slots long as per the basic data rate. Where EDR differs from the basic data rate is that in the same 1MHz symbol rate 2 or 3bits are used per symbol, compared to 1bit per symbol used by the basic data rate. To achieve the increase in number of bits symbol, two new modulation schemes have been introduced as summarised in Table 10.1 and the modulation schemes are explained further in Section 10.2 and Section 0.

Data Rate Scheme	Bits Per Symbol	Modulation
Basic data rate	1	GFSK
EDR	2	π/4 DQPSK
EDR	3	8DPSK (optional)

Table 10.1: Data Rate Schemes

Although the EDR uses new packets Link establishment and management are unchanged and still use Basic Rate packets

#### **10.2** Enhanced Data Rate $\pi/4$ DQPSK

8-state Differential Phase Shift Keying

2 bits determine phase shift between consecutive symbols

2 bits determine phase shift between consecutive symbols

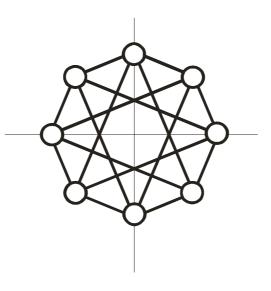
 $\pi/4$  rotation avoids phase shift of  $\pi$ , which would cause large amplitude variation

Raised Cosine pulse shaping filter to further reduce side band emissions

Bit Pattern		Phase Shift
00	$\sim$	π/4
01		3π/4
11	All a	-3π/4
10	$\bigvee$	-π/4

Table 10.2: 2 bits determine phase shift between consecutive symbols





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Table 10.3: π/4 DQPSK

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#### 10.3 Enhanced Data Rate 8DPSK

8-state Differential Phase-Shift Keying

Three bits determine phase shift between consecutive symbols

Bit Pattern		Phase Shift
000		Ø
001		TU4
011		π/2
010	C	3π/4
110	0000	π
111	C//Pr	-3π/4
101	a Cler	-π/2
100		-π/4

Table 10.4: 3 bits determine phase shift between consecutive symbols



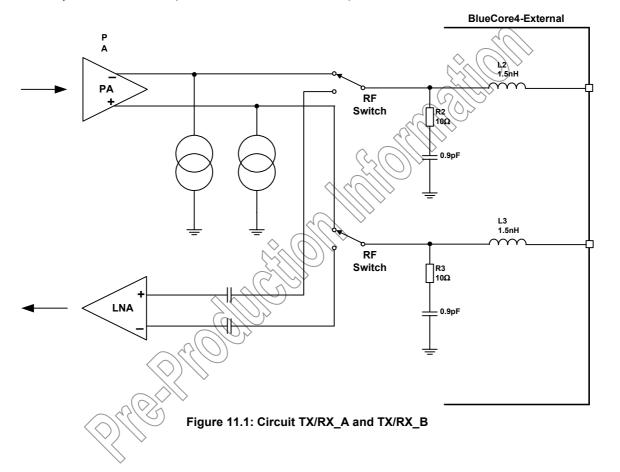
# **11 Device Terminal Descriptions**

# 11.1 RF Ports

The BlueCore4-External RF\_IN terminal can be configured as either a single ended or differential input. The operational mode is determined by the setting the PS Key PSKEY\_TXRX\_PIO\_CONTROL (0x20).

# 11.1.1 TX\_A and TX\_B

TX\_A and TX\_B form a complementary balanced pair. On transmit; their outputs are combined using a balun into the single-ended output required for the antenna. Similarly, on receive their input signals are combined internally. Both terminals present similar complex impedances that require matching networks between them and the balun. Starting from the substrate (chip side), the outputs can each be modelled as an ideal current source in parallel with a lossy resistance and a capacitor. The bond wire can be represented as series inductance.

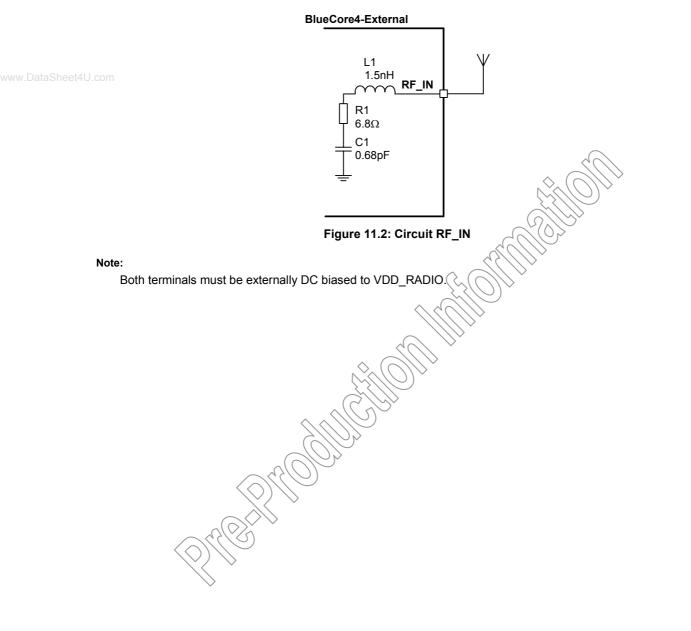




### 11.1.2 Single-Ended Input (RF\_IN)

This is the single ended RF input from the antenna. The input presents a complex impedance that requires a matching network between the terminal and the antenna. Starting from the substrate (chip) side, the input can be modelled as a lossy capacitor with the bond wire to the ball grid represented as a series inductance.

The terminal is DC blocked. The DC level must not exceed (VSS\_RADIO -0.3V to VDD\_RADIO + 0.3V).





### 11.2 External Reference Clock Input (XTAL\_IN)

The BlueCore4-External RF local oscillator and internal digital clocks are derived from the reference clock at the BlueCore4-External XTAL\_IN input. This reference may be either an external clock or from a crystal connected between XTAL\_IN and XTAL\_OUT. The crystal mode is described in Section 11.2.5.

#### 11.2.1 External Mode

BlueCore4-External can be configured to accept an external reference clock (from another device, such as TCXO) at XTAL\_IN by connecting XTAL\_OUT to ground. The external clock can either be a digital level square wave or sinusoidal and this may be directly coupled to XTAL\_IN without the need for additional components. If the peaks of the reference clock are below VSS\_ANA or above VDD\_ANA, it must be driven through a DC blocking capacitor (~33pF) connected to XTAL\_IN. A digital level reference clock gives superior noise immunity as the high slew rate clock edges have lower voltage to phase conversion.

The external clock signal should meet the specifications in Table 11.1:

	Min	Тур	Max
Frequency <sup>(1)</sup>	7.5MHz	16MHz	40MHz
Duty cycle	20:80	50:50	80:20
Edge Jitter (At Zero Crossing)	-	-6//5	15ps rms
Signal Level	400mV pk-pk		VDD_ANA (2)(3)

#### Table 11.1: External Clock Specifications

#### Notes:

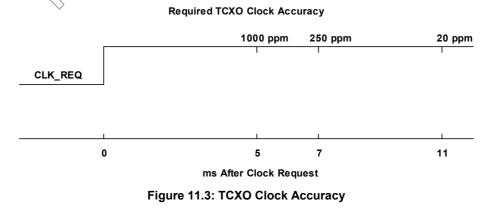
- <sup>(1)</sup> The frequency should be an integer multiple of 250kHz except for the CDMA/3G frequencies
- <sup>(2)</sup> VDD\_ANA is 1.8V nominal
- <sup>(3)</sup> If the external clock is driven through a DC blocking capacitor then maximum allowable amplitude is reduced from VDD\_ANA to 800mV pk-pk

### 11.2.2 XTAL\_IN Impedance in External Mode

The impedance of the XTAL\_IN will not change significantly between operating modes, typically 10fF. When transitioning from deep sleep to an active state a spike of up to 1pC may be measured. For this reason it is recommended that a buffered clock input be used.

# 11.2.3 Clock Timing Accuracy

As Figure 11.3 indicates, the 250ppm timing accuracy on the external clock is required 7ms after the assertion of the system clock request line. This is to guarantee that the firmware can maintain timing accuracy in accordance with the Bluetooth v1.2 specification. Radio activity may occur after 11ms, therefore at this point, the timing accuracy of the external clock source must be within 20ppm.

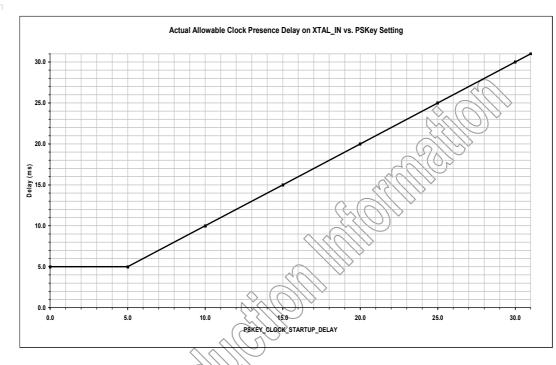




# 11.2.4 Clock Start-Up Delay

BlueCore4-External hardware incorporates an automatic 5ms delay after the assertion of the system clock request signal before running firmware. This is suitable for most applications using an external clock source. However, there may be scenarios where the clock cannot be guaranteed to either exist or be stable after this period. Under these conditions, BlueCore4-External firmware provides a software function which will extend the system clock request signal by a period stored in PSKEY\_CLOCK\_STARTUP\_DELAY. This value is set in milliseconds from 5-31ms.

This PS Key allows the designer to optimise a system where clock latencies may be longer than 5ms while still keeping the current consumption of BlueCore4-External as low as possible. BlueCore4-External will consume about 2mA of current for the duration of PSKEY\_CLOCK\_STARTUP\_DELAY before activating the firmware.









# 11.2.5 Input Frequencies and PS Key Settings

BlueCore4-External should be configured to operate with the chosen reference frequency. This is accomplished by setting the PS Key PSKEY\_ANA\_FREQ (0x1fe) for all frequencies with an integer multiple of 250kHz. The input frequency default setting in BlueCore4-External is 26MHz.

The following CDMA/3G TCXO frequencies are also catered for: 7.68, 14.4, 15.36, 16.2, 16.8, 19.2, 19.44, 19.68, 19.8 and 38.4MHz.

	Reference Crystal Frequency (MHz)	PSKEY_ANA_FREQ (0x1fe) (Units of 1kHz)
ww.DataSheet4U	7.68	7680
ww.DataSheet+0 <del></del>	14.40	14400
	15.36	15360
	16.20	16200
	16.80	16800
	19.20	19200
	19.44	19440
	19.68	19680
	19.80	19800
	38.40	38400
	n x 250kHz	
	+26.00 Default	26000

Table 11.2: PS Key Values for CDMA/3G Phone TCXO Frequencies



### 11.3 Crystal Oscillator (XTAL\_IN, XTAL\_OUT)

The BlueCore4-External RF local oscillator and internal digital clocks are derived from the reference clock at the BlueCore4-External XTAL\_IN input. This reference may be either an external clock or from a crystal connected between XTAL\_IN and XTAL\_OUT. The external reference clock mode is described in Section 11.2.

#### 11.3.1 XTAL Mode

BlueCore4-External contains a crystal driver circuit. This operates with an external crystal and capacitors to form a Pierce oscillator.

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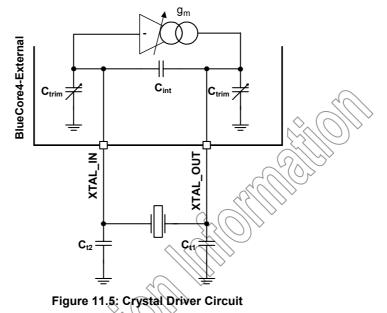


Figure 11.6 shows an electrical equivalent circuit for a crystal. The crystal appears inductive near its resonant

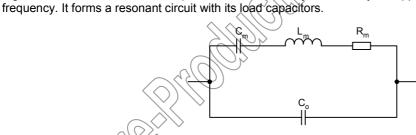


Figure 11.6: Crystal Equivalent Circuit

The resonant frequency may be trimmed with the crystal load capacitance. BlueCore4-External contains variable internal capacitors to provide a fine trim.

The BlueCore4-External driver circuit is a transconductance amplifier. A voltage at XTAL\_IN generates a current at XTAL\_OUT. The value of transconductance is variable and may be set for optimum performance.



#### 11.3.2 Load Capacitance

For resonance at the correct frequency the crystal should be loaded with its specified load capacitance, which is defined for the crystal. This is the total capacitance across the crystal viewed from its terminals. BlueCore4-External provides some of this load with the capacitors  $C_{trim}$  and  $C_{int}$ . The remainder should be from the external capacitors labelled  $C_{t1}$  and  $C_{t2}$ .  $C_{t1}$  should be three times the value of  $C_{t2}$  for best noise performance. This maximises the signal swing, hence slew rate at XTAL\_IN, to which all on chip clocks are referred. Crystal load capacitance,  $C_i$  is calculated with the following equation:

$$C_{I} = C_{int} + \frac{C_{trim}}{2} + \frac{C_{t1} \cdot C_{t2}}{C_{t1} + C_{t2}}$$

#### **Equation 11.1: Load Capacitance**

#### Where:

Ctrim = 3.4pF nominal (Mid range setting)

C<sub>int</sub> = 1.5pF

Note:

Cint does not include the crystal internal self capacitance, it is the driver self capacitance

#### 11.3.3 Frequency Trim

BlueCore4-External enables frequency adjustments to be made. This feature is typically used to remove initial tolerance frequency errors associated with the crystal. Frequency trim is achieved by adjusting the crystal load capacitance with on chip trim capacitors,  $C_{trim}$ . The value of  $C_{trim}$  is set by a 6-bit word in the Persistent Store Key PSKEY\_ANA\_FTRIM (0x1f6). Its value is calculated thus:

#### Equation 11.2: Trim Capacitance

There are two C<sub>trim</sub> capacitors, which are both connected to ground. When viewed from the crystal terminals, they appear in series so each least significant bit (LSB) increment of frequency trim presents a load across the crystal of 55fF.

The frequency trim is described by Equation 11.3.

 $\Delta(F_x) = \text{pullability} \times 55 \times 10^{-3} (\text{ppm/LSB})$ 

Equation 11.3: Frequency Trim

Where  $F_x$  is the crystal frequency and pullability is a crystal parameter with units of ppm/pF. Total trim range is 63 times the value above.

If not specified, the pullability of a crystal may be calculated from its motional capacitance with Equation 11.4:

$$\frac{\partial(\mathsf{F}_{x})}{\partial(\mathsf{C})} = \mathsf{F}_{x} \cdot \frac{\mathsf{C}_{\mathsf{m}}}{4(\mathsf{C}_{\mathsf{I}} + \mathsf{C}_{\mathsf{n}})^{2}}$$

#### Equation 11.4: Pullability

#### Where:

C<sub>0</sub> = Crystal self capacitance (shunt capacitance)

C<sub>m</sub> = Crystal motional capacitance (series branch capacitance in crystal model). See Figure 11.6.

Note:

It is a Bluetooth requirement that the frequency is always within  $\pm 20$  ppm. The trim range should be sufficient to pull the crystal within  $\pm 5$  ppm of the exact frequency. This leaves a margin of  $\pm 15$  ppm for frequency drift with ageing and temperature. A crystal with an ageing and temperature drift specification of better than  $\pm 15$  ppm is required.



# **11.3.4 Transconductance Driver Model**

The crystal and its load capacitors should be viewed as a transimpedance element, whereby a current applied to one terminal generates a voltage at the other. The transconductance amplifier in BlueCore4-External uses the voltage at its input, XTAL\_IN, to generate a current at its output, XTAL\_OUT. Therefore, the circuit will oscillate if the transconductance, transimpedance product is greater than unity. For sufficient oscillation amplitude, the product should be greater than 3. The transconductance required for oscillation is defined by the following relationship:

$$g_{m} > \frac{3(C_{t1} + C_{trim})(C_{t2} + C_{trim})}{(2\pi F_{x})^{2}R_{m}((C_{0} + C_{int})(C_{t1} + C_{t2} + 2C_{trim}) + (C_{t1} + C_{trim})(C_{t2} + C_{trim}))^{2}}$$

Equation 11.5: Transconductance Required for Oscillation

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BlueCore4-External guarantees a transconductance value of at least 2mA/V at maximum drive level.

#### Notes:

More drive strength is required for higher frequency crystals, higher loss crystals (larger  $R_m$ ) or higher capacitance loading.

Optimum drive level is attained when the level at XTAL\_IN is approximately 1V pk-pk. The drive level is determined by the crystal driver transconductance, by setting the Persistent Store KEX XTAL LVL (0x241).

#### 11.3.5 Negative Resistance Model

An alternative representation of the crystal and its load capacitors is a frequency dependent resistive element. The driver amplifier may be considered as a circuit that provides negative resistance. For oscillation, the value of the negative resistance must be greater than that of the crystal circuit equivalent resistance. Although the BlueCore4-External crystal driver circuit is based on a transimpedance amplifier, an equivalent negative resistance may be calculated for it with the following formula in Equation 11.6:

$$R_{neg} > \frac{3(C_{t1} + C_{trim})(C_{t2} + C_{trim})}{g_m (2\pi F_x)^2 (C_0 + C_{int})(C_{t1} + C_{t2} + 2C_{trim}) + (C_{t1} + C_{trim})(C_{t2} + C_{trim})^2}$$

#### Equation 11.6: Equivalent Negative Resistance

This formula shows the negative resistance of the BueCore4-External driver as a function of its drive strength.

The value of the driver negative resistance may be easily measured by placing an additional resistance in series with the crystal. The maximum value of this resistor (oscillation occurs) is the equivalent negative resistance of the oscillator.

	Min	Тур	Мах
Frequency	8MHz	16MHz	32MHz
Initial Tolerance	- (2) 7)	±25ppm	-
Pullability	-	±20ppm/pF	-

Table 11.3: Oscillator Negative Resistance

### 11.3.6 Crystal PS Key Settings

See tables in Section 11.2.5.



# 11.3.7 Crystal Oscillator Characteristics

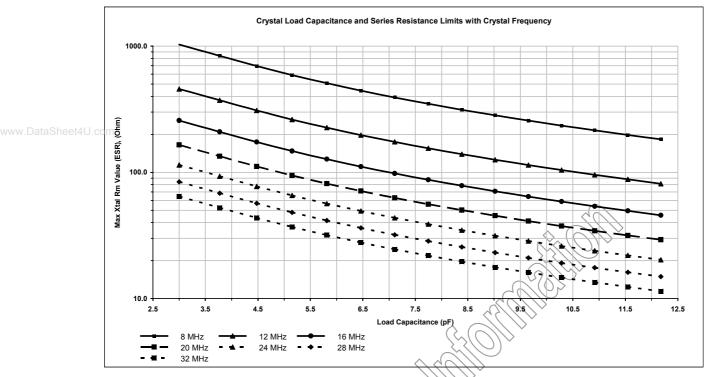


Figure 11.7: Crystal Load Capacitance and Series Resistance Limits with Crystal Frequency

#### Note:

Graph shows results for BlueCore4-External crystal driver at maximum drive level.

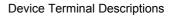
#### Conditions:

 $C_{trim}$  = 3.4pF centre value Crystal  $C_o$  = 2pF

Transconductance setting = 2mA/N

Loop gain = 3

 $C_{t1}/C_{t2} = 3$ 





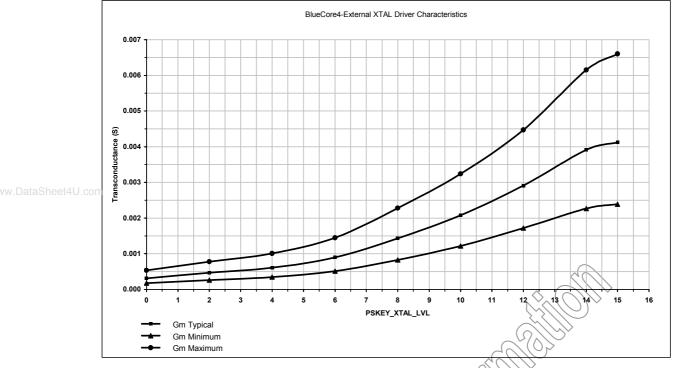


Figure 11.8: Crystal Driver Transconductance vs. Driver Level Register Setting

#### Note:

Drive level is set by Persistent Store Key PSKEY\_XTAL\_LVL (0x241).

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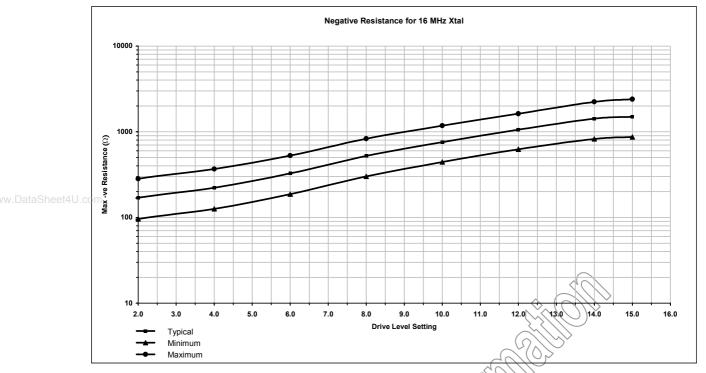


Figure 11.9: Crystal Driver Negative Resistance as a Function of Drive Level Setting

#### Crystal parameters:

Crystal frequency 16MHz (Please refer to your software build release note for frequencies supported);

Crystal C<sub>0</sub> = 0.75pF

#### Circuit parameters:

C<sub>trim</sub> = 8pF, maximum value

 $C_{t1}$ , $C_{t2}$  = 5pF (3.9pF plus 1.1 pF stray)

(Crystal total load capacitance 8.5pF)

#### Note:

This is for a specific crystal and load capacitance.



## 11.4 Off-Chip Program Memory

The external memory port provides a facility to interface up to 8Mbits of 16-bit external memory. This off-chip storage is used to store BlueCore4-External settings and program code. Flash is the storage mechanism typically used by BlueCore4-External modules, however external masked-ROM may also be used if the host takes over responsibility for storing configuration data.

The external memory port consists of 16 bi-directional data lines, D[15:0]; 19 output address lines, A[18:0] and three active low output control signals (WEB, CEB, REB). WEB is asserted when data is written to external memory. REB is asserted when data is read from external memory and the chip select line. CSB is asserted when any data transfer (read or write) is required. All of the external memory port connections are implemented using CMOS technology and use standard 0V and VDD\_MEM (1.8-3.6V) signalling levels.

Parameter	Value	
Data width	16-bit	
Minimum total capacity	4Mbit (256kWord)	
Maximum access time	90ns @125°C 50pF load	
	110ns @85°C 10pF load	

#### Table 11.4: Flash Device Hardware Requirements

In addition to these hardware requirements, particular care should be taken to ensure that the sector organisation of the extended memory has the correct format. A sector is defined as an individually erasable area of external Flash.

It is important to make sure that external memory devices meet certain minimum specifications. In addition particular care should be taken to ensure that the sector organisation of the extended memory has the correct format.



### 11.4.1 Minimum Flash Specification

The flash device used with BlueCore4-External must meet the following criteria:

- Standard or extended form of either the JEDEC (AMD/Fujitsu/SST) or Intel command set.
- Access time must be ≤90ns @125°C 50pF load or ≤110ns @85°C 10pF load.
- Write strobe of 100ns.
- Accessible in word mode, i.e., via a 16-bit data bus.
- Support changing different bits within each word from 1 to 0 in at least two separate programming operations.
- Programming and erase times must have fixed upper limits.
- Must be bottom boot or uniform sector.

Must have independently erasable sectors with at least the following boundaries (see Memory Map for more information).

Word Address	Size (kWords)
0x00000 - 0x01FFF	8
0x02000 - 0x02FFF	(A)
0x03000 - 0x03FFF	
0x04000 - 0x07FFF	16
0x08000 - 0x0FFFF	32
0x10000 - 0x17FFF	32
0x18000	Don't care

#### Table 11.5: Flash Sector Boundaries

#### Important Note:

Satisfaction of these criteria is not sufficient for a particular device to be used; it must also support the Common Flash Interface described in Section 11.4.2 or be supported in the BlueCore4-External firmware and host-side tools.

### 11.4.2 Common Flash Interface

The firmware can adapt automatically to work with some flash devices. If in addition to satisfying the minimum Flash specification described above, they meet the following criteria:

The device must support the Common Flash Interface, as defined by JEDEC standard JESD68.

The device must return one of the following codes for either the Primary or Alternative Algorithm Command Set (offset 0x13b or 0x17 of the Query Structure Output):

Code	Description
0x0001	Intel/Sharp Extended Command Set
0x0002	AMD/Fujitsu Standard Command Set
0x0003	Intel Standard Command Set
0x0701	AMD/Fujitsu Extended Command Set

#### Table 11.6: Common Flash Interface Return Codes

The device must return one of the following patterns of Erase Block Region Information (beginning at offset 0x2d of the Query Structure Output):

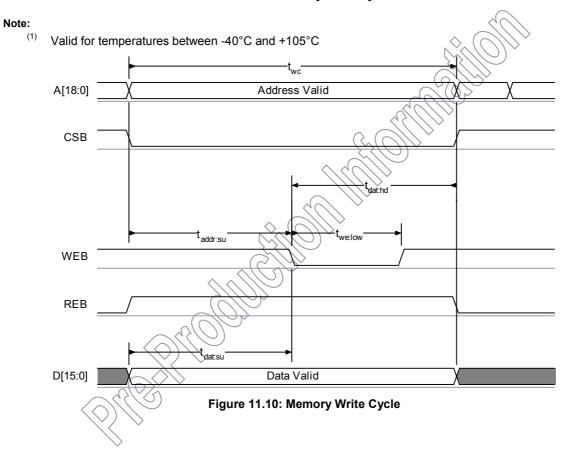
If any of these criteria is not met, then the device will **not** work unless the device is supported by the BlueCore4-External firmware.

## 11.4.3 Memory Timing

#### Memory Write Cycle

Symbol	Parameter	Min <sup>(1)</sup>	Тур	Max <sup>(1)</sup>	Unit
t <sub>wc</sub>	Write cycle time	300	-	-	ns
t <sub>dat:su</sub>	Data set-up time	150	-	-	ns
t <sub>dat:hd</sub>	Data hold time	150	-	-	ns
taddr:su	Address set-up time	150	-	-	ns
t <sub>we:low</sub>	WEB low	100	-	-	ns

#### Table 9.7: Memory Write Cycle



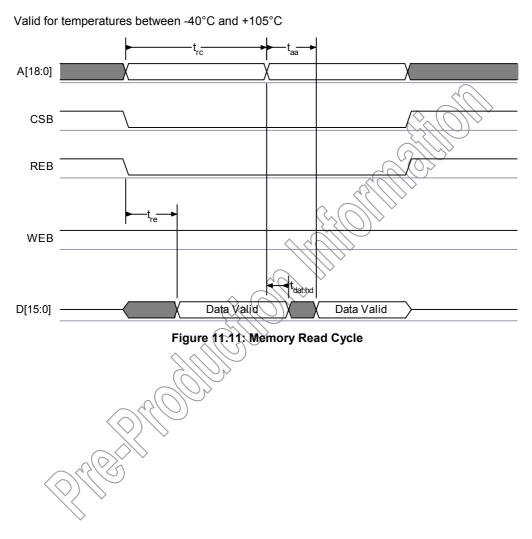


#### Memory Read Cycle

Symbol	Parameter	Min <sup>(1)</sup>	Тур	Max <sup>(1)</sup>	Unit
t <sub>rc</sub>	Read cycle time	114	125	-	ns
t <sub>aa</sub>	Address access time	-	-	110	ns
t <sub>re</sub>	Read enable access time	-	-	110	ns
t <sub>dat:hd</sub>	Data hold time from address line		-	-	ns

Table 9.8: Memory Read Cycle

#### Note: www.DataSheet4U.com (1)





## 11.5 UART Interface

BlueCore4-External Universal Asynchronous Receiver Transmitter (UART) interface provides a simple mechanism for communicating with other serial devices using the RS232 standard<sup>(1)</sup>.

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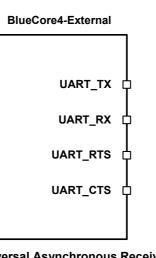


Figure 11.12: Universal Asynchronous Receiver

Four signals are used to implement the UART function, as shown in Figure 11.12. When BlueCore4-External is connected to another digital device, UART\_RX and UART\_TX transfer data between the two devices. The remaining two signals, UART\_CTS and UART\_RTS, can be used to implement RS232 hardware flow control where both are active low indicators. All UART connections are implemented using CMOS technology and have signalling levels of 0V and VDD\_PADS.

UART configuration parameters, such as Baud rate and packet format, are set using BlueCore4-External software.

#### Notes:

In order to communicate with the UART at its maximum data rate using a standard PC, an accelerated serial port adapter card is required for the PC.

(1)				
(1)	Uses RS232 protocol but voltage	levels are 0V to VDD	USB.	(requires external RS232 transceiver chip)
			,	

Parameter		Possible Values		
Baud Rate	Minimum	1200 Baud (≤2%Error)		
		<sup>)</sup> 9600 Baud (≤1%Error)		
	Maximum	1.5MBaud (≤1%Error)		
Flow Control		RTS/CTS or None		
Parity		None, Odd or Even		
Number of Stop Bits		1 or 2		
Bits per channel		8		

#### Table 11.7: Possible UART Settings



The UART interface is capable of resetting BlueCore4-External upon reception of a break signal. A Break is identified by a continuous logic low (0V) on the UART\_RX terminal, as shown in Figure 11.13. If  $t_{BRK}$  is longer than the value, defined by the PS Key PSKEY\_HOST\_IO\_UART\_RESET\_TIMEOUT, (0x1a4), a reset will occur. This feature allows a host to initialise the system to a known state. Also, BlueCore4-External can emit a Break character that may be used to wake the Host.

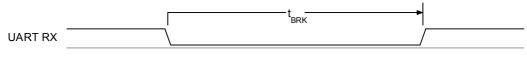
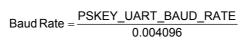


Figure 11.13: Break Signal

#### Note:

<sup>om</sup> The DFU boot loader must be loaded into the Flash device before the UART or USB interfaces can be used. This initial flash programming can be done via the SPI.

Table 11.3 shows a list of commonly used Baud rates and their associated values for the Persistent Store Key PSKEY\_UART\_BAUD\_RATE (0x204). There is no requirement to use these standard values. Any Baud rate within the supported range can be set in the Persistent Store Key according to the formula in Equation 11.7.



#### Equation 11.7: Baud Rate

Baud Rate	Persistent S	Error				
Daud Nate	Hex	Dec	End			
1200	0x0005	5	1.73%			
2400	0x000a	10	1.73%			
4800	0x0014	) → 20	1.73%			
9600	0x0027	39	-0.82%			
19200	0x004f	79	0.45%			
38400	0x009d	157	-0.18%			
57600	0x00ec	236	0.03%			
76800	0x013b	315	0.14%			
115200	💛 🚫 0x01d8	472	0.03%			
230400	0x03b0	944	0.03%			
460800	) 0x075f	1887	-0.02%			
921600	0x0ebf	3775	0.00%			
1382400	0x161e	5662	-0.01%			

Table 11.8: Standard Baud Rates



# 11.5.1 UART Bypass

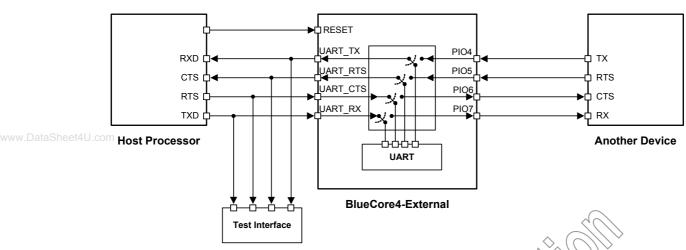


Figure 11.14: UART Bypass Architecture

# 11.5.2 UART Configuration While RESET is Active

The UART interface for BlueCore4-External while the chip is being held in reset is tri-state. This will allow the user to daisy chain devices onto the physical UART bus. The constraint on this method is that any devices connected to this bus must tri-state when BlueCore4-External reset is de-asserted and the firmware begins to run.

# 11.5.3 UART Bypass Mode

Alternatively, for devices that do not tri-state the UART bus, the UART bypass mode on BlueCore4-External can be used. The default state of BlueCore4-External after reset is de-asserted, this is for the host UART bus to be connected to the BlueCore4-External UART, thereby allowing communication to BlueCore4-External via the UART.

In order to apply the UART bypass mode, a BCCMD command will be issued to BlueCore4-External upon this, it will switch the bypass to PIO[7:4] as shown in Figure 11.14. Once the bypass mode has been invoked, BlueCore4-External will enter the deep sleep state indefinitely.

In order to re-establish communication with BlueCore4-External, the chip must be reset so that the default configuration takes affect.

It is important for the host to ensure a clean Bluetooth disconnection of any active links before the bypass mode is invoked. Therefore it is not possible to have active Bluetooth links while operating the bypass mode.

# 11.5.4 Current Consumption in UART Bypass Mode

The current consumption for a device in UART Bypass Mode is equal to the values quoted for a device in standby mode.



### 11.6 USB Interface

BlueCore4-External USB devices contain a full speed (12Mbits/s) USB interface that is capable of driving a USB cable directly. No external USB transceiver is required. The device operates as a USB peripheral, responding to requests from a master host controller such as a PC. Both the OHCI and the UHCI standards are supported. The set of USB endpoints implemented can behave as specified in the USB section of the Bluetooth specification v1.2 or alternatively can appear as a set of endpoint appropriate to USB audio devices such as speakers.

As USB is a Master/Slave oriented system (in common with other USB peripherals), BlueCore4-External only supports USB Slave operation.

# 11.6.1 USB Data Connections

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The USB data lines emerge as pins USB\_DP and USB\_DN. These terminals are connected to the internal USB I/O buffers of the BlueCore4-External and therefore have a low output impedance. To match the connection to the characteristic impedance of the USB cable, resistors must be placed in series with USB\_DP / USB\_DN and the cable.

## 11.6.2 USB Pull-Up Resistor

BlueCore4-External features an internal USB pull-up resistor. This pulls the USB\_DP pin weakly high when BlueCore4-External is ready to enumerate. It signals to the PC that it is a full speed (12Mbt/s) USB device.

The USB internal pull-up is implemented as a current source, and is compliant with Section 7.1.5 of the USB specification v1.2. The internal pull-up pulls USB\_DP high to at least 2.8V when loaded with a  $15k\Omega \pm 5\%$  pull-down resistor (in the hub/host) when VDD\_PADS=3.1V. This presents a Thevenin resistance to the host of at least 900 $\Omega$ . Alternatively, an external 1.5k $\Omega$  pull-up resistor can be placed between a PIO line and D+ on the USB cable. The firmware must be alerted to which mode is used by setting PS Key PSKEY\_USB\_PIO\_PULLUP appropriately. The default setting uses the internal pull-up resistor.

### 11.6.3 Power Supply

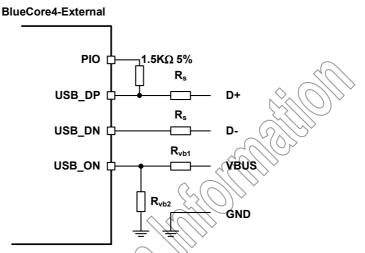
The USB specification dictates that the minimum output high voltage for USB data lines is 2.8V. To safely meet the USB specification, the voltage on the VDD\_USB supply terminals must be an absolute minimum of 3.1V. CSR recommends 3.3V for optimal USB signal quality.



### 11.6.4 Self Powered Mode

In self powered mode, the circuit is powered from its own power supply and not from the VBUS (5V) line of the USB cable. It draws only a small leakage current (below 0.5mA) from VBUS on the USB cable. This is the easier mode for which to design for, as the design is not limited by the power that can be drawn from the USB hub or root port. However, it requires that VBUS be connected to BlueCore4-External via a resistor network ( $R_{vb1}$  and  $R_{vb2}$ ), so BlueCore4-External can detect when VBUS is powered up. BlueCore4-External will not pull USB\_DP high when VBUS is off.

Self powered USB designs (powered from a battery or PSU) must ensure that a PIO line is allocated for USB pull-up purposes. A 1.5K 5% pull-up resistor between USB\_DP and the selected PIO line should be fitted to the design. Failure to fit this resistor may result in the design failing to be USB compliant in self powered mode. The www.DataSheet4U.internal pull-up in BlueCore is only suitable for bus powered USB devices i.e. dongles.



#### Figure 11.15: USB Connections for Self Powered Mode

The terminal marked USB\_ON can be any free PIO pin. The PIO pin selected must be registered by setting PSKEY\_USB\_PIO\_VBUS to the corresponding pin number.



# 11.6.5 Bus Powered Mode

In bus powered mode the application circuit draws its current from the 5V VBUS supply on the USB cable. BlueCore4-External negotiates with the PC during the USB enumeration stage about how much current it is allowed to consume.

For Class 2 Bluetooth applications, CSR recommends that the regulator used to derive 3.3V from VBUS is rated at 100mA average current and should be able to handle peaks of 120mA without foldback or limiting. In bus powered mode, BlueCore4-External requests 100mA during enumeration.

For Class 1 Bluetooth applications, the USB power descriptor should be altered to reflect the amount of power required. This is accomplished by setting the PS Key PSKEY\_USB\_MAX\_POWER (0x2c6). This is higher than for a Class 2 application due to the extra current drawn by the Transmit RF PA.

When selecting a regulator, be aware that VBUS may go as low as 4.4V. The inrush current (when charging reservoir and supply decoupling capacitors) is limited by the USB specification (see USB specification v1.1, Section 7.2.4.1). Some applications may require soft start circuitry to limit inrush current if more than  $10\mu$ F is present between VBUS and GND.

The 5V VBUS line emerging from a PC is often electrically noisy. As well as regulation down to 3.3V and 1.8V, applications should include careful filtering of the 5V line to attenuate noise that is above the voltage regulator bandwidth. Excessive noise on the 1.8V supply to the analogue supply pins of BlueCore4-External will result in reduced receive sensitivity and a distorted RF transmit signal.

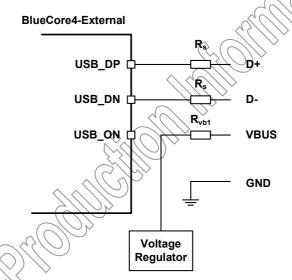


Figure 11.16: USB Connections for Bus Powered Mode

Note:

USB\_ON is shared with BlueCore4-External PIO terminals

Identifier	Value	Function
Rs	$27\Omega$ nominal	Impedance matching to USB cable
R <sub>vb1</sub>	22kΩ 5%	VBUS ON sense divider
R <sub>vb2</sub>	47kΩ 5%	VBUS ON sense divider

Table 11.9: USB Interface Component Values



## 11.6.6 Suspend Current

All USB devices must permit the USB controller to place them in a USB Suspend mode. While in USB Suspend, bus powered devices must not draw more than 0.5mA from USB VBUS (self powered devices may draw more than 0.5mA from their own supply). This current draw requirement prevents operation of the radio by bus powered devices during USB Suspend.

The voltage regulator circuit itself should draw only a small quiescent current (typically less than 100 $\mu$ A) to ensure adherence to the suspend current requirement of the USB specification. This is not normally a problem with modern regulators. Ensure that external LEDs and/or amplifiers can be turned off by BlueCore4-External. The entire circuit must be able to enter the suspend mode. (For more details on USB Suspend, see separate CSR documentation).

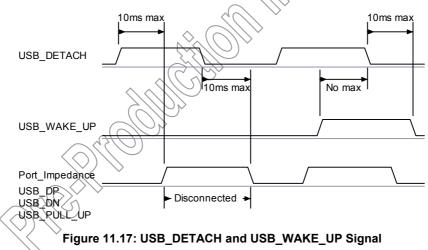
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#### 11.6.7 Detach and Wake-Up Signalling

BlueCore4-External can provide out-of-band signalling to a host controller by using the control lines called 'USB\_DETACH' and 'USB\_WAKE\_UP'. These are outside the USB specification (no wires exist for them inside the USB cable), but can be useful when embedding BlueCore4-External into a circuit where no external USB is visible to the user. Both control lines are shared with PIO pins and can be assigned to any PIO pin by setting the PS Keys PSKEY\_USB\_PIO\_DETACH and PSKEY\_USB\_PIO\_WAKEUP to the selected PIO number.

USB\_DETACH is an input which, when asserted high, causes BlueCore4-External to put USB\_DN and USB\_DP in a high impedance state and turned off the pull-up resistor on D+. This detaches the device from the bus and is logically equivalent to unplugging the device. When USB\_DETACH is taken low, BlueCore4-External will connect back to USB and await enumeration by the USB host.

USB\_WAKE\_UP is an active high output (used only when USB\_DETACH is active) to wake up the host and allow USB communication to recommence. It replaces the function of the software USB WAKE\_UP message (which runs over the USB cable), and cannot be sent while BlueCore4-External is effectively disconnected from the bus.



#### 11.6.8 USB Driver

A USB Bluetooth device driver is required to provide a software interface between BlueCore4-External and Bluetooth software running on the host computer. Suitable drivers are available from www.csrsupport.com.



## 11.6.9 USB 1.1 Compliance

BlueCore4-External is qualified to the USB specification v1.1, details of which are available from http://www.usb.org. The specification contains valuable information on aspects such as PCB track impedance, supply inrush current and product labelling.

Although BlueCore4-External meets the USB specification, CSR cannot guarantee that an application circuit designed around the chip is USB compliant. The choice of application circuit, component choice and PCB layout all affect USB signal quality and electrical characteristics. The information in this document is intended as a guide and should be read in association with the USB specification, with particular attention being given to Chapter 7. Independent USB qualification must be sought before an application is deemed USB compliant and can bear the USB logo. Such qualification can be obtained from a USB plugfest or from an independent USB test house.

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Terminals USB\_DP and USB\_DN adhere to the USB specification 2.0 (Chapter 7) electrical requirements.

#### 11.6.10 USB 2.0 Compatibility

BlueCore4-External is compatible with USB v2.0 host controllers; under these circumstances the two ends agree the mutually acceptable rate of 12Mbits/s according to the USB v2.0 specification.

#### 11.7 Serial Peripheral Interface

BlueCore4-External uses 16-bit data and 16-bit address serial peripheral interface, where transactions may occur when the internal processor is running or is stopped. This section details the considerations required when interfacing to BlueCore4-External via the four dedicated serial peripheral interface terminals. Data may be written or read one word at a time or the auto increment feature may be used to access blocks.

### **11.7.1 Instruction Cycle**

The BlueCore4-External is the slave and receives commands on SPI\_MOSI and outputs data on SPI\_MISO. The instruction cycle for a SPI transaction is shown in Table 11.10.

1	Reset the SPI interface	Hold SPI_CSB high for two SPI_CLK cycles
2	Write the command word	Take SPI_CSB low and clock in the 8 bit command
3	Write the address	Clock in the 16-bit address word
4	Write or read data words	Clock in or out 16-bit data word(s)
5	Termination	Take SPI_CSB high

#### Table 11.10: Instruction Cycle for an SPI Transaction

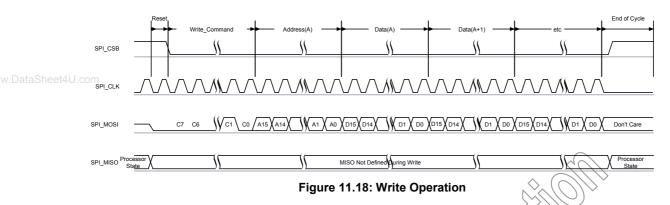
With the exception of reset, SPL\_CSB must be held low during the transaction. Data on SPI\_MOSI is clocked into the BlueCore4-External on the rising edge of the clock line SPI\_CLK. When reading, BlueCore4-External will reply to the master on SPI\_MISO with the data changing on the falling edge of the SPI\_CLK. The master provides the clock on SPI\_CLK. The transaction is terminated by taking SPI\_CSB high.

Sending a command word and the address of a register for every time it is to be read or written is a significant overhead, especially when large amounts of data are to be transferred. To overcome this BlueCore4-External offers increased data transfer efficiency via an auto increment operation. To invoke auto increment, SPI\_CSB is kept low, which auto increments the address, while providing an extra 16 clock cycles for each extra word to be written or read.



#### 11.7.2 Writing to BlueCore4-External

To write to BlueCore4-External, the 8-bit write command (00000010) is sent first (C[7:0]) followed by a 16-bit address (A[15:0]). The next 16-bits (D[15:0]) clocked in on SPI\_MOSI are written to the location set by the address (A). Thereafter for each subsequent 16-bits clocked in, the address (A) is incremented and the data written to consecutive locations until the transaction terminates when SPI\_CSB is taken high.

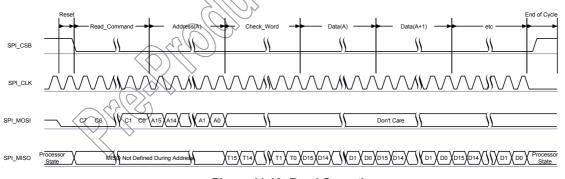


### 11.7.3 Reading from BlueCore4-External

Reading from BlueCore4-External is similar to writing to it. An 8-bit read command (00000011) is sent first (C[7:0]), followed by the address of the location to be read (A[15:0]). BlueCore4-External then outputs on SPI\_MISO a check word during T[15:0] followed by the 16-bit contents of the addressed location during bits D[15:0].

The check word is composed of {command, address [15:8]}. The check word may be used to confirm a read operation to a memory location. This overcomes the problems encountered with typical serial peripheral interface slaves, whereby it is impossible to determine whether the data returned by a read operation is valid data or the result of the slave device not responding.

If SPI\_CSB is kept low, data from consecutive locations is read out on SPI\_MISO for each subsequent 16 clocks, until the transaction terminates when SPI\_CSB is taken high.





### 11.7.4 Multi Slave Operation

BlueCore4-External should not be connected in a multi slave arrangement by simple parallel connection of slave MISO lines. When BlueCore4-External is deselected (SPI\_CSB = 1), the SPI\_MISO line does not float, instead, BlueCore4-External outputs 0 if the processor is running or 1 if it is stopped.



#### 11.8 PCM Interface

Pulse Code Modulation (PCM) is a standard method used to digitise audio (particularly voice) patterns for transmission over digital communication channels. Through its PCM interface, BlueCore4-External has hardware support for continual transmission and reception of PCM data, thus reducing processor overhead for wireless headset applications. BlueCore4-External offers a bi directional digital audio interface that routes directly into the baseband layer of the on chip firmware. It does not pass through the HCI protocol layer.

Hardware on BlueCore4-External allows the data to be sent to and received from a SCO connection.

Up to three SCO connections can be supported by the PCM interface at any one time<sup>(1)</sup>.

www.DataSheet4U BlueCore4-External can operate as the PCM interface Master generating an output clock of 128, 256 or 512kHz. When configured as PCM interface slave it can operate with an input clock up to 2048kHz. BlueCore4-External is compatible with a variety of clock formats, including Long Frame Sync, Short Frame Sync and GCI timing environments.

It supports 13 or 16-bit linear, 8-bit µ-law or A-law companded sample formats at 8ksamples/s and can receive and transmit on any selection of three of the first four slots following PCM\_SYNC. The PCM configuration options are enabled by setting the PS Key PS KEY\_PCM\_CONFIG32 (0x1b3).

BlueCore4-External interfaces directly to PCM audio devices including the following

- Qualcomm MSM 3000 series and MSM 5000 series CDMA baseband devices
- OKI MSM7705 four channel A-law and µ-law CODEC
- Motorola MC145481 8-bit A-law and µ-law CODEC
- Motorola MC145483 13-bit linear CODEC
- STW 5093 and 5094 14-bit linear CODECs
- BlueCore4-External is also compatible with the Motorola SSI<sup>™</sup> interface

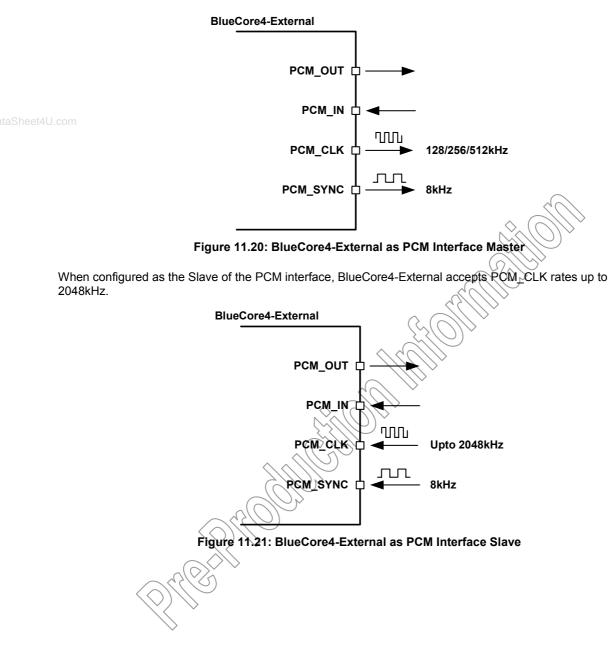
#### Note:

<sup>(1)</sup> Subject to firmware support, contact CSR for current status.



## 11.8.1 PCM Interface Master/Slave

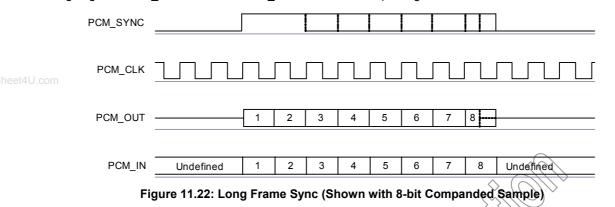
When configured as the Master of the PCM interface, BlueCore4-External generates PCM\_CLK and PCM\_SYNC.





### 11.8.2 Long Frame Sync

Long Frame Sync is the name given to a clocking format that controls the transfer of PCM data words or samples. In Long Frame Sync, the rising edge of PCM\_SYNC indicates the start of the PCM word. When BlueCore4-External is configured as PCM Master, generating PCM\_SYNC and PCM\_CLK, then PCM\_SYNC is 8-bits long. When BlueCore4-External is configured as PCM Slave, PCM\_SYNC may be from two consecutive falling edges of PCM CLK to half the PCM SYNC rate, i.e. 62.5µs long.



BlueCore4-External samples PCM\_IN on the falling edge of PCM\_CLK and transmits PCM\_OUT on the rising edge. PCM\_OUT may be configured to be high impedance on the falling edge of PCM\_CLK in the LSB position or on the rising edge.

### 11.8.3 Short Frame Sync

In Short Frame Sync the falling edge of PCM\_SYNC indicates the start of the PCM word. PCM\_SYNC is always one clock cycle long.

PCM_SYNC							$\leq$	0	$\langle \rangle$	$\searrow$							
PCM_CLK				Д					>								
PCM_OUT		1 2	3	A	5	6	7	8	9	10	11	12	13	14	15	16-	
PCM_IN	Undefined	1 3	3	4	5	6	7	8	9	10	11	12	13	14	15	16	Undefined
Figure 11.23: Short Frame Sync (Shown with 16-bit Sample)																	

As with Long Frame Sync, BlueCore4-External samples PCM\_IN on the falling edge of PCM\_CLK and transmits PCM\_OUT on the rising edge. PCM\_OUT may be configured to be high impedance on the falling edge of PCM\_CLK in the LSB position or on the rising edge.



## 11.8.4 Multi Slot Operation

More than one SCO connection over the PCM interface is supported using multiple slots. Up to three SCO connections can be carried over any of the first four slots.

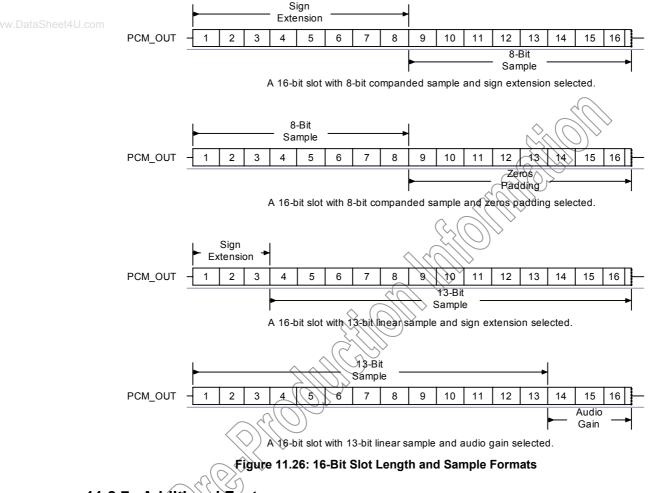
LONG_PCM_SYNC	
Or	
SHORT_PCM_SYNC	
z.DataSheet4U.com PCM_CLK	
PCM_OUT	
PCM_IN	Do Not Care 1 2 3 4 5 6 7 8 1 2 3 4 5 6 7 8 Do Not Care
Figure 11.	24: Multi Slot Operation with Two Slots and 8-bit Companded Samples
11.8.5 GCI Interf	ace
	ompatible with the General Circuit Interface, a standard synchronous 2B+D ISDN timing os B channels can be accessed when this mode is configured.
PCM_SYNC	
РСМ_СLК ]	
PCM_OUT	
PCM_IN Do N Gare	<sup>lot</sup> 1 2 3 4 5 6 7 8 1 2 3 4 5 6 7 8 <sup>Do Not</sup> Care
	B1 Channel B2 Channel
	Figure 11.25: GCI Interface
The start of frame is indi Slave mode, the frequen	cated by the rising edge of PCM_SYNC and runs at 8kHz. With BlueCore4-External in in vor of PCM_CLK can be up to 4.096MHz.
	$\diamond$



## 11.8.6 Slots and Sample Formats

BlueCore4-External can receive and transmit on any selection of the first four slots following each sync pulse. Slot durations can be either 8 or 16 clock cycles. Duration's of 8 clock cycles may only be used with 8-bit sample formats. Durations of 16 clocks may be used with 8, 13 or 16-bit sample formats.

BlueCore4-External supports 13-bit linear, 16-bit linear and 8-bit  $\mu$ -law or A-law sample formats. The sample rate is 8ksamples/s. The bit order may be little or big Endian. When 16-bit slots are used, the 3 or 8 unused bits in each slot may be filled with sign extension, padded with zeros or a programmable 3-bit audio attenuation compatible with some Motorola CODECs.



# 11.8.7 Additional Features

BlueCore4-External has a mute facility that forces PCM\_OUT to be 0. In Master mode, PCM\_SYNC may also be forced to 0 while keeping PCM\_CLK running which some CODECS use to control power down.



## 11.8.8 PCM Timing Information

	Symbol	Para	meter	Min	Тур	Max	Unit
			4MHz DDS generation. Selection of frequency is programmable, see Table 11.13	-	128 256 512	-	kHz
.DataSheet4U.	f <sub>mclk</sub>	PCM_CLK frequency	48MHz DDS generation. Selection of frequency is programmable, see Table 11.14 and Section 11.8.10	2.9		-	kHz
	-	PCM_SYNC frequency		-	8		kHz
	t <sub>mclkh</sub> <sup>(1)</sup>	PCM_CLK high	4MHz DDS generation	980	-	-	ns
	t <sub>mclkl</sub> <sup>(1)</sup>	PCM_CLK low	4MHz DDS generation	730	-		ns
	-	PCM_CLK jitter	48MHz DDS generation		<	21	ns pk-pk
	t <sub>dmclksynch</sub>	Delay time from PCM_CL high	K high to PCM_SYNC	-	<u>}</u>	20	ns
	t <sub>dmclkpout</sub>	Delay time from PCM_CL	K high to valid PCM_OUT	-		20	ns
	t <sub>dmclklsyncl</sub>	Delay time from PCM_CL (Long Frame Sync only)	K low to PCM_SYNC low	-		20	ns
	t <sub>dmclkhsyncl</sub>	Delay time from PCM_CL	K high to PCM_SYNC low		> <u>-</u>	20	ns
	t <sub>dmclklpoutz</sub>	Delay time from PCM_CL impedance	K low to PCM_OUT high		-	20	ns
	t <sub>dmclkhpoutz</sub>	Delay time from PCM_CL impedance	K high to PCM_OUT high	<u> </u>	-	20	ns
	t <sub>supinclkl</sub>	Set-up time for PCM_IN v	alid to PCM_CLK low	30	-	-	ns
	t <sub>hpinclkl</sub>	Hold time for PCM_CLK le	ow to PCM_IN invalid	10	-	-	ns

#### Table 11, 14: PCM Master Timing

Note:

<sup>(1)</sup> Assumes normal system clock operation. Figures will vary during low power modes, when system clock speeds are reduced.



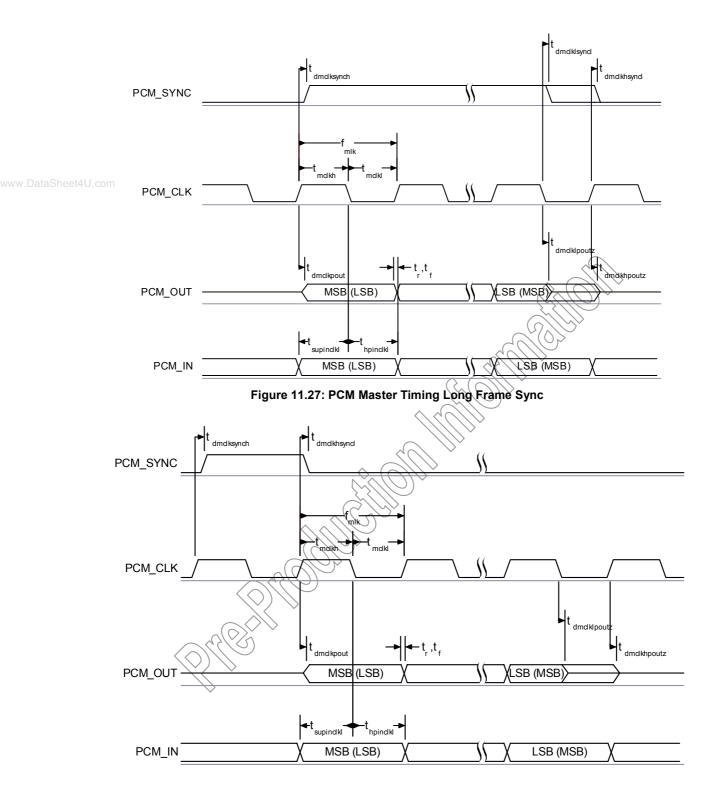


Figure 11.28: PCM Master Timing Short Frame Sync



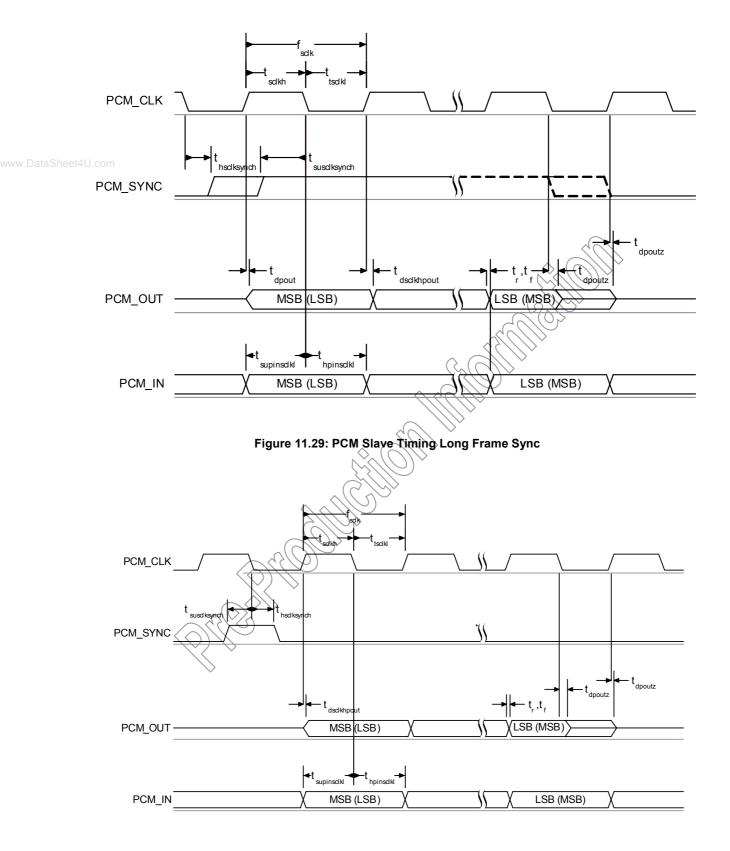
## 11.8.9 PCM Slave Timing

Symbol	Parameter	Min	Тур	Max	Unit
f <sub>sclk</sub>	PCM clock frequency (Slave mode: input)	64	-	2048	kHz
f <sub>sclk</sub>	PCM clock frequency (GCI mode)	128	-	4096	kHz
t <sub>sciki</sub>	PCM_CLK low time	200	-	-	ns
t <sub>sclkh</sub>	t <sub>scikh</sub> PCM_CLK high time			-	ns
t <sub>hsclksynch</sub>	Hold time from PCM_CLK low to PCM_SYNC high	30	-	-	ns
t <sub>susclksynch</sub>	Set-up time for PCM_SYNC high to PCM_CLK low	30	-	-	ns
com t <sub>dpout</sub>	Delay time from PCM_SYNC or PCM_CLK whichever is later, to valid PCM_OUT data (Long Frame Sync only)			20	ns
tdsclkhpout	Delay time from CLK high to PCM_OUT valid data	-	-	20	ns
t <sub>dpoutz</sub> Delay time from PCM_SYNC or PCM_CLK low, whichever is later, to PCM_OUT data line high impedance		-	-	20	ns
t <sub>supinsclkl</sub>	Set-up time for PCM_IN valid to CLK low	30	72-	<u> - </u>	ns
t <sub>hpinsclkl</sub>	Hold time for PCM_CLK low to PCM_IN invalid	30	$\langle (-) \rangle$	>	ns

Table 11.12: PCM Slave Timing

PROOL









## 11.8.10 PCM\_CLK and PCM\_SYNC Generation

BlueCore4-External has two methods of generating PCM\_CLK and PCM\_SYNC in master mode. The first is generating these signals by Direct Digital Synthesis (DDS) from BlueCore4-External internal 4MHz clock. Using this mode limits PCM\_CLK to 128, 256 or 512kHz and PCM\_SYNC to 8kHz. The second is generating PCM\_CLK and PCM\_SYNC by DDS from an internal 48MHz clock which allows a greater range of frequencies to be generated with low jitter but consumes more power. This second method is selected by setting bit '48M\_PCM\_CLK\_GEN\_EN' in PSKEY\_PCM\_CONFIG32. When in this mode and with long frame sync, the length of PCM\_SYNC can be either 8 or 16 cycles of PCM\_CLK, determined by 'LONG\_LENGTH\_SYNC\_EN' in PSKEY\_PCM\_CONFIG32.

The Equation 11.8 describes PCM\_CLK frequency when being generated using the internal 48MHz clock:

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$$f = \frac{CNT\_RATE}{CNT\_LIMIT} \times 24MHz$$

#### Equation 11.8: PCM\_CLK Frequency When Being Generated Using the Internal 48MHz clock

The frequency of PCM\_SYNC relative to PCM\_CLK can be set using following equation:

$$f = \frac{PCM\_CLK}{SYNC \quad LIMIT \times 8}$$

#### Equation 11.9: PCM\_SYNC Frequency Relative to PCM\_CLK

CNT\_RATE, CNT\_LIMIT and SYNC\_LIMIT are set using PSKEY\_PCM\_LOW\_JTTER\_CONFIG. As an example, to generate PCM\_CLK at 512kHz with PCM\_SYNC at 8kHz, set PSKEY\_PCM\_LOW\_JITTER\_CONFIG to 0x08080177.



# 11.8.11 PCM Configuration

The PCM configuration is set using two PS Keys, PSKEY\_PCM\_CONFIG32 and

PSKEY\_PCM\_LOW\_JITTER\_CONFIG. The following tables detail these PS Keys. PSKEY\_PCM\_CONFIG32. The default for this key is 0x00800000 i.e. first slot following sync is active, 13-bit linear voice format, long frame sync and interface master generating 256kHz PCM\_CLK from 4MHz internal clock with no tri-stating of PCM\_OUT. PSKEY\_PCM\_LOW\_JITTER\_CONFIG is described in Table 11.14.

	Name	Bit Position	Description
	-	0	Set to 0.
J.ci	SLAVE_MODE_EN	1	0 selects Master mode with internal generation of PCM_CLK and PCM_SYNC. 1 selects Slave mode requiring externally generated PCM_CLK and PCM_SYNC. This should be set to 1 if 48M_PCM_CLK_GEN_EN (bit 11) is set.
	SHORT_SYNC_EN	2	0 selects long frame sync (rising edge indicates start of frame), 1 selects short frame sync (falling edge indicates start of frame).
	-	3	Set to 0.
	SIGN_EXTEND_EN	4	0 selects padding of 8 or 13-bit voice sample into a 16- bit slot by inserting extra LSBs, 1 selects sign extension. When padding is selected with 13-bit voice sample, the 3 padding bits are the audio gain setting; with 8-bit samples the 8 padding bits are zeroes.
	LSB_FIRST_EN	5	0 transmits and receives voice samples MSB first, 1 uses LSB first.
	TX_TRISTATE_EN	6	0 drives PCM_OUT continuously, 1 tri-states PCM_OUT immediately after the falling edge of PCM_CLK in the last bit of an active slot, assuming the next slot is not active.
	TX_TRISTATE_RISING_EDGE_EN	7	Othistates PCM_OUT immediately after the falling edge of PCM_CLK in the last bit of an active slot, assuming the next slot is also not active. 1 tristates PCM_OUT after the rising edge of PCM_CLK.
	SYNC_SUPPRESS_EN		0 enables PCM_SYNC output when master, 1 suppresses PCM_SYNC whilst keeping PCM_CLK running. Some CODECS utilise this to enter a low power state.
	GCI_MODE_EN	9	1 enables GCI mode.
	MUTE_EN	10	1 forces PCM_OUT to 0.
	48M_PCM_CLK_GEN_EN	11	0 sets PCM_CLK and PCM_SYNC generation via DDS from internal 4 MHz clock, as for BlueCore4-External. 1 sets PCM_CLK and PCM_SYNC generation via DDS from internal 48 MHz clock.
	LONG_LENGTH_SYNC_EN	12	0 sets PCM_SYNC length to 8 PCM_CLK cycles and 1 sets length to 16 PCM_CLK cycles. Only applies for long frame sync and with 48M_PCM_CLK_GEN_EN set to 1.
	-	[20:16]	Set to 0b00000.
	MASTER_CLK_RATE	[22:21]	Selects 128 (0b01), 256 (0b00), 512 (0b10) kHz PCM_CLK frequency when master and 48M_PCM_CLK_GEN_EN (bit 11) is low.
	ACTIVE_SLOT	[26:23]	Default is '0001'. Ignored by firmware.
	SAMPLE_FORMAT	[28:27]	Selects between 13 (0b00), 16 (0b01), 8 (0b10) bit sample with 16 cycle slot duration or 8 (0b11) bit sample with 8 cycle slot duration.

#### Table 11.13: PSKEY\_PCM\_CONFIG32 Description



Name	Bit Position	Description		
CNT_LIMIT	[12:0]	Sets PCM_CLK counter limit.		
CNT_RATE	[23:16]	Sets PCM_CLK count rate.		
SYNC_LIMIT	[31:24]	Sets PCM_SYNC division relative to PCM_CLK.		

#### Table 11.14: PSKEY\_PCM\_LOW\_JITTER\_CONFIG Description

#### 11.9 I/O Parallel Ports

Fifteen lines of programmable bi-directional input/outputs (I/O) are provided. PIO[11:8] and PIO[3:0] are powered from VDD\_PIO. PIO[7:4] are powered from VDD\_PADS. AIO [2:0] are powered from VDD\_MEM.

PIO lines can be configured through software to have either weak or strong pull-ups or pull-downs. All PIO lines are configured as inputs with weak pull-downs at reset.

PIO[0] and PIO[1] are normally dedicated to RXEN and TXEN respectively, but they are available for general use.

Any of the PIO lines can be configured as interrupt request lines or as wake-up lines from sleep modes. PIO[6] or PIO [2] can be configured as a request line for an external clock source. This is useful when the clock to BlueCore4-External is provided from a system application specific integrated circuit (ASIC).

BlueCore4-External has three general purpose analogue interface pins, AIQ[0], AIQ[1] and AIO[2]. These are used to access internal circuitry and control signals. One pin is allocated to accoupling for the on-chip band gap reference voltage, the other three may be configured to provide additional functionality.

Auxiliary functions available via these pins include an 8-bit ADC and at 8-bit DAC. Typically the ADC is used for battery voltage measurement. Signals selectable at these pins include the band gap reference voltage and a variety of clock signals; 48, 24, 16, 8MHz and the XTAL clock frequency. When used with analogue signals the voltage range is constrained by the analogue supply voltage (1.8V). When configured to drive out digital level signals (clocks) generated from within the analogue part of the device, the output voltage level is determined by VDD\_MEM (1.8V).

# 11.9.1 PIO Defaults for BTv1.2 HCI Level Bluetooth Stack

I/O Terminal	Description			
PIO[0]	Pull high on boot up to select USB transport rather than BCSP			
	Control output for external LNA after boot up completion			
PIO[1]	Pull high on boot up to select 16MHz reference clock frequency rather than 26MHz			
G	Control output for external PA (Class 1 operation) after boot up completion			
PIO[2]	Clock request output			
PIO[3]	Clock request "OR" gate input			
PIO[4]	UART Bypass (UART_TX)			
PIO[5]	UART Bypass (UART_RTS)			
PIO[6]	UART Bypass (UART_CTS)			
PIO[7]	UART Bypass (UART_RX)			
AIO[0]	32kHz watchdog input			
AIO[2]	Vref output. Must be decoupled			

#### Table 11.15: PIO Defaults

#### Important Note:

CSR cannot guarantee that these terminal functions remain the same. Please refer to the software release note for the implementation of these PIO lines, as they are firmware build specific.



# 11.10 I<sup>2</sup>C Interface

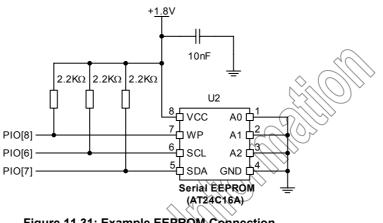
PIO[8:6] can be used to form a Master I<sup>2</sup>C interface. The interface is formed using software to drive these lines. Therefore it is suited only to relatively slow functions such as driving a dot matrix liquid crystal display (LCD), keyboard scanner or EEPROM.

Note:

PIO[7:6] dual functions, UART bypass and EEPROM support, therefore devices using an EEPROM cannot support UART bypass mode

PIO lines need to be pulled-up through  $2.2k\Omega$  resistors.

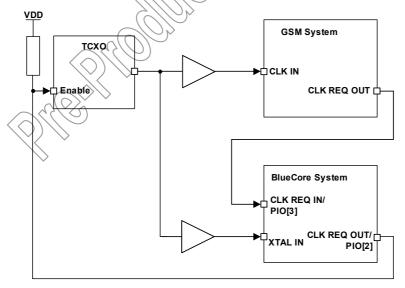
For connection to EEPROMs, refer to CSR documentation on I<sup>2</sup>C EEPROMS for use with BlueCore. This provides information on the type of devices which are currently supported.



#### Figure 11.31: Example EEPROM Connection

#### **TCXO Enable OR Function** 11.11

An OR function exists for clock enable signals from a host controller and BlueCore4-External where either device can turn on the clock without having to wake up the other device. PIO[3] can be used as the Host clock enables input and PIO[2] can be used as the OR output with the TCXO enable signal from BlueCore4-External.





On reset and up to the time the PIO has been configured, PIO[2] will be tri-stated. Therefore, the developer must ensure that the circuitry connected to this pin is pulled via a 470k $\Omega$  resistor to the appropriate power rail. This ensures that the TCXO is oscillating at start up.



#### 11.12 **RESETB**

BlueCore4-External may be reset from several sources: RESETB pin, power on reset, a UART break character or via a software configured watchdog timer.

The RESETB pin is an active low reset and is internally filtered using the internal low frequency clock oscillator. A reset will be performed between 1.5 and 4.0ms following RESETB being active. It is recommended that RESETB be applied for a period greater than 5ms.

The power on reset occurs when the VDD\_CORE supply falls below typically 1.5V and is released when VDD\_CORE rises above typically 1.6V.

At reset the digital I/O pins are set to inputs for bi-directional pins and outputs are tri-stated. The PIOs have weak pull-downs.

Following a reset, BlueCore4-External assumes the maximum XTAL\_IN frequency, which ensures that the internal clocks run at a safe (low) frequency until BlueCore4-External is configured for the actual XTAL\_IN frequency. If no clock is present at XTAL\_IN, the oscillator in BlueCore4-External free runs, again at a safe frequency.

### 11.12.1 Pin States on Reset

Pin Name	State: BlueCore4-External					
PIO[11:0]	Input with weak pull-down					
PCM_OUT	Tri-stated with weak pull-down					
PCM_IN	Input with weak pull-down					
PCM_SYNC	Input with weak pull-down					
PCM_CLK	Input with weak pull-down					
UART_TX	Output tri-stated with weak pull-up					
UART_RX	Input with weak pull-down					
UART_RTS	Dutput tri-stated with weak pull-up					
UART_CTS	Input with weak pull <sub>t</sub> down					
USB_DP	Input with weak pull-down					
USB_DN	Input with weak pull-down					
SPI_CSB	Input with weak pull-up					
SPI_CLK	Input with weak pull-down					
SPI_MOSI	Imput with weak pull-down					
SPI_MISO	Output tri-stated with weak pull-down					
AIO[2:0]	Output, driving low					
RESET	Input with weak pull-down					
RESETB	Input with weak pull-up					
TEST_EN	Input with strong pull-down					
TX_A	High impedance					
TX_B	High impedance					
RF_IN	High impedance					
XTAL_IN	High impedance, 250k to XTAL_OUT					
XTAL_OUT	High impedance, 250k to XTAL_IN					

Table 11.16 shows the pin states of BlueCore4-External on reset.

Table 11.16: Pin States of BlueCore4-External on Reset

#### 11.12.2 Status after Reset

The chip status after a reset is as follows:

- Warm Reset: Baud rate and RAM data remain available
- Cold Reset<sup>(1)</sup>: Baud rate and RAM data not available

#### Note:

- <sup>(1)</sup> Cold Reset consititutes one of the following:
  - Power cycle
  - System reset (firmware fault code)
    - Reset signal, see Section 11.12

#### 11.13 Power Supply

#### 11.13.1 Voltage Regulator

An on-chip linear voltage regulator can be used to power the 1.8V dependent supplies. It is advised that a smoothing circuit using a  $2.2\mu$ F low ESR capacitor and  $2.2\Omega$  resistor be placed on the output VDD\_ANA adjacent to VREG\_IN.

The regulator is switched into a low power mode when the device is sent into deep sleep mode. When the on chip regulator is not required VDD\_ANA is a 1.8V input and VREG\_IN must be either open circuit or tied to VDD\_ANA.

An on chip linear voltage regulator can be used to power the 1.8V dependent supplies. The regulator is switched into a low power mode when the device is sent into deep sleep mode. When the on chip regulator is not required VDD\_ANA is a 1.8V input and VREG\_IN must be either open circuit or tied to VDD\_ANA.

#### 11.13.2 Sequencing

It is recommended that VDD\_CORE, VDD\_RADIO, VDD\_LO and VDD\_ANA be powered at the same time. The order of powering supplies for VDD\_CORE, VDD\_PIO, VDD\_PADS and VDD\_USB is not important. However if VDD\_CORE is not present, all inputs have a weak pull-down irrespective of the reset state.

### 11.13.3 Sensitivity to Disturbances

It is recommended that if you are supplying BlueCore4-External from an external voltage source that VDD\_LO, VDD\_ANA and VDD\_RADIO should have less than 10mV rms noise levels between 0 to 10MHz. Single tone frequencies are also to be avoided. A simple RC filter is recommended for VDD\_CORE as this reduces transients put back onto the power supply rails

The remaining supplies VDD\_MEM, VDD\_PIO, VDD\_PADS and VDD\_USB can be connected together with the VREG\_IN to the 3.3V supply and simply decoupled as shown in Figure 12.1.

The transient response of the regulator is also important. At the start of a packet, power consumption will jump to high levels, see average current consumption section. The regulator should have a response time of  $20\mu s$  or less, it is essential that the power rail recovers quickly.



# **12** Application Schematic

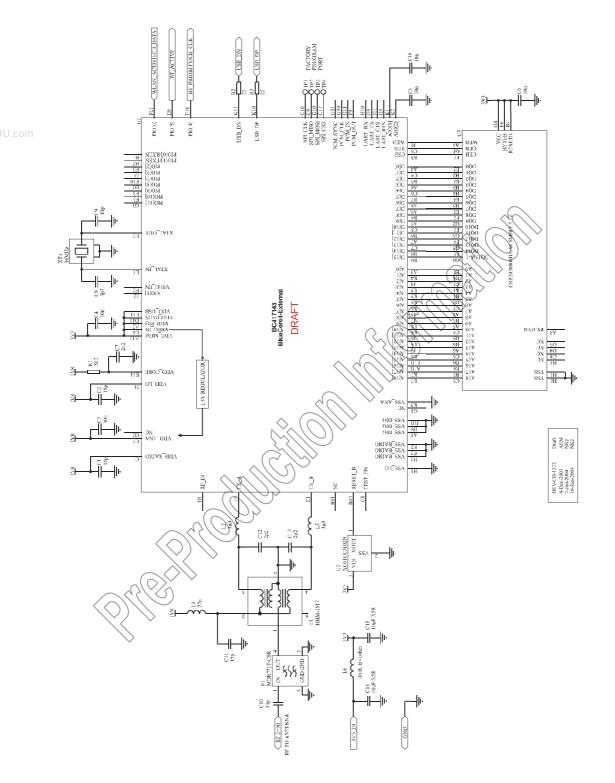


Figure 12.1: Application Circuit for Radio Characteristics Specification with 8 x 8mm VFBGA Package



#### Notes:

Circuit is for PC application circuit

C14, L4, C15  $\pi$  filter is for potentially noisy PC motherboard power supply. Less filtering may be necessary depending on actual application

U2 voltage detector holds BlueCore4-External in reset if voltage is low enough for 3.3V flash operation to be unreliable. Not required if 1.8V flash used

Various alternative components can be used in the place of discrete RF matching circuit, balun and filter. Please refer to CSR example designs

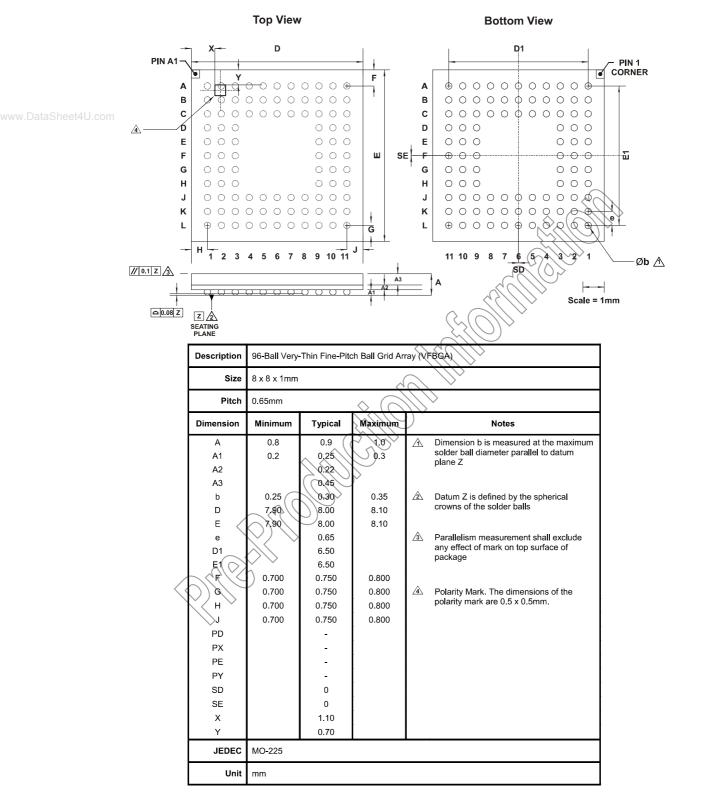
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# 13 Package Dimensions

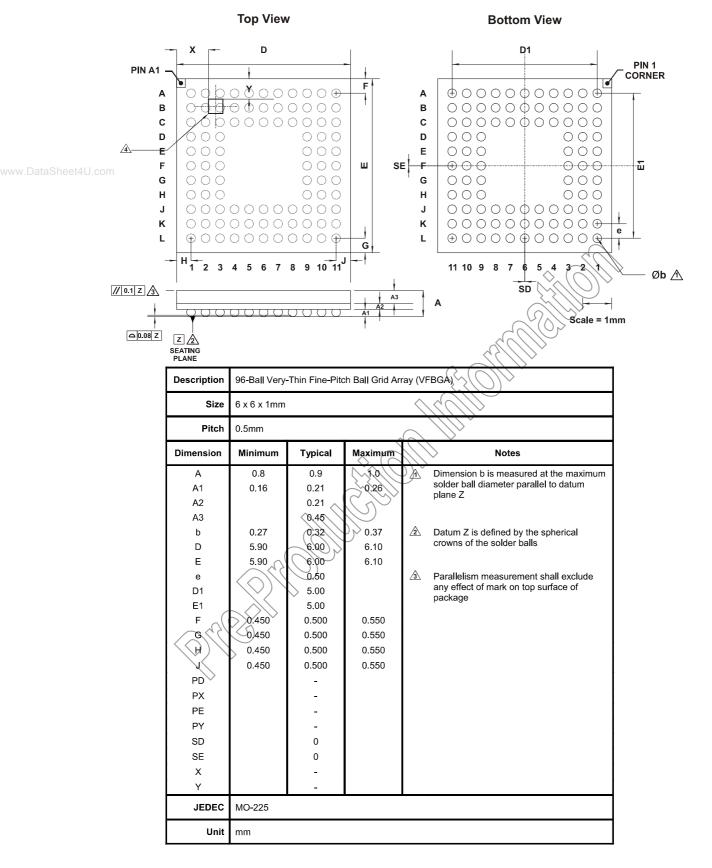
## 13.1 8 x 8mm VFBGA 96-Ball Package







#### 13.2 6 x 6mm VFBGA 96-Ball Package



BlueCore<sup>m</sup>4-External Product Data Shee



# **14 Solder Profiles**

The soldering profile depends on various parameters necessitating a set up for each application. The data here is given only for guidance on solder re-flow. There are four zones:

- 1. **Preheat Zone** This zone raises the temperature at a controlled rate, typically 1-2.5°C/s.
- 2. **Equilibrium Zone** This zone brings the board to a uniform temperature and also activates the flux. The duration in this zone (typically 2-3 minutes) will need to be adjusted to optimise the out gassing of the flux.
- 3. **Reflow Zone** The peak temperature should be high enough to achieve good wetting but not so high as to cause component discoloration or damage. Excessive soldering time can lead to intermetallic growth which can result in a brittle joint.
- 4. **Cooling Zone** The cooling rate should be fast, to keep the solder grains small which will give a longer lasting joint. Typical rates will be 2-5°C/s.

#### 14.1 Solder Re-Flow Profile for Devices with Lead-Free Solder Balls

Composition of the solder ball: Sn 95.5%, Ag 4.0%, Cu 0.5%

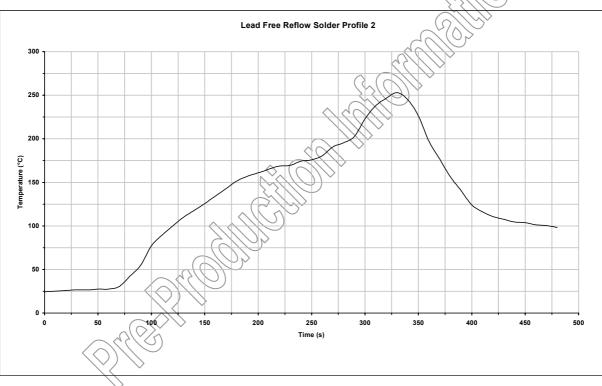


Figure 14.1: Typical Lead-Free Re-flow Solder Profile

Key features of the profile:

- Initial Ramp = 1-2.5°C/sec to 175°C±25°C equilibrium
- Equilibrium time = 60 to 180 seconds
- Ramp to Maximum temperature (250°C) = 3°C/sec max.
- Time above liquidus temperature (217°C): 45-90 seconds
- Device absolute maximum reflow temperature: 260°C

Devices will withstand the specified profile. Lead-free devices will withstand up to 3 reflows to a maximum temperature of 260°C.



# **15 Ordering Information**

#### 15.1 **BlueCore4-External**

Interface		Package		Order Number		
Version	Туре	Size	Shipment Method	Order Number		
UART and USB	96-Ball VFBGA (Pb free)	8 x 8 x 1mm	Tape and reel	BC417143B-ES-IQN-E <sup>(1)</sup>		
UART and USB	96-Ball VFBGA (Pb free)	6 x 6 x 1mm	Tape and reel	BC417143B-ES-IRN-E <sup>(1)</sup>		

#### Maximum Order Engineering Sample Quantity

PROCEED.

2kpcs Taped and Reeled

Note:

When BlueCore4-External reaches Production status order numbers will be BC417143B-IQN-E4 and (1) BC417143B-IRN-E4.

#### **Minimum Order Production Quantity**

2kpcs Taped and Reeled

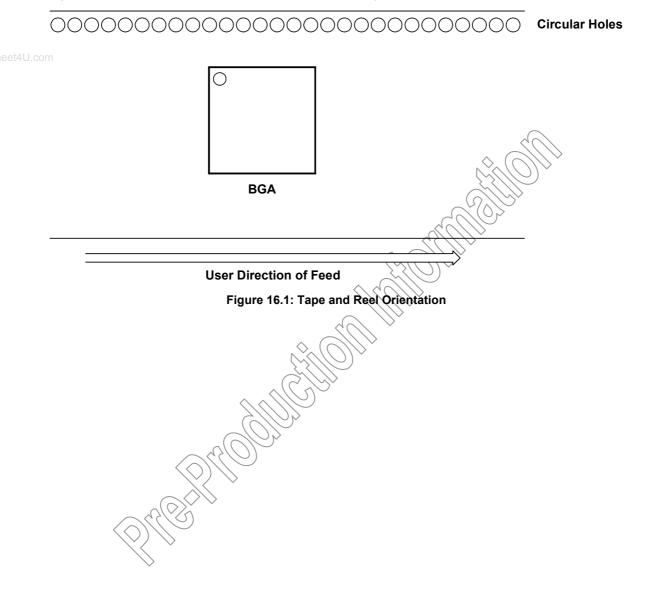


# 16 Tape and Reel Information

Tape and reel is in accordance with EIA-481-2.

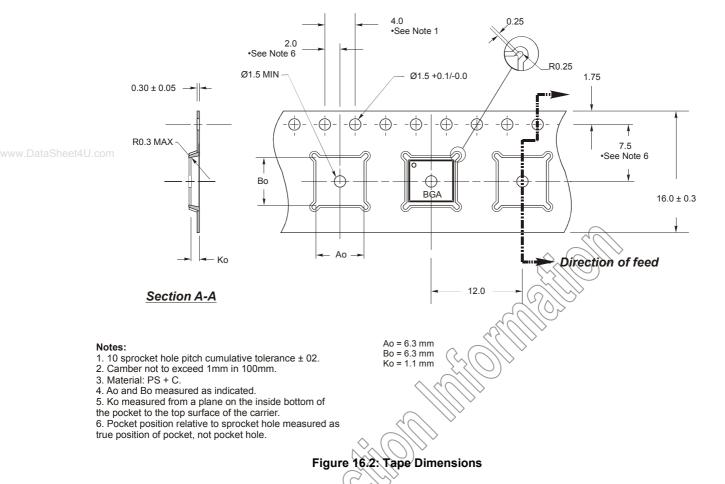
#### **16.1** Tape Orientation and Dimensions

The general orientation of the BGA in the tape is as shown in Figure 16.1.





As a detailed example, the diagram shown in Figure 16.2 outlines the dimensions of the tape used for 8x8x1mm VFBGA devices:



The cover tape has a total peel strength of 0.11 to 1.3N. The direction of the pull should be opposite the direction of the carrier tape such that the cover tape makes an angle of between 165° and 180° with the top of the carrier tape. The carrier and/or cover tape should be pulled with a velocity of 300±10mm during peeling.

Maximum component rotation inside the cavity is  $10^{\circ}$  in accordance with EIA-481-2. The cavity pitch tolerance (dimension P1) is  $\pm 0.1$ mm.

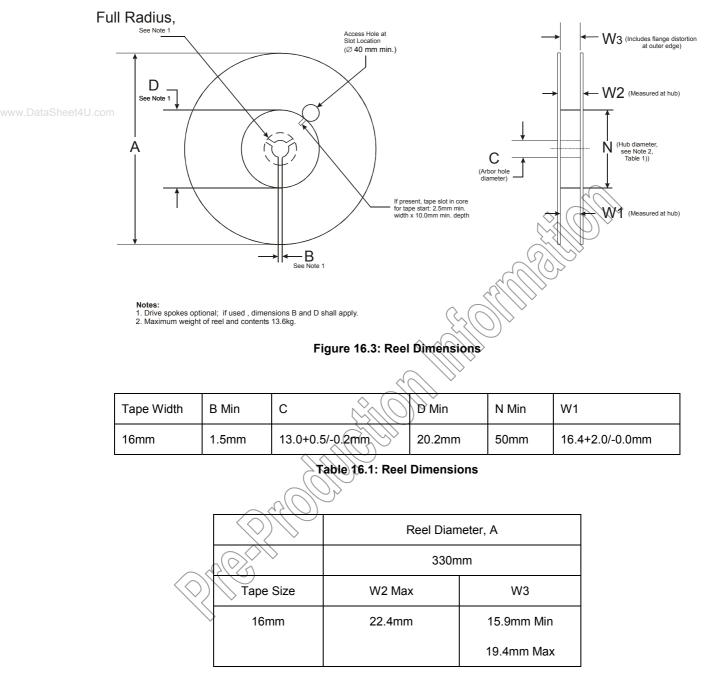
The reel is made of high impact injection moulded polystyrene. The carrier tape is made of polystyrene with carbon. The cover tape is made of antistatic polyester film and an antistatic heat activated adhesive coating.



## 16.2 Reel Information

#### Reel dimensions

(All dimensions in millimeters)



#### **Table 16.2: Diameter Dependent Dimensions**

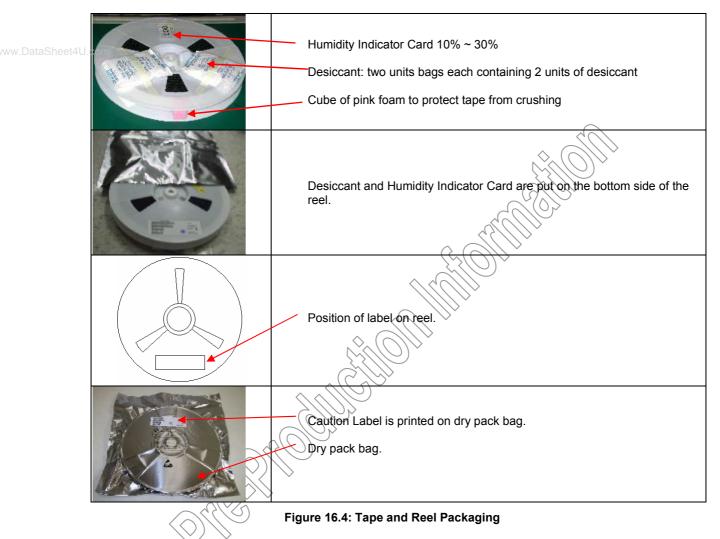


## 16.3 Dry Pack Information

The primary packed product is dry packed in accordance with Joint IPC / JEDEC J-STD-033.

All materials used in dry packing conform to EIA-541 and EIA-583.

Some illustrative views of reel dry packs are shown in Figure 16.4.



Devices shipped in dry-pack bags will withstand storage in normal environmental conditions, such as 30°C and 70% RH for a minimum of one year as long as the dry-pack bag has not become punctured. Humidity indicators inside the dry-pack bag will confirm this when the bag is opened.



## 16.4 Baking Conditions

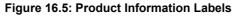
Devices may, if necessary, be re-baked at 125°C for 24 hours. If devices are still on the reel, which cannot withstand such high temperatures, they should be baked at 45°C for 192 hours at relative humidity less than 5%.

Solder wettability of parts will be unaffected by three such bakes.

#### **16.5 Product Information**

Example product information labels are shown is Figure 16.5.





A product information label is placed on each reel, primary package and shipment package



# **17 Contact Information**

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To contact a CSR representative, go to http://www.csr.com/contacts.htm



# **18 Document References**

Document:	Reference, Date:
Specification of the Bluetooth System	v1.2, 05 November 2003
Universal Serial Bus Specification	v2.0, 27 April 2000
Selection of I <sup>2</sup> C EEPROMS for Use with BlueCore	bcore-an-008Pb, 30 September 2003
EDR RF Test Specification v0.9	v0.90, D07r22, 16 March 2004
RF Prototyping Specification for Enhanced Data Rate IP	v.90, r29, 2004

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# Acronyms and Definitions

8DPSK	8 phase Differential Phase Shift Keying		
π/4 DQPSK	pi/4 rotated Differential Quaternary Phase Shift Keying		
BlueCore™	Group term for CSR's range of Bluetooth chips		
Bluetooth™	Set of technologies providing audio and data transfer over short-range radio connections		
CSR	Cambridge Silicon Radio		
ACL	Asynchronous Connection-Less. A Bluetooth data packet.		
ADC	Analogue to Digital Converter		
AGC	Automatic Gain Control		
A-law	Audio encoding standard		
API	Application Programming Interface		
ASIC	Application Specific Integrated Circuit		
BCSP	BlueCore™ Serial Protocol		
BER	Bit Error Rate. Used to measure the quality of a link		
BIST	Built-In Self-Test		
BMC	Burst Mode Controller		
CMOS	Complementary Metal Oxide Semiconductor		
CODEC	Coder Decoder		
CSB	Chip Select (Active Low)		
CSR	Cambridge Silicon Radio		
CTS	Clear to Send		
CVSD	Continuous Variable Slope Delta Modulation		
DAC	Digital to Analogue Converter		
dBm	Decibels relative to 1mW		
DC	Direct Current		
DEVM	Differential Error Vector Magnitude		
DFU	Device Firmware Upgrade		
DPSK	Differential Phase Shift Keying		
DQPSK	Differential Quaternary Phase Shift Keying		
ESR	Equivalent Series Resistance		
FSK	Frequency Shift Keying		
GSM	Global System for Mobile communications		
нсі	Host Controller Interface		
IQ Modulation	In-Phase and Quadrature Modulation		
IF	Intermediate Frequency		
IIR	Infinite Impulse Response		
ISDN	Integrated Services Digital Network		
ISM	Industrial, Scientific and Medical		
ksps	KiloSamples Per Second		
L2CAP	Logical Link Control and Adaptation Protocol (protocol layer)		
LC	Link Controller		
LCD	Liquid Crystal Display		
LFBGA	Low profile Fine Ball Grid Array		
LNA	Low Noise Amplifier		



Г		
-	LPF	Low Pass Filter
-	LSB	Least-Significant Bit
F	μ-law	Audio Encoding Standard
-	MCU	MicroController Unit
-	MMU	Memory Management Unit
	MISO	Master In Serial Out
	MOSI	Master Out Slave In
	Mbps	Mega bits per second
	OHCI	Open Host Controller Interface
ww.DataSheet4U.c	o <b>PA</b>	Power Amplifier
-	PCM	Pulse Code Modulation. Refers to digital voice data
	PIO	Parallel Input Output
	PLL	Phase Lock Loop
-	ppm	parts per million
	PS Key	Persistent Store Key
	RAM	Random Access Memory
	REB	Read enable (Active Low)
	REF	Reference. Represents dimension for reference use only
	RF	Radio Frequency
-	RFCOMM	Protocol layer providing serial port emulation over L2CAP
-	RISC	Reduced Instruction Set Computer
	rms	root mean squared
	RoHS	The Restriction of Hazardous Substances in Electrical and Electronic Equipment Directive (2002/95/EC)
-	RSSI	Receive Signal Strength Indication
-	RTS	Ready To Send
-	RX	Receive or Receiver
	SCO	Synchronous Connection-Oriented. Voice oriented Bluetooth packet
	SD	Secure Digital
	SDK	Software Development Kit
	SDP	Service Discovery Protocol
-	SPI	Serial Reripheral Interface
	ТВА	To Be Announced
	ТВD	To Be Defined
	тх	Transmit or Transmitter
	UART	Universal Asynchronous Receiver Transmitter
	USB	Universal Serial Bus or Upper Side Band (depending on context)
ľ	VCO	Voltage Controlled Oscillator
	VFBGA	Very Fine Ball Grid Array
ľ	VM	Virtual Machine
1	W-CDMA	Wideband Code Division Multiple Access
-	WEB	Write Enable (Active Low)
-	www	world wide web
L		



# **Record of Changes**

Date:	Revision	Reason for Change:
03 JUN 04	а	Original publication of Pre-Production Information Product Data Sheet (CSR reference: BC417143B-ds-001Pa)
15 JUN 04	b	Numbering changes made to AIO pins.
06 SEPT 04	С	6 x 6mm package option added to data sheet and AUX DAC removed.

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# BlueCore™4-External

# **Product Data Sheet**

# BC417143B-ds-001Pc

# September 2004